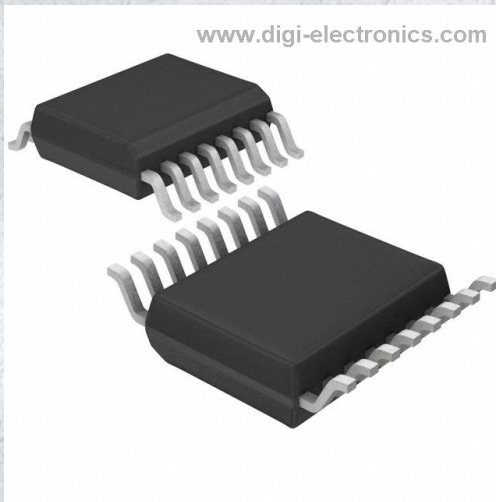


# SA572DTB Datasheet



<https://www.DiGi-Electronics.com>

DiGi Electronics Part Number	SA572DTB-DG
Manufacturer	<a href="#">onsemi</a>
Manufacturer Product Number	SA572DTB
Description	IC COMPANDOR 16TSSOP
Detailed Description	Audio Compandor 2 Channel 16-TSSOP



Tel: +00 852-30501935

RFQ Email: [Info@DiGi-Electronics.com](mailto:Info@DiGi-Electronics.com)

DiGi is a global authorized distributor of electronic components.

## Purchase and inquiry

Manufacturer Product Number:

SA572DTB

Series:

-

Function:

Companzor

Number of Channels:

2

Voltage - Supply:

6V ~ 22V

Specifications:

110dB

Package / Case:

16-TSSOP (0.173", 4.40mm Width)

Base Product Number:

SA572

Manufacturer:

onsemi

Product Status:

Obsolete

Applications:

Consumer Audio, Telecommunication Systems

Interface:

Analog

Operating Temperature:

-40°C ~ 85°C (TA)

Mounting Type:

Surface Mount

Supplier Device Package:

16-TSSOP

## Environmental & Export classification

RoHS Status:

RoHS non-compliant

REACH Status:

REACH Unaffected

HTSUS:

8542.39.0001

Moisture Sensitivity Level (MSL):

1 (Unlimited)

ECCN:

EAR99



# Programmable Analog Compandor

## SA572

The SA572 is a dual-channel, high-performance gain control circuit in which either channel may be used for dynamic range compression or expansion. Each channel has a full-wave rectifier to detect the average value of input signal, a linearized, temperature-compensated variable gain cell ( $\Delta G$ ) and a dynamic time constant buffer. The buffer permits independent control of dynamic attack and recovery time with minimum external components and improved low frequency gain control ripple distortion over previous compandors.

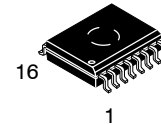
The SA572 is intended for noise reduction in high-performance audio systems. It can also be used in a wide range of communication systems and video recording applications.

### Features

- Independent Control of Attack and Recovery Time
- Improved Low Frequency Gain Control Ripple
- Complementary Gain Compression and Expansion with External Op Amp
- Wide Dynamic Range – Greater than 110 dB
- Temperature-Compensated Gain Control
- Low Distortion Gain Cell
- Low Noise – 6.0  $\mu\text{V}$  Typical
- Wide Supply Voltage Range – 6.0 V-22 V
- System Level Adjustable with External Components
- This is a Pb-Free Device\*

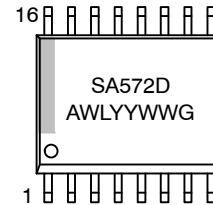
### Applications

- Dynamic Noise Reduction System
- Voltage Control Amplifier
- Stereo Expander
- Automatic Level Control
- High-Level Limiter
- Low-Level Noise Gate
- State Variable Filter



**SOIC-16 WB**  
**D SUFFIX**  
**CASE 751G**

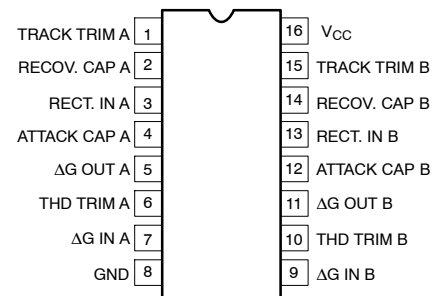
### MARKING DIAGRAM



A = Assembly Location  
 WL = Wafer Lot  
 YY = Year  
 WW = Work Week  
 ■ = Pb-Free Package

(Note: Microdot may be in either location)

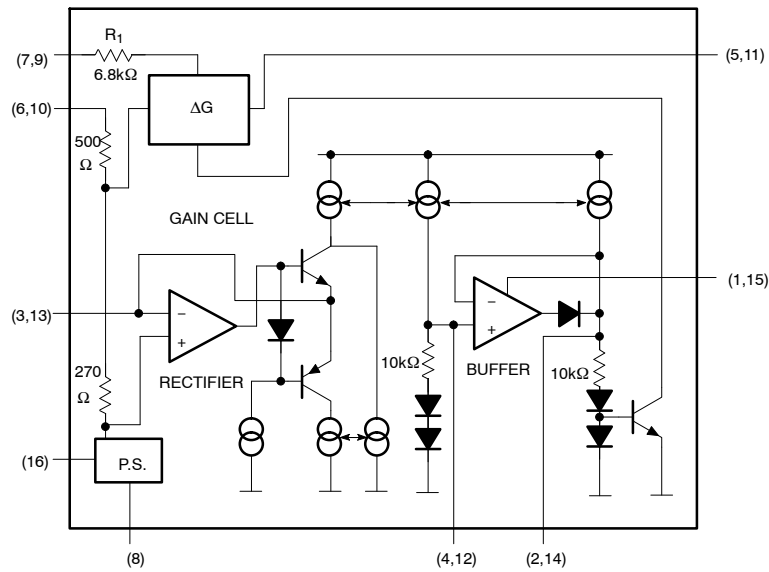
### PIN CONNECTIONS



### ORDERING INFORMATION

See detailed ordering and shipping information in the package dimensions section on page 10 of this data sheet.

\*For additional information on our Pb-Free strategy and soldering details, please download the [onsemi Soldering and Mounting Techniques Reference Manual, SOLDERRM/D](#).

**SA572****Figure 1. Block Diagram****PIN FUNCTION DESCRIPTION**

Pin	Symbol	Description
1	TRACK TRIM A	Tracking Trim A
2	RECOV. CAP A	Recovery Capacitor A
3	RECT. IN A	Rectifier A Input
4	ATTACK CAP A	Attack Capacitor A
5	$\Delta G$ OUT A	Variable Gain Cell A Output
6	THD TRIM A	Total Harmonic Distortion Trim A
7	$\Delta G$ IN A	Variable Gain Cell A Input
8	GND	Ground
9	$\Delta G$ IN B	Variable Gain Cell B Input
10	THD TRIM B	Total Harmonic Distortion Trim B
11	$\Delta G$ OUT B	Variable Gain Cell B Output
12	ATTACK CAP B	Attack Capacitor B
13	RECT. IN B	Rectifier B Input
14	RECOV. CAP B	Recovery Capacitor B
15	TRACK TRIM B	Tracking Trim B
16	$V_{CC}$	Positive Power Supply



## SA572

### Audio Signal Processing IC Combines VCA and Fast Attack/Slow Recovery Level Sensor

In high-performance audio gain control applications, it is desirable to independently control the attack and recovery time of the gain control signal. This is true, for example, in compandor applications for noise reduction. In high end systems the input signal is usually split into two or more frequency bands to optimize the dynamic behavior for each band. This reduces low frequency distortion due to control signal ripple, phase distortion, high frequency channel overload and noise modulation. Because of the expense in hardware, multiple band signal processing up to now was limited to professional audio applications.

With the introduction of the SA572 this high-performance noise reduction concept becomes feasible for consumer hi fi applications. The SA572 is a dual channel gain control IC. Each channel has a linearized, temperature-compensated gain cell and an improved level sensor. In conjunction with an external low noise op amp for current-to-voltage conversion, the VCA features low distortion, low noise and wide dynamic range.

The novel level sensor which provides gain control current for the VCA gives lower gain control ripple and independent control of fast attack, slow recovery dynamic response. An attack capacitor  $C_A$  with an internal  $10\text{ k}\Omega$  resistor  $R_A$  defines the attack time  $\tau_A$ . The recovery time  $\tau_R$  of a tone burst is defined by a recovery capacitor  $C_R$  and an internal  $10\text{ k}\Omega$  resistor  $R_R$ . Typical attack time of  $4.0\text{ ms}$  for the high-frequency spectrum and  $40\text{ ms}$  for the low frequency band can be obtained with  $0.1\text{ }\mu\text{F}$  and  $1.0\text{ }\mu\text{F}$  attack capacitors, respectively. Recovery time of  $200\text{ ms}$  can be obtained with a  $4.7\text{ }\mu\text{F}$  recovery capacitor for a  $100\text{ Hz}$  signal, the third harmonic distortion is improved by more than  $10\text{ dB}$  over the simple RC ripple filter with a single  $1.0\text{ }\mu\text{F}$  attack and recovery capacitor, while the attack time remains the same.

The SA572 is assembled in a standard 16-pin dual in-line plastic package and in oversized SOL package. It operates over a wide supply range from  $6.0\text{ V}$  to  $22\text{ V}$ . Supply current is less than  $6.0\text{ mA}$ . The SA572 is designed for applications from  $-40^\circ\text{C}$  to  $+85^\circ\text{C}$ .

## BASIC APPLICATIONS

### Description

The SA572 consists of two linearized, temperature-compensated gain cells ( $\Delta G$ ), each with a full-wave rectifier and a buffer amplifier as shown in the block diagram. The two channels share a  $2.5\text{ V}$  common bias reference derived from the power supply but otherwise operate independently. Because of inherent low distortion, low noise and the capability to linearize large signals, a wide dynamic range can be obtained. The buffer amplifiers are provided to permit control of attack time and recovery time independent of each other. Partitioned as shown in the block diagram, the IC allows flexibility in the design of system levels that optimize DC shift, ripple distortion, tracking accuracy and noise floor for a wide range of application requirements.

### Gain Cell

Figure 3 shows the circuit configuration of the gain cell. Bases of the differential pairs  $Q_1$ - $Q_2$  and  $Q_3$ - $Q_4$  are both tied to the output and inputs of OPA  $A_1$ . The negative feedback through  $Q_1$  holds the  $V_{BE}$  of  $Q_1$ - $Q_2$  and the  $V_{BE}$  of  $Q_3$ - $Q_4$  equal. The following relationship can be derived from the transistor model equation in the forward active region.

$$\Delta V_{BE_{Q3Q4}} = \Delta V_{BE_{Q1Q2}}$$

$$(V_{BE} = V_T \ln IC/IS)$$

$$V_{Tn} \ln \left( \frac{\frac{1}{2}I_G + \frac{1}{2}I_O}{I_S} \right) - V_{Tn} \ln \left( \frac{\frac{1}{2}I_G - \frac{1}{2}I_O}{I_S} \right) \quad (\text{eq. 1})$$

$$= V_{Tn} \ln \left( \frac{I_1 + I_{IN}}{I_S} \right) - V_{Tn} \ln \left( \frac{I_2 - I_1 - I_{IN}}{I_S} \right)$$

$$\text{where } I_{IN} = \frac{V_{IN}}{R_1}$$

$$R_1 = 6.8\text{ k}\Omega$$

$$I_1 = 140\text{ }\mu\text{A}$$

$$I_2 = 280\text{ }\mu\text{A}$$

$I_O$  is the differential output current of the gain cell and  $I_G$  is the gain control current of the gain cell.

If all transistors  $Q_1$  through  $Q_4$  are of the same size, equation 1 can be simplified to:

$$I_O = \frac{2}{I_2} \cdot I_{IN} \cdot I_G - \frac{1}{I_2} (I_2 - 2I_1) \cdot I_G \quad (\text{eq. 2})$$

The first term of equation 2 shows the multiplier relationship of a linearized two quadrant transconductance amplifier. The second term is the gain control feedthrough due to the mismatch of devices. In the design, this has been minimized by large matched devices and careful layout. Offset voltage is caused by the device mismatch and it leads to even harmonic distortion. The offset voltage can be trimmed out by feeding a current source within  $\pm 25\text{ }\mu\text{A}$  into the THD trim pin.

## SA572

The residual distortion is third harmonic distortion and is caused by gain control ripple. In a compandor system, available control of fast attack and slow recovery improve ripple distortion significantly. At the unity gain level of 100 mV, the gain cell gives THD (total harmonic distortion) of 0.17% typ. Output noise with no input signals

is only 6.0  $\mu\text{V}$  in the audio spectrum (10 Hz-20 kHz). The output current  $I_O$  must feed the virtual ground input of an operational amplifier with a resistor from output to inverting input. The non-inverting input of the operational amplifier has to be biased at  $V_{\text{REF}}$  if the output current  $I_O$  is DC coupled.

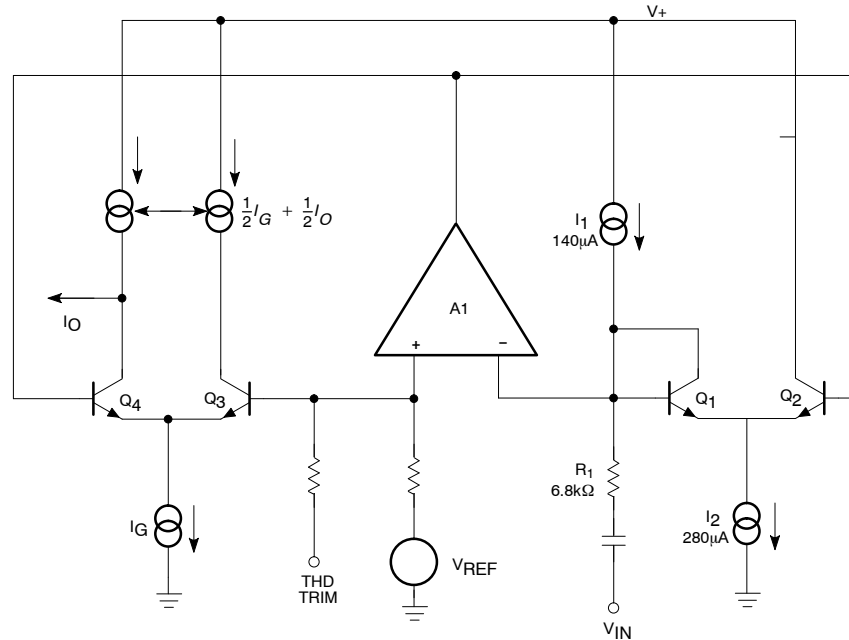


Figure 3. Basic Gain Cell Schematic

### Rectifier

The rectifier is a full-wave design as shown in Figure 4. The input voltage is converted to current through the input resistor  $R_2$  and turns on either  $Q_5$  or  $Q_6$  depending on the signal polarity. Deadband of the voltage to current converter is reduced by the loop gain of the gain block  $A_2$ . If AC coupling is used, the rectifier error comes only from input bias current of gain block  $A_2$ . The input bias current is typically about 70 nA. Frequency response of the gain block  $A_2$  also causes second-order error at high frequency. The collector current of  $Q_6$  is mirrored and summed at the collector of  $Q_5$  to form the full wave rectified output current  $I_R$ . The rectifier transfer function is:

$$I_R = \frac{V_{\text{IN}} - V_{\text{REF}}}{R_2} \quad (\text{eq. 3})$$

If  $V_{\text{IN}}$  is AC-coupled, then the equation will be reduced to:

$$I_{\text{RAC}} = \frac{V_{\text{IN}}(\text{AVG})}{R_2}$$

The internal bias scheme limits the maximum output current  $I_R$  to be around 300  $\mu\text{A}$ . Within a  $\pm 1.0$  dB error band the input range of the rectifier is about 52 dB.

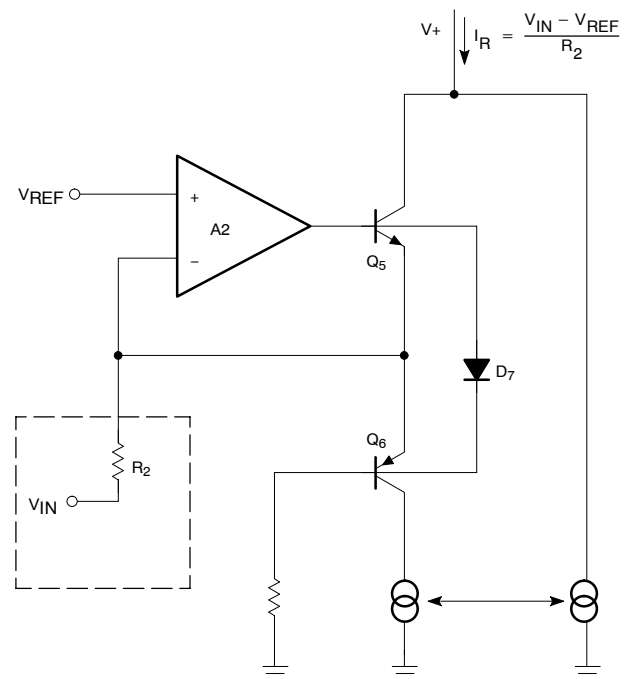


Figure 4. Simplified Rectifier Schematic

## SA572

## Buffer Amplifier

In audio systems, it is desirable to have fast attack time and slow recovery time for a tone burst input. The fast attack time reduces transient channel overload but also causes low-frequency ripple distortion. The low-frequency ripple distortion can be improved with the slow recovery time. If different attack times are implemented in corresponding frequency spectrums in a split band audio system, high quality performance can be achieved. The buffer amplifier is designed to make this feature available with minimum external components. Referring to Figure 5, the rectifier output current is mirrored into the input and output of the unipolar buffer amplifier A<sub>3</sub> through Q<sub>8</sub>, Q<sub>9</sub> and Q<sub>10</sub>. Diodes D<sub>11</sub> and D<sub>12</sub> improve tracking accuracy and provide common-mode bias for A<sub>3</sub>. For a positive-going input signal, the buffer amplifier acts like a voltage-follower. Therefore, the output impedance of A<sub>3</sub> makes the contribution of capacitor C<sub>R</sub> to attack time insignificant. Neglecting diode impedance, the gain G<sub>A</sub>(t) for ΔG can be expressed as follows:

$$G_A(t) = (G_{A_{INT}} - G_{A_{FNL}}) e^{-\frac{t}{\tau_A}} + G_{A_{FNL}}$$

G<sub>A<sub>INT</sub></sub> = Initial Gain

G<sub>A<sub>FNL</sub></sub> = Final Gain

$$\tau_A = R_A \cdot C_A = 10 \text{ k}\Omega \cdot C_A$$

where τ<sub>A</sub> is the attack time constant and R<sub>A</sub> is a 10 kΩ internal resistor. Diode D<sub>15</sub> opens the feedback loop of A<sub>3</sub> for a negative-going signal if the value of capacitor C<sub>R</sub> is larger than capacitor C<sub>A</sub>. The recovery time depends only on C<sub>R</sub> • R<sub>R</sub>. If the diode impedance is assumed negligible, the dynamic gain G<sub>R</sub>(t) for ΔG is expressed as follows:

$$G_R(t) = (G_{R_{INT}} - G_{R_{FNL}}) e^{-\frac{t}{\tau_R}} + G_{R_{FNL}}$$

$$G_R(t) = (G_{R_{INT}} - G_{R_{FNL}}) e^{-\frac{t}{\tau_R}} + G_{R_{FNL}}$$

$$\tau_R = R_R \cdot C_R = 10 \text{ k}\Omega \cdot C_R$$

where τ<sub>R</sub> is the recovery time constant and R<sub>R</sub> is a 10 kΩ internal resistor. The gain control current is mirrored to the gain cell through Q<sub>14</sub>. The low level gain errors due to input bias current of A<sub>2</sub> and A<sub>3</sub> can be trimmed through the tracking trim pin into A<sub>3</sub> with a current source of ± 3.0 μA.

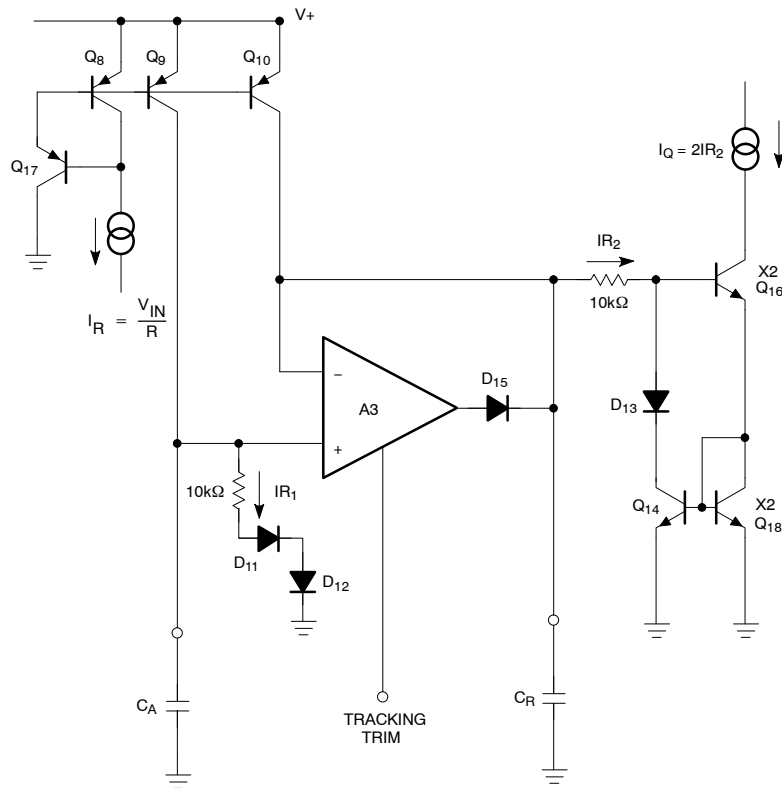


Figure 5. Buffer Amplifier Schematic



## SA572

### Basic Expander

Figure 6 shows an application of the circuit as a simple expander. The gain expression of the system is given by:

$$\frac{V_{OUT}}{V_{IN}} = \left( \frac{2}{I_1} \cdot \frac{R_3 \cdot V_{IN(AVG)}}{R_2 \cdot R_1} \right)^2 \quad (\text{eq. 4})$$

( $I_1 = 140 \mu\text{A}$ )

Both the resistors  $R_1$  and  $R_2$  are tied to internal summing nodes.  $R_1$  is a 6.8 k $\Omega$  internal resistor. The maximum input current into the gain cell can be as large as 140  $\mu\text{A}$ . This corresponds to a voltage level of 140  $\mu\text{A} \cdot 6.8 \text{ k}\Omega = 952 \text{ mV}$  peak. The input peak current into the rectifier is limited to 300  $\mu\text{A}$  by the internal bias system. Note that the value of  $R_1$  can be increased to accommodate higher input level.  $R_2$  and  $R_3$  are external resistors. It is easy to adjust the ratio of  $R_3/R_2$  for desirable system voltage and current levels. A small  $R_2$  results in higher gain control current and smaller static and dynamic tracking error. However, an impedance

buffer  $A_1$  may be necessary if the input is voltage driven with large source impedance.

The gain cell output current feeds the summing node of the external OPA  $A_2$ .  $R_3$  and  $A_2$  convert the gain cell output current to the output voltage. In high-performance applications,  $A_2$  has to be low-noise, high-speed and wide band so that the high-performance output of the gain cell will not be degraded. The non-inverting input of  $A_2$  can be biased at the low noise internal reference Pin 6 or 10. Resistor  $R_4$  is used to bias up the output DC level of  $A_2$  for maximum swing. The output DC level of  $A_2$  is given by:

$$V_{OUT DC} = V_{REF} \left( 1 + \frac{R_3}{R_4} \right) - V_B \frac{R_3}{R_4} \quad (\text{eq. 5})$$

$V_B$  can be tied to a regulated power supply for a dual supply system and be grounded for a single supply system.  $C_A$  sets the attack time constant and  $C_R$  sets the recovery time constant.

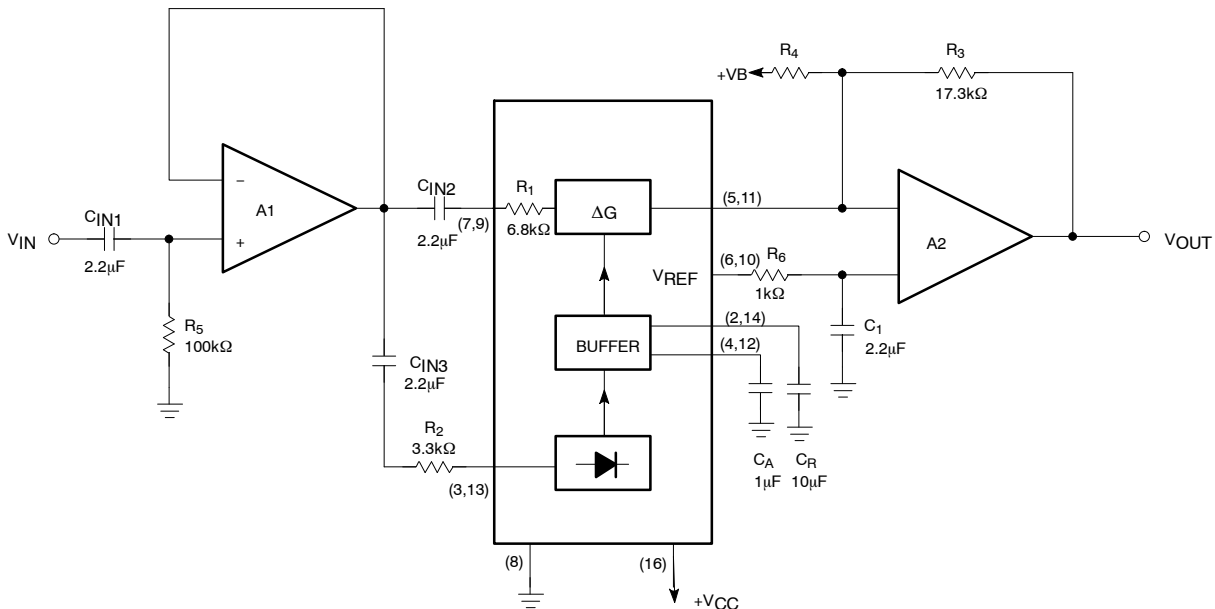


Figure 6. Basic Expander Schematic

### Basic Compressor

Figure 7 shows the hook-up of the circuit as a compressor. The IC is put in the feedback loop of the OPA  $A_1$ . The system gain expression is as follows:

$$\frac{V_{OUT}}{V_{IN}} = \left( \frac{I_1}{2} \cdot \frac{R_2 \cdot R_1}{R_3 \cdot V_{IN(AVG)}} \right)^{\frac{1}{2}} \quad (\text{eq. 6})$$

( $I_1 = 140 \mu\text{A}$ )

$R_{DC1}$ ,  $R_{DC2}$ , and  $CDC$  form a DC feedback for  $A_1$ . The output DC level of  $A_1$  is given by:

$$V_{OUT DC} = V_{REF} \left( 1 + \frac{R_{DC1} + R_{DC2}}{R_4} \right) - V_B \cdot \left( \frac{R_{DC1} + R_{DC2}}{R_4} \right) \quad (\text{eq. 7})$$

The zener diodes  $D_1$  and  $D_2$  are used for channel overload protection.

### SA572

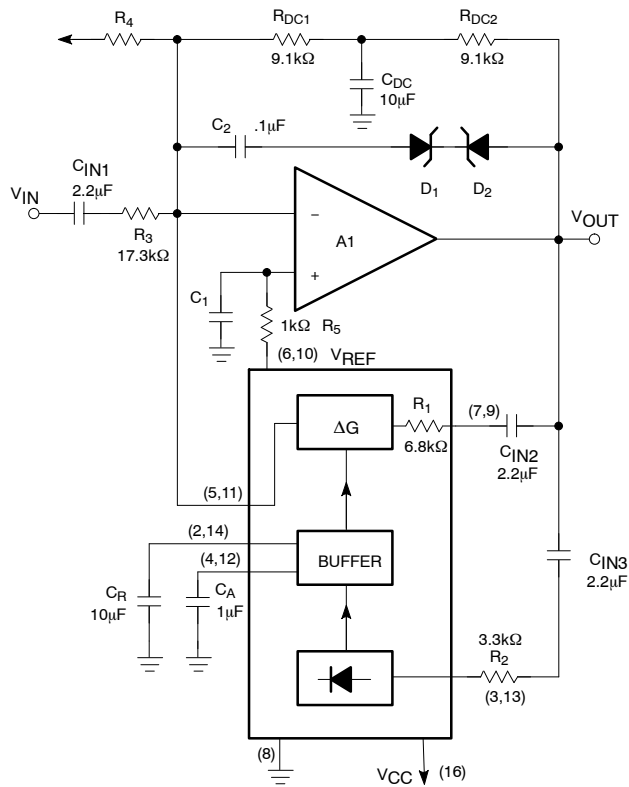


Figure 7. Basic Compressor Schematic

#### Basic Compressor System

The above basic compressor and expander can be applied to systems such as tape/disc noise reduction, digital audio, bucket brigade delay lines. Additional system design techniques such as bandlimiting, band splitting,

pre-emphasis, de-emphasis and equalization are easy to incorporate. The IC is a versatile functional block to achieve a high performance audio system. Figure 8 shows the system level diagram for reference.

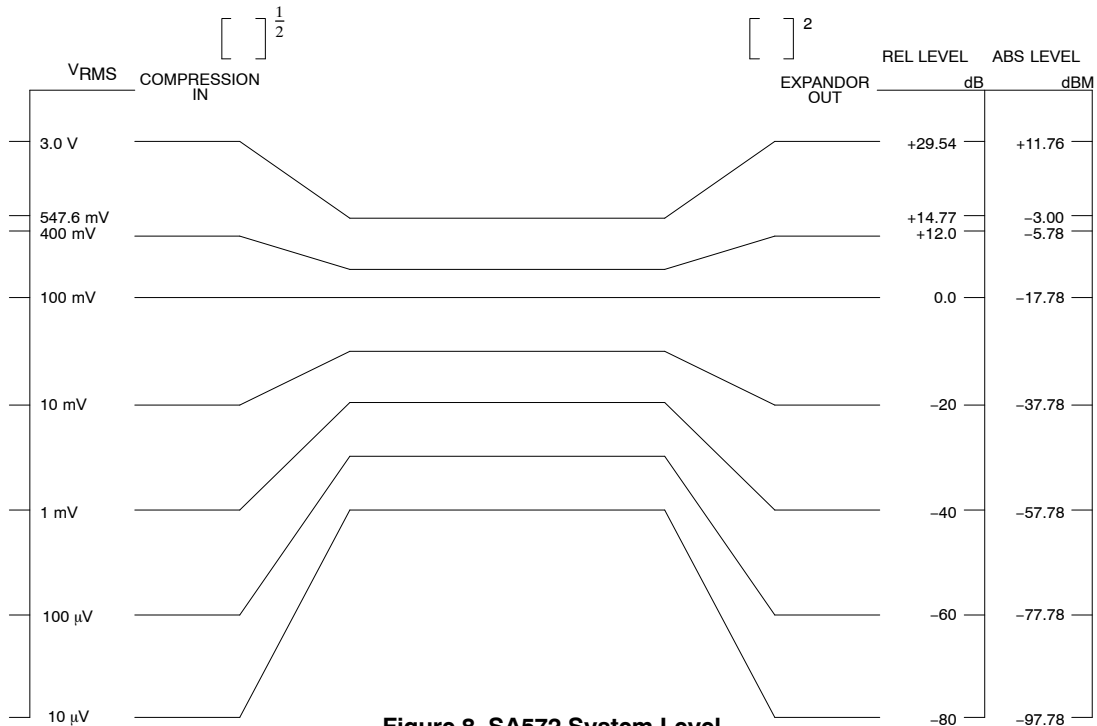


Figure 8. SA572 System Level



**SA572****ORDERING INFORMATION**

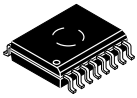
<b>Device</b>	<b>Description</b>	<b>Package</b>	<b>Temperature Range</b>	<b>Shipping<sup>†</sup></b>
SA572DR2G	16-Pin Plastic Small Outline Package (Pb-Free)	SO-16 WB	-40 to +85°C	1000 / Tape & Reel

<sup>†</sup>For information on tape and reel specifications, including part orientation and tape sizes, please refer to our Tape and Reel Packaging Specification Brochure, [BRD8011/D](#).





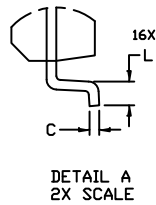
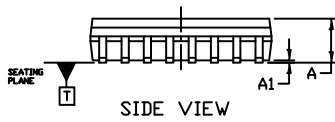
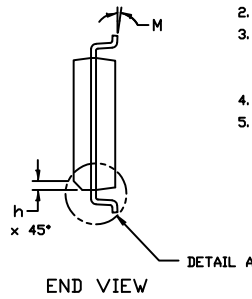
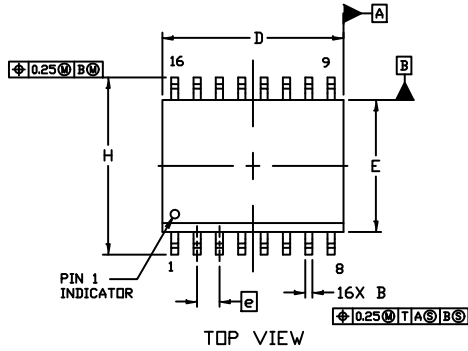
**MECHANICAL CASE OUTLINE  
PACKAGE DIMENSIONS**



1  
SCALE 1:1

**SOIC-16 WB  
CASE 751G  
ISSUE E**

DATE 08 OCT 2021

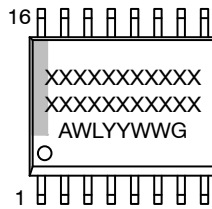


NOTES:

1. DIMENSIONING AND TOLERANCING PER ASME Y14.5M, 1994.
2. CONTROLLING DIMENSION: MILLIMETERS
3. DIMENSION b DOES NOT INCLUDE DAMBAR PROTRUSION. ALLOWABLE PROTRUSION SHALL BE 0.13 TOTAL IN EXCESS OF B DIMENSION AT MAXIMUM MATERIAL CONDITION.
4. DIMENSIONS D AND E DO NOT INCLUDE MOLD PROTRUSIONS.
5. MAXIMUM MOLD PROTRUSION OR FLASH TO BE 0.15 PER SIDE.

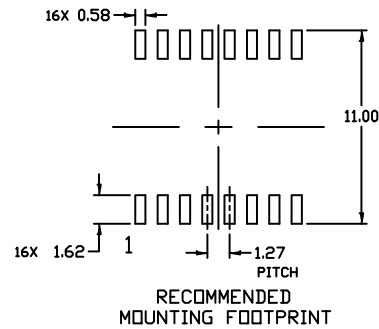
DIM	MILLIMETERS	
	MIN.	MAX.
A	2.35	2.65
A1	0.10	0.25
B	0.35	0.49
C	0.23	0.32
D	10.15	10.45
E	7.40	7.60
e	1.27 BSC	
H	10.05	10.55
h	0.53 REF	
L	0.50	0.90
M	0°	7°

**GENERIC  
MARKING DIAGRAM\***



- XXXXXX = Specific Device Code
- A = Assembly Location
- WL = Wafer Lot
- YY = Year
- WW = Work Week
- G = Pb-Free Package

\*This information is generic. Please refer to device data sheet for actual part marking. Pb-Free indicator, "G" or microdot "•", may or may not be present. Some products may not follow the Generic Marking.



<b>DOCUMENT NUMBER:</b>	<b>98ASB42567B</b>	Electronic versions are uncontrolled except when accessed directly from the Document Repository. Printed versions are uncontrolled except when stamped "CONTROLLED COPY" in red.
<b>DESCRIPTION:</b>	<b>SOIC-16 WB</b>	<b>PAGE 1 OF 1</b>

onsemi and onSEMI are trademarks of Semiconductor Components Industries, LLC dba onsemi or its subsidiaries in the United States and/or other countries. onsemi reserves the right to make changes without further notice to any products herein. onsemi makes no warranty, representation or guarantee regarding the suitability of its products for any particular purpose, nor does onsemi assume any liability arising out of the application or use of any product or circuit, and specifically disclaims any and all liability, including without limitation special, consequential or incidental damages. onsemi does not convey any license under its patent rights nor the rights of others.

**onsemi**, **Onsemi**, and other names, marks, and brands are registered and/or common law trademarks of Semiconductor Components Industries, LLC dba "**onsemi**" or its affiliates and/or subsidiaries in the United States and/or other countries. **onsemi** owns the rights to a number of patents, trademarks, copyrights, trade secrets, and other intellectual property. A listing of **onsemi**'s product/patent coverage may be accessed at [www.onsemi.com/site/pdf/Patent-Marking.pdf](http://www.onsemi.com/site/pdf/Patent-Marking.pdf). **onsemi** reserves the right to make changes at any time to any products or information herein, without notice. The information herein is provided "as-is" and **onsemi** makes no warranty, representation or guarantee regarding the accuracy of the information, product features, availability, functionality, or suitability of its products for any particular purpose, nor does **onsemi** assume any liability arising out of the application or use of any product or circuit, and specifically disclaims any and all liability, including without limitation special, consequential or incidental damages. Buyer is responsible for its products and applications using **onsemi** products, including compliance with all laws, regulations and safety requirements or standards, regardless of any support or applications information provided by **onsemi**. "Typical" parameters which may be provided in **onsemi** data sheets and/or specifications can and do vary in different applications and actual performance may vary over time. All operating parameters, including "Typicals" must be validated for each customer application by customer's technical experts. **onsemi** does not convey any license under any of its intellectual property rights nor the rights of others. **onsemi** products are not designed, intended, or authorized for use as a critical component in life support systems or any FDA Class 3 medical devices or medical devices with a same or similar classification in a foreign jurisdiction or any devices intended for implantation in the human body. Should Buyer purchase or use **onsemi** products for any such unintended or unauthorized application, Buyer shall indemnify and hold **onsemi** and its officers, employees, subsidiaries, affiliates, and distributors harmless against all claims, costs, damages, and expenses, and reasonable attorney fees arising out of, directly or indirectly, any claim of personal injury or death associated with such unintended or unauthorized use, even if such claim alleges that **onsemi** was negligent regarding the design or manufacture of the part. **onsemi** is an Equal Opportunity/Affirmative Action Employer. This literature is subject to all applicable copyright laws and is not for resale in any manner.

#### ADDITIONAL INFORMATION

**TECHNICAL PUBLICATIONS:**

Technical Library: [www.onsemi.com/design/resources/technical-documentation](http://www.onsemi.com/design/resources/technical-documentation)  
onsemi Website: [www.onsemi.com](http://www.onsemi.com)

**ONLINE SUPPORT:** [www.onsemi.com/support](http://www.onsemi.com/support)

For additional information, please contact your local Sales Representative at [www.onsemi.com/support/sales](http://www.onsemi.com/support/sales)

## OUR CERTIFICATE

DiGi provide top-quality products and perfect service for customer worldwide through standardization, technological innovation and continuous improvement. DiGi through third-party certification, we stricly control the quality of products and services. Welcome your RFQ to

Email: [Info@DiGi-Electronics.com](mailto:Info@DiGi-Electronics.com)



Tel: +00 852-30501935

RFQ Email: [Info@DiGi-Electronics.com](mailto:Info@DiGi-Electronics.com)

DiGi is a global authorized distributor of electronic components.