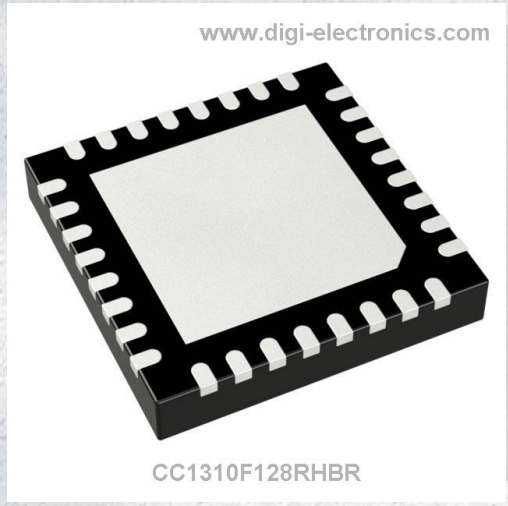
CC1310F128RHBR Datasheet



https://www.DiGi-Electronics.com

DiGi Electronics Part Number	CC1310F128RHBR-DG
Manufacturer	Texas Instruments
Manufacturer Product Number	CC1310F128RHBR
Description	IC RF TXRX+MCU ISM<1GHZ 32VFQFN
Detailed Description	IC RF TxRx + MCU General ISM < 1GHz 300MHz ~ 930 MHz 32-VFQFN Exposed Pad

Tel: +00 852-30501935

RFQ Email: Info@DiGi-Electronics.com

Purchase and inquiry

Manufacturer Product Number:	Manufacturer:
CC1310F128RHBR	Texas Instruments
Series:	Product Status:
SimpleLink™	Active
DiGi-Electronics Programmable:	Type:
Not Verified	TxRx + MCU
RF Family/Standard:	Protocol:
General ISM < 1GHz	-
Modulation:	Frequency:
DSSS, GFSK	300MHz ~ 930MHz
Data Rate (Max):	Power - Output:
50kbps	14dBm
Sensitivity:	Memory Size:
-124dBm	128kB Flash, 20kB RAM
Serial Interfaces:	GPIO:
I2C, I2S, JTAG, SPI, UART	15
Voltage - Supply:	Current - Receiving:
1.8V ~ 3.8V	5.5mA
Current - Transmitting:	Operating Temperature:
12.9mA ~ 22.6mA	-40°C ~ 85°C
Mounting Type:	Package / Case:
Surface Mount	32-VFQFN Exposed Pad
Supplier Device Package:	Base Product Number:
32-VQFN (5x5)	CC1310

Environmental & Export classification

RoHS Status:	Moisture Sensitivity Level (MSL):
ROHS3 Compliant	3 (168 Hours)
REACH Status:	ECCN:
REACH Unaffected	5A992C
HTSUS:	
8542.31.0001	

CC1310 SimpleLink™ Ultra-Low-Power Sub-1 GHz Wireless MCU

1 Device Overview

1.1 Features

- Microcontroller
  - Powerful Arm® Cortex®-M3 Processor
  - EEMBC CoreMark® Score: 142
  - EEMBC ULPBench™ Score: 158
  - Clock Speed up to 48-MHz
  - 32KB, 64KB, and 128KB of In-System Programmable Flash
  - 8KB of SRAM for Cache (or as General-Purpose RAM)
  - 20KB of Ultra-Low-Leakage SRAM
  - 2-Pin cJTAG and JTAG Debugging
  - Supports Over-the-Air (OTA) Update
- Ultra-Low-Power Sensor Controller
  - Can Run Autonomously From the Rest of the System
  - 16-Bit Architecture
  - 2KB of Ultra-Low-Leakage SRAM for Code and Data
- Efficient Code-Size Architecture, Placing Parts of TI-RTOS, Drivers, and Bootloader in ROM
- RoHS-Compliant Package
  - 7-mm × 7-mm RGZ VQFN48 (30 GPIOs)
  - 5-mm × 5-mm RHB VQFN32 (15 GPIOs)
  - 4-mm × 4-mm RSM VQFN32 (10 GPIOs)
- Peripherals
  - All Digital Peripheral Pins Can Be Routed to Any GPIO
  - Four General-Purpose Timer Modules (Eight 16-Bit or Four 32-Bit Timers, PWM Each)
  - 12-Bit ADC, 200 ksamples/s, 8-Channel Analog MUX
  - Continuous Time Comparator
  - Ultra-Low-Power Clocked Comparator
  - Programmable Current Source
  - UART
  - 2× SSI (SPI, MICROWIRE, TI)
  - I^2C, I2S
  - Real-Time Clock (RTC)
  - AES-128 Security Module
  - True Random Number Generator (TRNG)
  - Support for Eight Capacitive Sensing Buttons
  - Integrated Temperature Sensor
- External System
  - On-Chip Internal DC/DC Converter
  - Seamless Integration With the SimpleLink™ CC1190 Range Extender
- Low Power
  - Wide Supply Voltage Range: 1.8 to 3.8 V
  - RX: 5.4 mA
  - TX at +10 dBm: 13.4 mA
  - Active-Mode MCU 48 MHz Running Coremark: 2.5 mA (51 µA/MHz)
  - Active-Mode MCU: 48.5 CoreMark/mA
  - Active-Mode Sensor Controller at 24 MHz: 0.4 mA + 8.2 µA/MHz
  - Sensor Controller, One Wakeup Every Second Performing One 12-Bit ADC Sampling: 0.95 µA
  - Standby: 0.7 µA (RTC Running and RAM and CPU Retention)
  - Shutdown: 185 nA (Wakeup on External Events)
- RF Section
  - Excellent Receiver Sensitivity –124 dBm Using Long-Range Mode, –110 dBm at 50 kbps
  - Excellent Selectivity (±100 kHz): 56 dB
  - Excellent Blocking Performance (±10 MHz): 90 dB
  - Programmable Output Power up to +15 dBm
  - Single-Ended or Differential RF Interface
  - Suitable for Systems Targeting Compliance With Worldwide Radio Frequency Regulations
    - ETSI EN 300 220, EN 303 204 (Europe)
    - FCC CFR47 Part 15 (US)
    - ARIB STD-T108 (Japan)
  - Wireless M-Bus (EN 13757-4) and IEEE® 802.15.4g PHY
- Tools and Development Environment
  - Full-Feature and Low-Cost Development Kits
  - Multiple Reference Designs for Different RF Configurations
  - Packet Sniffer PC Software
  - Sensor Controller Studio
  - SmartRF™ Studio
  - SmartRF Flash Programmer 2
  - IAR Embedded Workbench® for Arm
  - Code Composer Studio™ (CCS) IDE
  - CCS UniFlash

1.2 Applications

- 315-, 433-, 470-, 500-, 779-, 868-, 915-, 920-MHz ISM and SRD Systems
- Low-Power Wireless Systems With 50-kHz to 5-MHz Channel Spacing
- Home and Building Automation
- Wireless Alarm and Security Systems
- Industrial Monitoring and Control
- Smart Grid and Automatic Meter Reading
- Wireless Healthcare Applications
- Wireless Sensor Networks
- Active RFID
- IEEE 802.15.4g, IP-Enabled Smart Objects (6LoWPAN), Wireless M-Bus, KNX Systems, Wi-SUN™, and Proprietary Systems
- Energy-Harvesting Applications
- Electronic Shelf Label (ESL)
- Long-Range Sensor Applications
- Heat-Cost Allocators

1.3 Description

The CC1310 device is a cost-effective, ultra-low-power, Sub-1 GHz RF device from Texas Instruments™ that is part of the SimpleLink™ microcontroller (MCU) platform. The platform consists of Wi-Fi®, *Bluetooth*® low energy, Sub-1 GHz, Ethernet, Zigbee®, Thread, and host MCUs. These devices all share a common, easy-to-use development environment with a single core software development kit (SDK) and a rich tool set. A one-time integration of the SimpleLink platform enables users to add any combination of devices from the portfolio into their design, allowing 100 percent code reuse when design requirements change. For more information, visit www.ti.com/simplelink.

With very low active RF and MCU current consumption, in addition to flexible low-power modes, the CC1310 device provides excellent battery life and allows long-range operation on small coin-cell batteries and in energy harvesting applications.

The CC1310 is a device in the CC13xx and CC26xx family of cost-effective, ultra-low-power wireless MCUs capable of handling Sub-1 GHz RF frequencies. The CC1310 device combines a flexible, very low-power RF transceiver with a powerful 48-MHz Arm® Cortex®-M3 microcontroller in a platform supporting multiple physical layers and RF standards. A dedicated Radio Controller (Cortex®-M0) handles low-level RF protocol commands that are stored in ROM or RAM, thus ensuring ultra-low power and flexibility. The low-power consumption of the CC1310 device does not come at the expense of RF performance; the CC1310 device has excellent sensitivity and robustness (selectivity and blocking) performance.

The CC1310 device is a highly integrated, true single-chip solution incorporating a complete RF system and an on-chip DC/DC converter.

Sensors can be handled in a very low-power manner by a dedicated autonomous ultra-low-power MCU that can be configured to handle analog and digital sensors; thus the main MCU (Arm® Cortex®-M3) can maximize sleep time.

The power and clock management and radio systems of the CC1310 device require specific configuration and handling by software to operate correctly, which has been implemented in the TI-RTOS. TI recommends using this software framework for all application development on the device. The complete TI-RTOS and device drivers are offered free of charge in source code.

Device Information(1)

PART NUMBER	PACKAGE	BODY SIZE (NOM)
CC1310F128RGZ	VQFN (48)	7.00 mm × 7.00 mm
CC1310F128RHB	VQFN (32)	5.00 mm × 5.00 mm
CC1310F128RSM	VQFN (32)	4.00 mm × 4.00 mm
CC1310F64RGZ	VQFN (48)	7.00 mm × 7.00 mm
CC1310F64RHB	VQFN (32)	5.00 mm × 5.00 mm
CC1310F64RSM	VQFN (32)	4.00 mm × 4.00 mm
CC1310F32RGZ	VQFN (48)	7.00 mm × 7.00 mm
CC1310F32RHB	VQFN (32)	5.00 mm × 5.00 mm
CC1310F32RSM	VQFN (32)	4.00 mm × 4.00 mm

(1) For more information, see Section 9.

1.4 Functional Block Diagram

Figure 1-1 shows a block diagram for the CC1310 device.

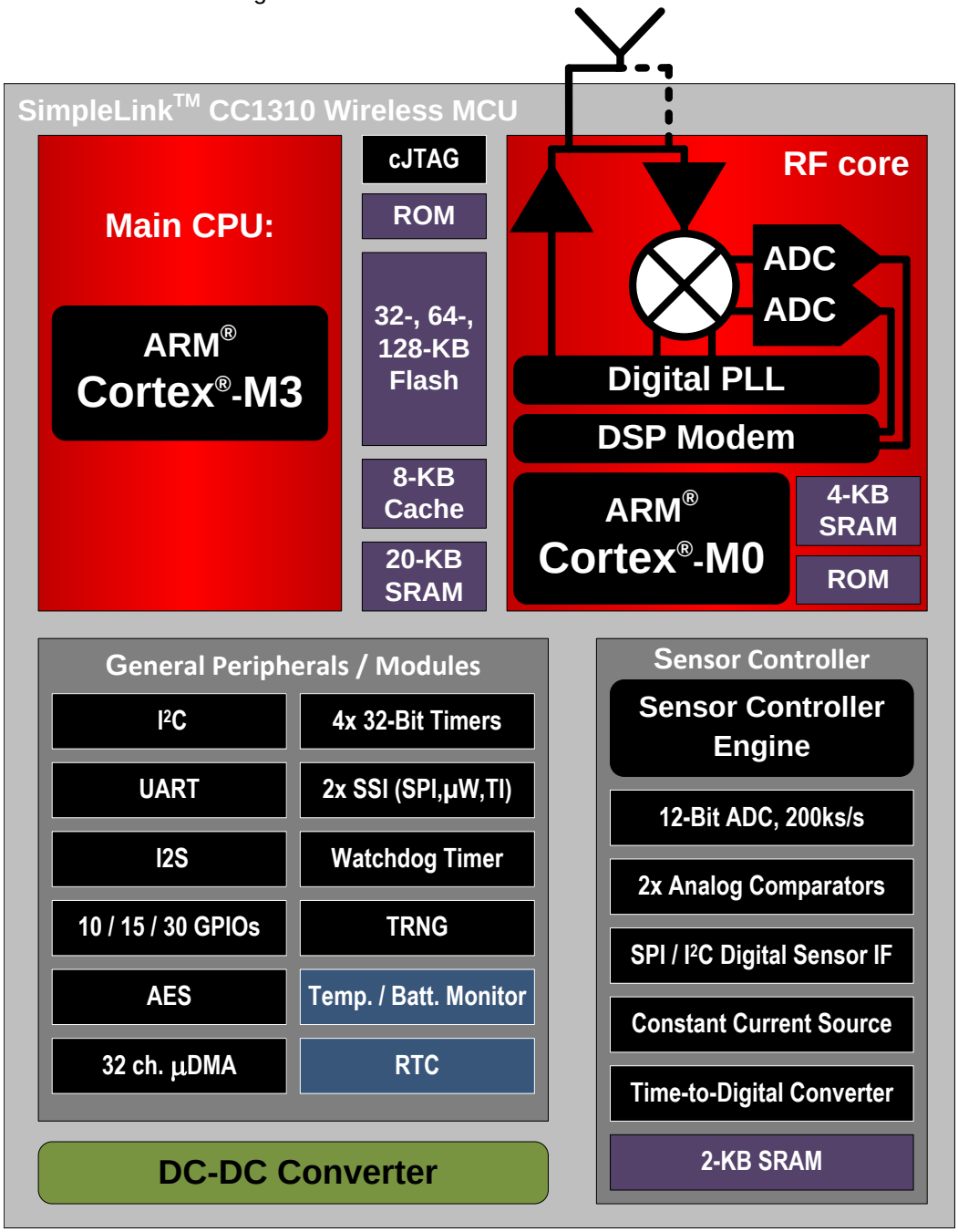


Figure 1-1. CC1310 Block Diagram

Table of Contents

2 Revision History

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

Changes from October 27, 2016 to July 13, 2018 **Page**

- Added Code Composer Studio UniFlash 1
- Changed *Description* section 2
- Changed Table 3-1 7
- Changed Figure 4-1 8
- Changed Figure 4-2 10
- Added support for split supply rail to Section 5.3 15
- Changed Operating supply voltage 15
- Added test conditions at 433.92 MHz to Section 5.4 16
- Moved footnote to specific values in Section 5.5 16
- Changed footnote in Section 5.5 16
- Changed test conditions for Receiver sensitivity, 50 kbps in Section 5.6 17
- Added parameters to Section 5.6 17
- Added Receiver sensitivity parameters to Section 5.7 23
- Changed 31
- Changed footnote in 31
- Added Software section 48

Changes from October 28, 2015 to October 27, 2016 **Page**

- Added 32KB and 64KB to the *Features* bullet for in-system programmable flash 1
- Changed to the correct pin count in the *Features* bullet *RoHS-Compliant Package* 1
- Changed CC1310 Block Diagram 4
- Changed Figure 4-2, corrected typo in pin name 10
- Changed the table note in Section 5.1 from: VDDS to: ground 15
- Changed ESD ratings for all pins in Section 5.2 15
- Added OOK modulation power consumption to Section 5.4 16
- Added OOK modulation sensitivity to Section 5.6 22
- Added receive parameters for 431-MHz to 527-MHz band in Section 5.7 23
- Added transmit parameters for 431-MHz to 527-MHz band in Section 5.9 26
- Changed ADC reference voltage to correct value in Section 5.11 27
- Added thermal characteristics for RHB and RSM packages in Section 5.18 30
- Changed *Standby MCU Current Consumption, 32-kHz Clock, RAM and MCU Retention* by extending the temperature 34
- Changed BOD restriction footnote in Table 6-2—restriction does not apply to die revision B and later 42
- Added Section 6.10 44
- Changed Figure 8-1 47

Changes from September 30, 2015 to October 28, 2015 **Page**

- Added the RSM and RHB packages 8

Changes from August 31, 2015 to September 30, 2015 **Page**

- Changed device status from: Product Preview to: Production Data 1
- Removed the RSM and RHB packages 8

3 Device Comparison

Table 3-1 lists the device family overview.

Table 3-1. Device Family Overview

DEVICE	RADIO SUPPORT	FLASH (KB)	RAM (KB)	GPIOs	PACKAGE SIZE
CC1310F128RGZ	Proprietary, Wireless M-Bus, IEEE 802.15.4g	128	20	30	RGZ (7 mm × 7 mm VQFN48)
CC1310F64RGZ		64	16	30	
CC1310F32RGZ		32	16	30	
CC1310F128RHB	Proprietary, Wireless M-Bus, IEEE 802.15.4g	128	20	15	RHB (5 mm × 5 mm VQFN32)
CC1310F64RHB		64	16	15	
CC1310F32RHB		32	16	15	
CC1310F128RSM	Proprietary, Wireless M-Bus, IEEE 802.15.4g	128	20	10	RSM (4 mm × 4 mm VQFN32)
CC1310F64RSM		64	16	10	
CC1310F32RSM		32	16	10	
CC1350	Sub-1 GHz Bluetooth low energy	128	20	10-30	RGZ (7 mm × 7 mm VQFN48) RHB (5 mm × 5 mm VQFN32) RSM (4 mm × 4 mm VQFN32)
CC2640R2	Bluetooth 5 low energy 2.4-GHz proprietary FSK-based formats	128	20	10-31	RGZ (7 mm × 7 mm VQFN48) RHB (5 mm × 5 mm VQFN32) RSM (4 mm × 4 mm VQFN32) YFV (2.7 mm × 2.7 mm DSBGA34)
CC1312R	Sub-1 GHz Proprietary, Wireless M-Bus, IEEE 802.15.4g	352	80	30	RGZ (7 mm × 7 mm VQFN48)
CC1352R	Dual-band (2.4-GHz and Sub-1 GHz) Multiprotocol	352	80	28	RGZ (7 mm × 7 mm VQFN48)
CC2652R	Multiprotocol Bluetooth 5 low energy Zigbee Thread 2.4-GHz proprietary FSK-based formats	352	80	31	RGZ (7 mm × 7 mm VQFN48)

3.1 Related Products

Wireless Connectivity The wireless connectivity portfolio offers a wide selection of low-power RF solutions suitable for a broad range of application. The offerings range from fully customized solutions to turnkey offerings with precertified hardware and software (protocol).

Sub-1 GHz Long-range, low power wireless connectivity solutions are offered in a wide range of Sub-1 GHz ISM bands.

Companion Products Review products that are frequently purchased or used with this product.

4 Terminal Configuration and Functions

4.1 Pin Diagram – RSM Package

Figure 4-1 shows the RSM pinout diagram.

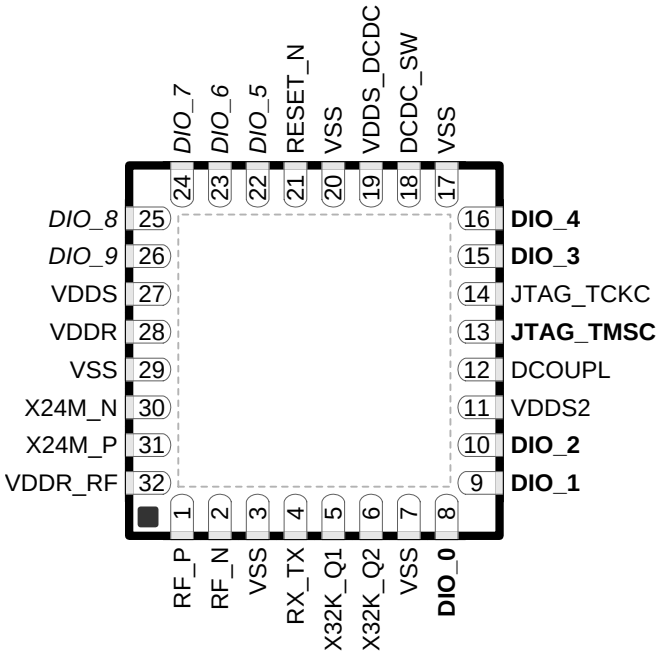


Figure 4-1. RSM (4-mm × 4-mm) Pinout, 0.4-mm Pitch
Top View

I/O pins marked in Figure 4-1 in **bold** have high-drive capabilities; they are as follows:

- Pin 8, DIO_0
- Pin 9, DIO_1
- Pin 10, DIO_2
- Pin 13, JTAG_TMSC
- Pin 15, DIO_3
- Pin 16, DIO_4

I/O pins marked in Figure 4-1 in *italics* have analog capabilities; they are as follows:

- Pin 22, DIO_5
- Pin 23, DIO_6
- Pin 24, DIO_7
- Pin 25, DIO_8
- Pin 26, DIO_9

4.2 Signal Descriptions – RSM Package

Table 4-1. Signal Descriptions – RSM Package

PIN		TYPE	DESCRIPTION
NAME	NO.		
DCDC_SW	18	Power	Output from internal DC/DC (1)
DCOUPL	12	Power	1.27-V regulated digital-supply decoupling capacitor (2)
DIO_0	8	Digital I/O	GPIO, Sensor Controller, high-drive capability
DIO_1	9	Digital I/O	GPIO, Sensor Controller, high-drive capability
DIO_2	10	Digital I/O	GPIO, Sensor Controller, high-drive capability
DIO_3	15	Digital I/O	GPIO, high-drive capability, JTAG_TDO
DIO_4	16	Digital I/O	GPIO, high-drive capability, JTAG_TDI
DIO_5	22	Digital or analog I/O	GPIO, Sensor Controller, analog
DIO_6	23	Digital or analog I/O	GPIO, Sensor Controller, analog
DIO_7	24	Digital or analog I/O	GPIO, Sensor Controller, analog
DIO_8	25	Digital or analog I/O	GPIO, Sensor Controller, analog
DIO_9	26	Digital or analog I/O	GPIO, Sensor Controller, analog
EGP	–	Power	Ground; exposed ground pad
JTAG_TMSC	13	Digital I/O	JTAG TMSC
JTAG_TCKC	14	Digital I/O	JTAG TCKC (3)
RESET_N	21	Digital input	Reset, active low. No internal pullup.
RF_N	2	RF I/O	Negative RF input signal to LNA during RX Negative RF output signal from PA during TX
RF_P	1	RF I/O	Positive RF input signal to LNA during RX Positive RF output signal from PA during TX
RX_TX	4	RF I/O	Optional bias pin for the RF LNA
VDDS	27	Power	1.8-V to 3.8-V main chip supply (1)
VDDS2	11	Power	1.8-V to 3.8-V GPIO supply (1)
VDDS_DCDC	19	Power	1.8-V to 3.8-V DC/DC supply
VDDR	28	Power	1.7-V to 1.95-V supply, connect to output of internal DC/DC (2) (4)
VDDR_RF	32	Power	1.7-V to 1.95-V supply, connect to output of internal DC/DC (2) (5)
VSS	3, 7, 17, 20, 29	Power	Ground
X32K_Q1	5	Analog I/O	32-kHz crystal oscillator pin 1
X32K_Q2	6	Analog I/O	32-kHz crystal oscillator pin 2
X24M_N	30	Analog I/O	24-MHz crystal oscillator pin 1
X24M_P	31	Analog I/O	24-MHz crystal oscillator pin 2

(1) See the technical reference manual listed in Section 8.3 for more details.
(2) Do not supply external circuitry from this pin.
(3) For design consideration regrading noise immunity for this pin, see the *JTAG Interface* chapter in the *CC13x0, CC26x0 SimpleLink™ Wireless MCU Technical Reference Manual*.
(4) If internal DC/DC is not used, this pin is supplied internally from the main LDO.
(5) If internal DC/DC is not used, this pin must be connected to VDDR for supply from the main LDO.

4.3 Pin Diagram – RHB Package

Figure 4-2 shows the RHB pinout diagram.

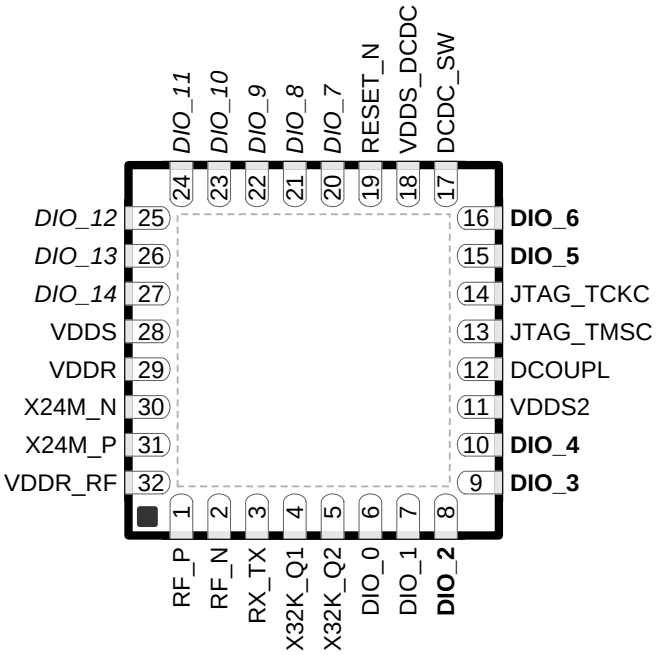


Figure 4-2. RHB (5-mm × 5-mm) Pinout, 0.5-mm Pitch
Top View

I/O pins marked in Figure 4-2 in **bold** have high-drive capabilities; they are as follows:

- Pin 8, DIO_2
- Pin 9, DIO_3
- Pin 10, DIO_4
- Pin 15, DIO_5
- Pin 16, DIO_6

I/O pins marked in Figure 4-2 in *italics* have analog capabilities; they are as follows:

- Pin 20, DIO_7
- Pin 21, DIO_8
- Pin 22, DIO_9
- Pin 23, DIO_10
- Pin 24, DIO_11
- Pin 25, DIO_12
- Pin 26, DIO_13
- Pin 27, DIO_14

4.4 Signal Descriptions – RHB Package

Table 4-2. Signal Descriptions – RHB Package

PIN		TYPE	DESCRIPTION
NAME	NO.		
DCDC_SW	17	Power	Output from internal DC/DC (1)
DCOUPL	12	Power	1.27-V regulated digital-supply decoupling (2)
DIO_0	6	Digital I/O	GPIO, Sensor Controller
DIO_1	7	Digital I/O	GPIO, Sensor Controller
DIO_2	8	Digital I/O	GPIO, Sensor Controller, high-drive capability
DIO_3	9	Digital I/O	GPIO, Sensor Controller, high-drive capability
DIO_4	10	Digital I/O	GPIO, Sensor Controller, high-drive capability
DIO_5	15	Digital I/O	GPIO, high-drive capability, JTAG_TDO
DIO_6	16	Digital I/O	GPIO, high-drive capability, JTAG_TDI
DIO_7	20	Digital or analog I/O	GPIO, Sensor Controller, analog
DIO_8	21	Digital or analog I/O	GPIO, Sensor Controller, analog
DIO_9	22	Digital or analog I/O	GPIO, Sensor Controller, analog
DIO_10	23	Digital or analog I/O	GPIO, Sensor Controller, Analog
DIO_11	24	Digital or analog I/O	GPIO, Sensor Controller, analog
DIO_12	25	Digital or analog I/O	GPIO, Sensor Controller, analog
DIO_13	26	Digital or analog I/O	GPIO, Sensor Controller, analog
DIO_14	27	Digital or analog I/O	GPIO, Sensor Controller, analog
EGP	–	Power	Ground; exposed ground pad
JTAG_TMSC	13	Digital I/O	JTAG TMSC, high-drive capability
JTAG_TCKC	14	Digital I/O	JTAG TCKC (3)
RESET_N	19	Digital input	Reset, active low. No internal pullup.
RF_N	2	RF I/O	Negative RF input signal to LNA during RX Negative RF output signal from PA during TX
RF_P	1	RF I/O	Positive RF input signal to LNA during RX Positive RF output signal from PA during TX
RX_TX	3	RF I/O	Optional bias pin for the RF LNA
VDDR	29	Power	1.7-V to 1.95-V supply, connect to output of internal DC/DC (2) (4)
VDDR_RF	32	Power	1.7-V to 1.95-V supply, connect to output of internal DC/DC (2) (5)
VDDS	28	Power	1.8-V to 3.8-V main chip supply (1)
VDDS2	11	Power	1.8-V to 3.8-V GPIO supply (1)
VDDS_DCDC	18	Power	1.8-V to 3.8-V DC/DC supply
X24M_N	30	Analog I/O	24-MHz crystal oscillator pin 1
X24M_P	31	Analog I/O	24-MHz crystal oscillator pin 2
X32K_Q1	4	Analog I/O	32-kHz crystal oscillator pin 1
X32K_Q2	5	Analog I/O	32-kHz crystal oscillator pin 2

(1) For more details, see the technical reference manual listed in Section 8.3.
(2) Do not supply external circuitry from this pin.
(3) For design consideration regrading noise immunity for this pin, see the *JTAG Interface* chapter in the *CC13x0, CC26x0 SimpleLink™ Wireless MCU Technical Reference Manual*.
(4) If internal DC/DC is not used, this pin is supplied internally from the main LDO.
(5) If internal DC/DC is not used, this pin must be connected to VDDR for supply from the main LDO.

4.5 Pin Diagram – RGZ Package

Figure 4-3 shows the RGZ pinout diagram.

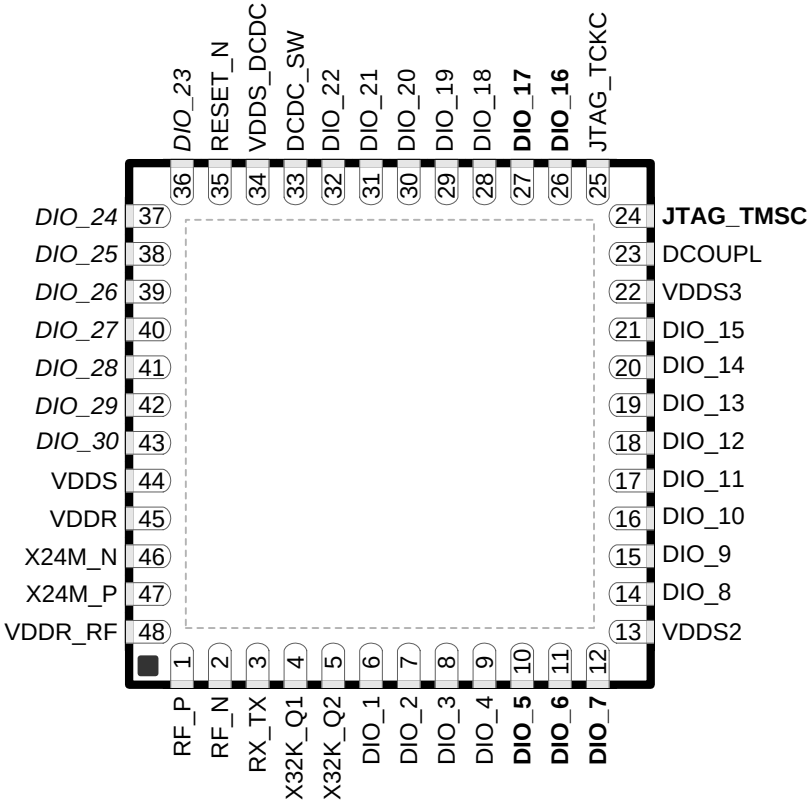


Figure 4-3. RGZ (7-mm × 7-mm) Pinout, 0.5-mm Pitch
Top View

I/O pins marked in Figure 4-3 in **bold** have high-drive capabilities; they are as follows:

- Pin 10, DIO_5
- Pin 11, DIO_6
- Pin 12, DIO_7
- Pin 24, JTAG_TMSC
- Pin 26, DIO_16
- Pin 27, DIO_17

I/O pins marked in Figure 4-3 in *italics* have analog capabilities; they are as follows:

- Pin 36, DIO_23
- Pin 37, DIO_24
- Pin 38, DIO_25
- Pin 39, DIO_26
- Pin 40, DIO_27
- Pin 41, DIO_28
- Pin 42, DIO_29
- Pin 43, DIO_30

4.6 Signal Descriptions – RGZ Package

Table 4-3. Signal Descriptions – RGZ Package

PIN		TYPE	DESCRIPTION
NAME	NO.		
DCDC_SW	33	Power	Output from internal DC/DC(1)(2)
DCOUPL	23	Power	1.27-V regulated digital-supply (decoupling capacitor)(2)
DIO_1	6	Digital I/O	GPIO, Sensor Controller
DIO_2	7	Digital I/O	GPIO, Sensor Controller
DIO_3	8	Digital I/O	GPIO, Sensor Controller
DIO_4	9	Digital I/O	GPIO, Sensor Controller
DIO_5	10	Digital I/O	GPIO, Sensor Controller, high-drive capability
DIO_6	11	Digital I/O	GPIO, Sensor Controller, high-drive capability
DIO_7	12	Digital I/O	GPIO, Sensor Controller, high-drive capability
DIO_8	14	Digital I/O	GPIO
DIO_9	15	Digital I/O	GPIO
DIO_10	16	Digital I/O	GPIO
DIO_11	17	Digital I/O	GPIO
DIO_12	18	Digital I/O	GPIO
DIO_13	19	Digital I/O	GPIO
DIO_14	20	Digital I/O	GPIO
DIO_15	21	Digital I/O	GPIO
DIO_16	26	Digital I/O	GPIO, JTAG_TDO, high-drive capability
DIO_17	27	Digital I/O	GPIO, JTAG_TDI, high-drive capability
DIO_18	28	Digital I/O	GPIO
DIO_19	29	Digital I/O	GPIO
DIO_20	30	Digital I/O	GPIO
DIO_21	31	Digital I/O	GPIO
DIO_22	32	Digital I/O	GPIO
DIO_23	36	Digital or analog I/O	GPIO, Sensor Controller, analog
DIO_24	37	Digital or analog I/O	GPIO, Sensor Controller, analog
DIO_25	38	Digital or analog I/O	GPIO, Sensor Controller, analog
DIO_26	39	Digital or analog I/O	GPIO, Sensor Controller, analog
DIO_27	40	Digital or analog I/O	GPIO, Sensor Controller, analog
DIO_28	41	Digital or analog I/O	GPIO, Sensor Controller, analog
DIO_29	42	Digital or analog I/O	GPIO, Sensor Controller, analog
DIO_30	43	Digital or analog I/O	GPIO, Sensor Controller, analog
EGP	–	Power	Ground; exposed ground pad
JTAG_TMSC	24	Digital I/O	JTAG TMSC, high-drive capability
JTAG_TCKC	25	Digital I/O	JTAG TCKC(3)
RESET_N	35	Digital input	Reset, active-low. No internal pullup.
RF_N	2	RF I/O	Negative RF input signal to LNA during RX Negative RF output signal from PA during TX
RF_P	1	RF I/O	Positive RF input signal to LNA during RX Positive RF output signal from PA during TX

(1) See technical reference manual listed in Section 8.3 for more details.
(2) Do not supply external circuitry from this pin.
(3) For design consideration regrading noise immunity for this pin, see the *JTAG Interface* chapter in the *CC13x0, CC26x0 SimpleLink™ Wireless MCU Technical Reference Manual*.

Table 4-3. Signal Descriptions – RGZ Package (continued)

PIN		TYPE	DESCRIPTION
NAME	NO.		
VDDR	45	Power	1.7-V to 1.95-V supply, connect to output of internal DC/DC (2) (4)
VDDR_RF	48	Power	1.7-V to 1.95-V supply, connect to output of internal DC/DC (2) (5)
VDDS	44	Power	1.8-V to 3.8-V main chip supply (1)
VDDS2	13	Power	1.8-V to 3.8-V DIO supply (1)
VDDS3	22	Power	1.8-V to 3.8-V DIO supply (1)
VDDS_DCDC	34	Power	1.8-V to 3.8-V DC/DC supply
X24M_N	46	Analog I/O	24-MHz crystal oscillator pin 1
X24M_P	47	Analog I/O	24-MHz crystal oscillator pin 2
RX_TX	3	RF I/O	Optional bias pin for the RF LNA
X32K_Q1	4	Analog I/O	32-kHz crystal oscillator pin 1
X32K_Q2	5	Analog I/O	32-kHz crystal oscillator pin 2

(4) If internal DC/DC is not used, this pin is supplied internally from the main LDO.
(5) If internal DC/DC is not used, this pin must be connected to VDDR for supply from the main LDO.

5 Specifications

5.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)(1)(2)

		MIN	MAX	UNIT
Supply voltage (VDDS, VDDS2, and VDDS3)		–0.3	4.1	V
Voltage on any digital pin(3)(4)		–0.3	VDDSn + 0.3, max 4.1	V
Voltage on crystal oscillator pins X32K_Q1, X32K_Q2, X24M_N, and X24M_P		–0.3	VDDR + 0.3, max 2.25	V
Voltage on ADC input (V_{in})	Voltage scaling enabled	–0.3	VDDS	V
	Voltage scaling disabled, internal reference	–0.3	1.49	
	Voltage scaling disabled, VDDS as reference	–0.3	VDDS / 2.9	
Input RF level			10	dBm
Storage temperature (T_{stg})		–40	150	°C

(1) Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Conditions* is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
(2) All voltage values are with respect to ground, unless otherwise noted.
(3) Including analog-capable DIO.
(4) Each pin is referenced to a specific VDDSn (VDDS, VDDS2 or VDDS3). For a pin-to-VDDS mapping table, see Table 6-3.

5.2 ESD Ratings

				VALUE	UNIT
V_{ESD}	Electrostatic discharge	Human body model (HBM), per ANSI/ESDA/JEDEC JS001(1)	All pins	±3000	V
		Charged device model (CDM), per JESD22-C101(2)	All pins	±500	

(1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.
(2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

5.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

			MIN	MAX	UNIT
Ambient temperature			–40	85	°C
Operating supply voltage (VDDS)	For operation in battery-powered and 3.3-V systems (internal DC/DC can be used to minimize power consumption)		1.8	3.8	V
Operating supply voltages (VDDS2 and VDDS3)		VDDS < 2.7 V	1.8	3.8	V
Operating supply voltages (VDDS2 and VDDS3)		VDDS ≥ 2.7 V	1.9	3.8	V
Rising supply voltage slew rate			0	100	mV/µs
Falling supply voltage slew rate			0	20	mV/µs
Falling supply voltage slew rate, with low-power flash setting(1)				3	mV/µs
Positive temperature gradient in standby(2)	No limitation for negative temperature gradient, or outside standby mode			5	°C/s

(1) For small coin-cell batteries, with high worst-case end-of-life equivalent source resistance, a 22-µF VDDS input capacitor must be used to ensure compliance with this slew rate.
(2) Applications using RCOSC_LF as sleep timer must also consider the drift in frequency caused by a change in temperature (see).

5.4 Power Consumption Summary

Measured on the Texas Instruments CC1310EM-7XD-7793 reference design unless otherwise noted. T_c = 25°C, V_{DDS} = 3.6 V with DC/DC enabled, unless otherwise noted. Using boost mode (increasing VDDR to 1.95 V), will increase currents in this table by 15% (does not apply to TX 14-dBm setting where this current is already included).

PARAMETER		TEST CONDITIONS	TYP	UNIT
I_{core}	Core current consumption	Reset. RESET_N pin asserted or VDDS below power-on-reset threshold	100	nA
		Shutdown. No clocks running, no retention	185	
		Standby. With RTC, CPU, RAM, and (partial) register retention. RCOSC_LF	0.7	µA
		Standby. With RTC, CPU, RAM, and (partial) register retention. XOSC_LF	0.8	
		Idle. Supply Systems and RAM powered.	570	
		Active. MCU running CoreMark at 48 MHz	1.2 mA + 25.5 µA/MHz	
		Active. MCU running CoreMark at 48 MHz	2.5	mA
		Active. MCU running CoreMark at 24 MHz	1.9	
		Radio RX, 868 MHz	5.5	mA
		Radio TX, 10-dBm output power, (G)FSK, 868 MHz	13.4	mA
		Radio TX, OOK modulation, 10-dBm output power, AVG	11.2	mA
		Radio TX, boost mode (VDDR = 1.95 V), 14-dBm output power, (G)FSK, 868 MHz	23.5	mA
		Radio TX, OOK modulation, boost mode (VDDR = 1.95 V), 14-dBm, AVG	14.8	mA
		Radio TX, boost mode (VDDR = 1.95 V), 15-dBm output power, (G)FSK, measured on CC1310EM-7XD-4251, 433.92 MHz	25.1	mA
		Radio TX, 10-dBm output power, measured on CC1310EM-7XD-4251, 433.92 MHz	13.2	mA
PERIPHERAL CURRENT CONSUMPTION (1)(2)(3)				
I_{peri}	Peripheral power domain	Delta current with domain enabled	20	µA
	Serial power domain	Delta current with domain enabled	13	
	RF core	Delta current with power domain enabled, clock enabled, RF core idle	237	
	µDMA	Delta current with clock enabled, module idle	130	
	Timers	Delta current with clock enabled, module idle	113	
	I^2C	Delta current with clock enabled, module idle	12	
	I2S	Delta current with clock enabled, module idle	36	
	SSI	Delta current with clock enabled, module idle	93	
	UART	Delta current with clock enabled, module idle	164	

(1) Adds to core current I_{core} for each peripheral unit activated
(2) I_{peri} is not supported in standby or shutdown modes.
(3) Measured at 3.0 V

5.5 RF Characteristics

over operating free-air temperature range (unless otherwise noted)

PARAMETER	MIN	TYP	MAX	UNIT
Frequency bands	287 (1)		351 (1)	MHz
	359 (1)		439 (1)	
	431		527	
	718 (1)		878 (1)	
	861		1054	

(1) These frequency bands are functionally verified. Radio settings for specific physical layer parameters can be made available upon request.

5.6 Receive (RX) Parameters, 861 MHz to 1054 MHz

Measured on the Texas Instruments CC1310EM-7XD-7793 reference design with T_c = 25°C, V_{DDS} = 3.0 V, DC/DC enabled, f_{RF} = 868 MHz, unless otherwise noted. All measurements are performed at the antenna input with a combined RX and TX path.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Data rate			Up to 4 Mbps		bps
Data rate offset tolerance, IEEE 802.15.4g PHY	50 kbps, GFSK, 25-kHz deviation, 100-kHz RX bandwidth (same modulation format as IEEE 802.15.4g mandatory mode), BER = 10^{-3}		1600		ppm
Data rate step size			1.5		bps
Digital channel filter programmable bandwidth	Using VCO divide by 5 setting	40		4000	kHz
Receiver sensitivity, 50 kbps	50 kbps, GFSK, 25-kHz deviation, 100-kHz RX bandwidth (same modulation format as IEEE 802.15.4g mandatory mode), BER = 10^{-2}. 868 MHz and 915 MHz		–110		dBm
Receiver saturation	50 kbps, GFSK, 25-kHz deviation, 100-kHz RX bandwidth (same modulation format as IEEE 802.15.4g mandatory mode), BER = 10^{-2}		10		dBm
Selectivity, ±200 kHz, 50 kbps	Wanted signal 3 dB above sensitivity limit. 50 kbps, GFSK, 25-kHz deviation, 100-kHz RX bandwidth (same modulation format as IEEE 802.15.4g mandatory mode), BER = 10^{-2}		44, 47		dB
Selectivity, ±400 kHz, 50 kbps	Wanted signal 3 dB above sensitivity limit. 50 kbps, GFSK, 25-kHz deviation, 100-kHz RX bandwidth (same modulation format as IEEE 802.15.4g mandatory mode), BER = 10^{-2}		48, 53		dB
Blocking ±1 MHz, 50 kbps	Wanted signal 3 dB above sensitivity limit. 50 kbps, GFSK, 25-kHz deviation, 100-kHz RX bandwidth (same modulation format as IEEE 802.15.4g mandatory mode), BER = 10^{-2}		59, 62		dB
Blocking ±2 MHz, 50 kbps	Wanted signal 3 dB above sensitivity limit. 50 kbps, GFSK, 25-kHz deviation, 100-kHz RX bandwidth (same modulation format as IEEE 802.15.4g mandatory mode), BER = 10^{-2}		64, 65		dB
Blocking ±5 MHz, 50 kbps	Wanted signal 3 dB above sensitivity limit. 50 kbps, GFSK, 25-kHz deviation, 100-kHz RX bandwidth (same modulation format as IEEE 802.15.4g mandatory mode), BER = 10^{-2}		67, 68		dB
Blocking ±10 MHz, 50 kbps	Wanted signal 3 dB above sensitivity limit. 50 kbps, GFSK, 25-kHz deviation, 100-kHz RX bandwidth (same modulation format as IEEE 802.15.4g mandatory mode), BER = 10^{-2}		76, 76		dB
Spurious emissions 1 GHz to 13 GHz (VCO leakage at 3.5 GHz) and 30 MHz to 1 GHz	Conducted emissions measured according to ETSI EN 300 220		–70		dBm
Image rejection (image compensation enabled, the image compensation is calibrated in production), 50 kbps	Wanted signal 3 dB above sensitivity limit. 50 kbps, GFSK, 25-kHz deviation, 100-kHz RX bandwidth (same modulation format as IEEE 802.15.4g mandatory mode), BER = 10^{-2}		44		dB
RSSI dynamic range	50 kbps, GFSK, 25-kHz deviation, 100-kHz RX bandwidth (same modulation format as IEEE 802.15.4g mandatory mode). Starting from the sensitivity limit. This range will give an accuracy of ±2 dB.		95		dB
RSSI accuracy	50 kbps, GFSK, 25-kHz deviation, 100-kHz RX bandwidth (same modulation format as IEEE 802.15.4g mandatory mode). Starting from the sensitivity limit across the given dynamic range.		±2		dB
Receiver sensitivity, 500 kbps	GFSK, 175-kHz deviation, 1.243-MHz RX bandwidth, BER = 10^{-2}		–97		dBm
Blocking, ±2 MHz, 500 kbps	Wanted signal 3 dB above sensitivity limit. 500 kbps, GFSK, 175-kHz deviation, 1.243-MHz RX bandwidth, BER = 10^{-2}		35, 36		dB

Receive (RX) Parameters, 861 MHz to 1054 MHz *(continued)*

Measured on the Texas Instruments CC1310EM-7XD-7793 reference design with T_c = 25°C, V_{DDS} = 3.0 V, DC/DC enabled, f_{RF} = 868 MHz, unless otherwise noted. All measurements are performed at the antenna input with a combined RX and TX path.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Blocking, ±10 MHz, 500 kbps	Wanted signal 3 dB above sensitivity limit. 500 kbps, GFSK, 175-kHz deviation, 1.243-MHz RX bandwidth, BER = 10^{-2}		55, 47		dB
Receiver sensitivity, long-range mode, 5 kbps	20 ksym/s, GFSK, 5-kHz deviation, FEC (half rate), DSSS = 2, 49-kHz RX bandwidth, BER = 10^{-2}. 868 MHz and 915 MHz		–119		dBm
Receiver sensitivity, long-range mode, 2.5 kbps	20 ksym/s, GFSK, 5-kHz deviation, FEC (half rate), DSSS = 4, 49-kHz RX bandwidth, BER = 10^{-2}. 868 MHz and 915 MHz		–120		dBm
Receiver sensitivity, long-range mode, 1.25 kbps	20 ksym/s, GFSK, 5-kHz deviation, FEC (half rate), DSSS = 8, 49-kHz RX bandwidth, BER = 10^{-2}. 868 MHz and 915 MHz		–121		dBm
Selectivity, ±100 kHz, long-range mode, 5 kbps	Wanted signal 3 dB above sensitivity limit. 20 ksym/s, GFSK, 5-kHz deviation, FEC (half rate), DSSS = 2, 49-kHz RX bandwidth, BER = 10^{-2}		47, 47		dB
Selectivity, ±200 kHz, long-range mode, 5 kbps	Wanted signal 3 dB above sensitivity limit. 20 ksym/s, GFSK, 5-kHz deviation, FEC (half rate), DSSS = 2, 49-kHz RX bandwidth, BER = 10^{-2}		54, 55		dB
Selectivity, ±300 kHz, long-range mode, 5 kbps	Wanted signal 3 dB above sensitivity limit. 20 ksym/s, GFSK, 5-kHz deviation, FEC (half rate), DSSS = 2, 49-kHz RX bandwidth, BER = 10^{-2}		57, 56		dB
Blocking, ±1 MHz, long-range mode, 5 kbps	Wanted signal 3 dB above sensitivity limit. 20 ksym/s, GFSK, 5-kHz deviation, FEC (half rate), DSSS = 2, 49-kHz RX bandwidth, BER = 10^{-2}		68, 67		dB
Blocking, ±2 MHz, long-range mode, 5 kbps	Wanted signal 3 dB above sensitivity limit. 20 ksym/s, GFSK, 5-kHz deviation, FEC (half rate), DSSS = 2, 49-kHz RX bandwidth, BER = 10^{-2}		74, 74		dB
Blocking, ±10 MHz, long-range mode, 5 kbps	Wanted signal 3 dB above sensitivity limit. 20 ksym/s, GFSK, 5-kHz deviation, FEC (half rate), DSSS = 2, 49-kHz RX bandwidth, BER = 10^{-2}		85, 85		dB
Image rejection (image compensation enabled, the image compensation is calibrated in production), long-range mode, 5 kbps	Wanted signal 3 dB above sensitivity limit. 20 ksym/s, GFSK, 5-kHz deviation, FEC (half rate), DSSS = 2, 49-kHz RX bandwidth, BER = 10^{-2}		52		dB
Receiver sensitivity, wM-BUS S2-mode, 32.768 kbps	f_{RF} = 868.3 MHz, 32.768 ksym/s, Manchester coding, FSK, 50-kHz deviation, 196-kHz RX bandwidth, BER = 10^{-2}		–111		dBm
Selectivity, ±200 kHz, wM-BUS S2-mode, 32.768 kbps	Wanted signal 3 dB above sensitivity limit. f_{RF} = 868.3 MHz, 32.768 ksym/s, Manchester coding, FSK, 50-kHz deviation, 196-kHz RX bandwidth, BER = 10^{-2}		42, 43		dB
Selectivity, ±400 kHz, wM-BUS S2-mode, 32.768 kbps	Wanted signal 3 dB above sensitivity limit. f_{RF} = 868.3 MHz, 32.768 ksym/s, Manchester coding, FSK, 50-kHz deviation, 196-kHz RX bandwidth, BER = 10^{-2}		41, 47		dB
Blocking, ±1 MHz, wM-BUS S2-mode, 32.768 kbps	Wanted signal 3 dB above sensitivity limit. f_{RF} = 868.3 MHz, 32.768 ksym/s, Manchester coding, FSK, 50-kHz deviation, 196-kHz RX bandwidth, BER = 10^{-2}		43, 52		dB
Blocking, ±2 MHz, wM-BUS S2-mode, 32.768 kbps	Wanted signal 3 dB above sensitivity limit. f_{RF} = 868.3 MHz, 32.768 ksym/s, Manchester coding, FSK, 50-kHz deviation, 196-kHz RX bandwidth, BER = 10^{-2}		52, 55		dB
Blocking, ±10 MHz, wM-BUS S2-mode, 32.768 kbps	Wanted signal 3 dB above sensitivity limit. f_{RF} = 868.3 MHz, 32.768 ksym/s, Manchester coding, FSK, 50-kHz deviation, 196-kHz RX bandwidth, BER = 10^{-2}		68, 72		dB

Receive (RX) Parameters, 861 MHz to 1054 MHz *(continued)*

Measured on the Texas Instruments CC1310EM-7XD-7793 reference design with T_c = 25°C, V_{DDS} = 3.0 V, DC/DC enabled, f_{RF} = 868 MHz, unless otherwise noted. All measurements are performed at the antenna input with a combined RX and TX path.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Image rejection (image compensation enabled, the image compensation is calibrated in production), wM-BUS S2-mode, 32.768 kbps	Wanted signal 3 dB above sensitivity limit. f_{RF} = 868.3 MHz, 32.768 ksym/s, Manchester coding, FSK, 50-kHz deviation, 196-kHz RX bandwidth, BER = 10^{-2}		43		dB
Receiver sensitivity, wM-BUS C-mode, 100 kbps	f_{RF} = 868.95 MHz, 100 ksym/s, NRZ coding, FSK, 45-kHz deviation, 196-kHz RX bandwidth, BER = 10^{-2}		–107		dBm
Selectivity, ±400 kHz, wM-BUS C-mode, 100 kbps	Wanted signal 3 dB above sensitivity limit. f_{RF} = 868.95 MHz, 100 ksym/s, NRZ coding, FSK, 45-kHz deviation, 196-kHz RX bandwidth, BER = 10^{-2}		41, 46		dB
Selectivity, ±800 kHz, wM-BUS C-mode, 100 kbps	Wanted signal 3 dB above sensitivity limit. f_{RF} = 868.95 MHz, 100 ksym/s, NRZ coding, FSK, 45-kHz deviation, 196-kHz RX bandwidth, BER = 10^{-2}		41, 50		dB
Blocking, ±1 MHz, wM-BUS C-mode, 100 kbps	Wanted signal 3 dB above sensitivity limit. f_{RF} = 868.95 MHz, 100 ksym/s, NRZ coding, FSK, 45-kHz deviation, 196-kHz RX bandwidth, BER = 10^{-2}		43, 51		dB
Blocking, ±2 MHz, wM-BUS C-mode, 100 kbps	Wanted signal 3 dB above sensitivity limit. f_{RF} = 868.95 MHz, 100 ksym/s, NRZ coding, FSK, 45-kHz deviation, 196-kHz RX bandwidth, BER = 10^{-2}		51, 53		dB
Blocking, ±5 MHz, wM-BUS C-mode, 100 kbps	Wanted signal 3 dB above sensitivity limit. f_{RF} = 868.95 MHz, 100 ksym/s, NRZ coding, FSK, 45-kHz deviation, 196-kHz RX bandwidth, BER = 10^{-2}		55, 61		dB
Blocking, ±10 MHz, wM-BUS C-mode, 100 kbps	Wanted signal 3 dB above sensitivity limit. f_{RF} = 868.95 MHz, 100 ksym/s, NRZ coding, FSK, 45-kHz deviation, 196-kHz RX bandwidth, BER = 10^{-2}		67, 68		dB
Receiver sensitivity, wM-BUS T-mode, 100 kbps	f_{RF} = 868.95 MHz, 100 ksym/s, 3 out of 6 coding, FSK, 45-kHz deviation, 196-kHz RX bandwidth, BER = 10^{-2}		–105		dBm
Selectivity, ±400 kHz, wM-BUS T-mode, 100 kbps	Wanted signal 3 dB above sensitivity limit. f_{RF} = 868.95 MHz, 100 ksym/s, 3 out of 6 coding, FSK, 45-kHz deviation, 196-kHz RX bandwidth, BER = 10^{-2}		41, 46		dB
Selectivity, ±800 kHz, wM-BUS T-mode, 100 kbps	Wanted signal 3 dB above sensitivity limit. f_{RF} = 868.95 MHz, 100 ksym/s, 3 out of 6 coding, FSK, 45-kHz deviation, 196-kHz RX bandwidth, BER = 10^{-2}		41, 50		dB
Blocking, ±1 MHz, wM-BUS T-mode, 100 kbps	Wanted signal 3 dB above sensitivity limit. f_{RF} = 868.95 MHz, 100 ksym/s, 3 out of 6 coding, FSK, 45-kHz deviation, 196-kHz RX bandwidth, BER = 10^{-2}		42, 51		dB
Blocking, ±2 MHz, wM-BUS T-mode, 100 kbps	Wanted signal 3 dB above sensitivity limit. f_{RF} = 868.95 MHz, 100 ksym/s, 3 out of 6 coding, FSK, 45-kHz deviation, 196-kHz RX bandwidth, BER = 10^{-2}		51, 52		dB
Blocking, ±5 MHz, wM-BUS T-mode, 100 kbps	Wanted signal 3 dB above sensitivity limit. f_{RF} = 868.95 MHz, 100 ksym/s, 3 out of 6 coding, FSK, 45-kHz deviation, 196-kHz RX bandwidth, BER = 10^{-2}		54, 60		dB
Blocking, ±10 MHz, wM-BUS T-mode, 100 kbps	Wanted signal 3 dB above sensitivity limit. f_{RF} = 868.95 MHz, 100 ksym/s, 3 out of 6 coding, FSK, 45-kHz deviation, 196-kHz RX bandwidth, BER = 10^{-2}		67, 68		dB
Receiver sensitivity, WideBand-DSSS (WB-DSSS), 30 kbps	f_{RF} = 915 MHz, 480 ksym/s, GFSK, 195-kHz deviation, FEC (half rate), DSSS = 8, 622-kHz RX bandwidth, BER = 10^{-2}		–109		dBm
Blocking, ±1 MHz, WB-DSSS, 30 kbps	Wanted signal 3 dB above sensitivity limit. f_{RF} = 915 MHz, 480 ksym/s, GFSK, 195-kHz deviation, FEC (half rate), DSSS = 8, 622-kHz RX bandwidth, BER = 10^{-2}		57, 57		dB
Blocking, ±2 MHz, WB-DSSS, 30 kbps	Wanted signal 3 dB above sensitivity limit. f_{RF} = 915 MHz, 480 ksym/s, GFSK, 195-kHz deviation, FEC (half rate), DSSS = 8, 622-kHz RX bandwidth, BER = 10^{-2}		58, 58		dB

Receive (RX) Parameters, 861 MHz to 1054 MHz *(continued)*

Measured on the Texas Instruments CC1310EM-7XD-7793 reference design with T_c = 25°C, V_{DDS} = 3.0 V, DC/DC enabled, f_{RF} = 868 MHz, unless otherwise noted. All measurements are performed at the antenna input with a combined RX and TX path.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Blocking, ±5 MHz, WB-DSSS, 30 kbps	Wanted signal 3 dB above sensitivity limit. f_{RF} = 915 MHz, 480 ksym/s, GFSK, 195-kHz deviation, FEC (half rate), DSSS = 8, 622-kHz RX bandwidth, BER = 10^{-2}		59, 57		dB
Blocking, ±10 MHz, WB-DSSS, 30 kbps	Wanted signal 3 dB above sensitivity limit. f_{RF} = 915 MHz, 480 ksym/s, GFSK, 195-kHz deviation, FEC (half rate), DSSS = 8, 622-kHz RX bandwidth, BER = 10^{-2}		71, 68		dB
Receiver sensitivity, WideBand-DSSS (WB-DSSS), 60 kbps	f_{RF} = 915 MHz, 480 ksym/s, GFSK, 195-kHz deviation, FEC (half rate), DSSS = 4, 622-kHz RX bandwidth, BER = 10^{-2}		–108		dBm
Blocking, ±1 MHz, WB-DSSS, 60 kbps	Wanted signal 3 dB above sensitivity limit. f_{RF} = 915 MHz, 480 ksym/s, GFSK, 195-kHz deviation, FEC (half rate), DSSS = 4, 622-kHz RX bandwidth, BER = 10^{-2}		56, 56		dB
Blocking, ±2 MHz, WB-DSSS, 60 kbps	Wanted signal 3 dB above sensitivity limit. f_{RF} = 915 MHz, 480 ksym/s, GFSK, 195-kHz deviation, FEC (half rate), DSSS = 4, 622-kHz RX bandwidth, BER = 10^{-2}		57, 57		dB
Blocking, ±5 MHz, WB-DSSS, 60 kbps	Wanted signal 3 dB above sensitivity limit. f_{RF} = 915 MHz, 480 ksym/s, GFSK, 195-kHz deviation, FEC (half rate), DSSS = 4, 622-kHz RX bandwidth, BER = 10^{-2}		57, 56		dB
Blocking, ±10 MHz, WB-DSSS, 60 kbps	Wanted signal 3 dB above sensitivity limit. f_{RF} = 915 MHz, 480 ksym/s, GFSK, 195-kHz deviation, FEC (half rate), DSSS = 4, 622-kHz RX bandwidth, BER = 10^{-2}		70, 67		dB
Receiver sensitivity, WideBand-DSSS (WB-DSSS), 120 kbps	f_{RF} = 915 MHz, 480 ksym/s, GFSK, 195-kHz deviation, FEC (half rate), DSSS = 2, 622-kHz RX bandwidth, BER = 10^{-2}		–106		dBm
Blocking, ±1 MHz, WB-DSSS, 120 kbps	Wanted signal 3 dB above sensitivity limit. f_{RF} = 915 MHz, 480 ksym/s, GFSK, 195-kHz deviation, FEC (half rate), DSSS = 2, 622-kHz RX bandwidth, BER = 10^{-2}		54, 54		dB
Blocking, ±2 MHz, WB-DSSS, 120 kbps	Wanted signal 3 dB above sensitivity limit. f_{RF} = 915 MHz, 480 ksym/s, GFSK, 195-kHz deviation, FEC (half rate), DSSS = 2, 622-kHz RX bandwidth, BER = 10^{-2}		55, 55		dB
Blocking, ±5 MHz, WB-DSSS, 120 kbps	Wanted signal 3 dB above sensitivity limit. f_{RF} = 915 MHz, 480 ksym/s, GFSK, 195-kHz deviation, FEC (half rate), DSSS = 2, 622-kHz RX bandwidth, BER = 10^{-2}		55, 54		dB
Blocking, ±10 MHz, WB-DSSS, 120 kbps	Wanted signal 3 dB above sensitivity limit. f_{RF} = 915 MHz, 480 ksym/s, GFSK, 195-kHz deviation, FEC (half rate), DSSS = 2, 622-kHz RX bandwidth, BER = 10^{-2}		69, 65		dB
Receiver sensitivity, WideBand-DSSS (WB-DSSS), 240 kbps	f_{RF} = 915 MHz, 480 ksym/s, GFSK, 195-kHz deviation, FEC (half rate), DSSS = 1, 622-kHz RX bandwidth, BER = 10^{-2}		–105		dBm
Blocking, ±1 MHz, WB-DSSS, 240 kbps	Wanted signal 3 dB above sensitivity limit. f_{RF} = 915 MHz, 480 ksym/s, GFSK, 195-kHz deviation, FEC (half rate), DSSS = 1, 622-kHz RX bandwidth, BER = 10^{-2}		53, 53		dB
Blocking, ±2 MHz, WB-DSSS, 240 kbps	Wanted signal 3 dB above sensitivity limit. f_{RF} = 915 MHz, 480 ksym/s, GFSK, 195-kHz deviation, FEC (half rate), DSSS = 1, 622-kHz RX bandwidth, BER = 10^{-2}		53, 54		dB

Receive (RX) Parameters, 861 MHz to 1054 MHz *(continued)*

Measured on the Texas Instruments CC1310EM-7XD-7793 reference design with T_c = 25°C, V_{DDS} = 3.0 V, DC/DC enabled, f_{RF} = 868 MHz, unless otherwise noted. All measurements are performed at the antenna input with a combined RX and TX path.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Blocking, ±5 MHz, WB-DSSS, 240 kbps	Wanted signal 3 dB above sensitivity limit. f_{RF} = 915 MHz, 480 ksym/s, GFSK, 195-kHz deviation, FEC (half rate), DSSS = 1, 622-kHz RX bandwidth, BER = 10^{-2}		53, 54		dB
Blocking, ±10 MHz, WB-DSSS, 240 kbps	Wanted signal 3 dB above sensitivity limit. f_{RF} = 915 MHz, 480 ksym/s, GFSK, 195-kHz deviation, FEC (half rate), DSSS = 1, 622-kHz RX bandwidth, BER = 10^{-2}		68, 64		dB
Receiver sensitivity, 10 kbps	GFSK, 19-kHz deviation, 78-kHz RX bandwidth, BER = 10^{-2}		–114		dBm
Selectivity, ±100 kHz, 10 kbps	Wanted signal 3 dB above sensitivity limit. 10 kbps, GFSK, 19-kHz deviation, 78-kHz RX bandwidth, BER = 10^{-2}		40, 40		dB
Selectivity, ±200 kHz, 10 kbps	Wanted signal 3 dB above sensitivity limit. 10 kbps, GFSK, 19-kHz deviation, 78-kHz RX bandwidth, BER = 10^{-2}		46, 44		dB
Selectivity, ±400 kHz, 10 kbps	Wanted signal 3 dB above sensitivity limit. 10 kbps, GFSK, 19-kHz deviation, 78-kHz RX bandwidth, BER = 10^{-2}		50, 45		dB
Blocking, ±2 MHz, 10 kbps	Wanted signal 3 dB above sensitivity limit. 10 kbps, GFSK, 19-kHz deviation, 78-kHz RX bandwidth, BER = 10^{-2}		62, 61		dB
Blocking, ±10 MHz, 10 kbps	Wanted signal 3 dB above sensitivity limit. 10 kbps, GFSK, 19-kHz deviation, 78-kHz RX bandwidth, BER = 10^{-2}		76, 72		dB
Image rejection (image compensation enabled, the image compensation is calibrated in production), 10 kbps	Wanted signal 3 dB above sensitivity limit. 10 kbps, GFSK, 19-kHz deviation, 78-kHz RX bandwidth, BER = 10^{-2}		43		dB
Receiver sensitivity, 4.8 kbps	GFSK, 5.2-kHz deviation, 49-kHz RX bandwidth, BER = 10^{-2}		–114		dBm
Selectivity, ±100 kHz, 4.8 kbps	Wanted signal 3 dB above sensitivity limit. 4.8 kbps, GFSK, 5.2-kHz deviation, 49-kHz RX bandwidth, BER = 10^{-2}		44, 43		dB
Selectivity, ±200 kHz, 4.8 kbps	Wanted signal 3 dB above sensitivity limit. 4.8 kbps, GFSK, 5.2-kHz deviation, 49-kHz RX bandwidth, BER = 10^{-2}		49, 48		dB
Selectivity, ±400 kHz, 4.8 kbps	Wanted signal 3 dB above sensitivity limit. 4.8 kbps, GFSK, 5.2-kHz deviation, 49-kHz RX bandwidth, BER = 10^{-2}		52, 49		dB
Blocking, ±2 MHz, 4.8 kbps	Wanted signal 3 dB above sensitivity limit. 4.8 kbps, GFSK, 5.2-kHz deviation, 49-kHz RX bandwidth, BER = 10^{-2}		64, 63		dB
Blocking, ±10 MHz, 4.8 kbps	Wanted signal 3 dB above sensitivity limit. 4.8 kbps, GFSK, 5.2-kHz deviation, 49-kHz RX bandwidth, BER = 10^{-2}		73, 72		dB
Image rejection (image compensation enabled, the image compensation is calibrated in production), 4.8 kbps	Wanted signal 3 dB above sensitivity limit. 4.8 kbps, GFSK, 5.2-kHz deviation, 49-kHz RX bandwidth, BER = 10^{-2}		43		dB
Receiver sensitivity, CC1101 compatible mode, 2.4 kbps	GFSK, 5.2-kHz deviation (commonly used settings on CC1101), 49-kHz RX bandwidth, BER = 10^{-2}		–116		dBm
Selectivity, ±100 kHz, CC1101 compatible mode, 2.4 kbps	Wanted signal 3 dB above sensitivity limit. 2.4 kbps, GFSK, 5.2-kHz deviation (commonly used settings on CC1101), 49-kHz RX bandwidth, BER = 10^{-2}		45, 44		dB
Selectivity, ±200 kHz, CC1101 compatible mode, 2.4 kbps	Wanted signal 3 dB above sensitivity limit. 2.4 kbps, GFSK, 5.2-kHz deviation (commonly used settings on CC1101), 49-kHz RX bandwidth, BER = 10^{-2}		51, 47		dB

Receive (RX) Parameters, 861 MHz to 1054 MHz *(continued)*

Measured on the Texas Instruments CC1310EM-7XD-7793 reference design with T_c = 25°C, V_{DDS} = 3.0 V, DC/DC enabled, f_{RF} = 868 MHz, unless otherwise noted. All measurements are performed at the antenna input with a combined RX and TX path.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Blocking, ±2 MHz, CC1101 compatible mode, 2.4 kbps	Wanted signal 3 dB above sensitivity limit. 2.4 kbps, GFSK, 5.2-kHz deviation (commonly used settings on CC1101), 49-kHz RX bandwidth, BER = 10^{-2}		63, 62		dB
Blocking, ±10 MHz, CC1101 compatible mode, 2.4 kbps	Wanted signal 3 dB above sensitivity limit. 2.4 kbps, GFSK, 5.2-kHz deviation (commonly used settings on CC1101), 49-kHz RX bandwidth, BER = 10^{-2}		76, 71		dB
Image rejection (image compensation enabled, the image compensation is calibrated in production), CC1101 compatible mode, 2.4 kbps	Wanted signal 3 dB above sensitivity limit. 2.4 kbps, GFSK, 5.2-kHz deviation (commonly used settings on CC1101), 49-kHz RX bandwidth, BER = 10^{-2}		45		dB
Receiver sensitivity, CC1101 compatible mode, 1.2 kbps	GFSK, 5.2-kHz deviation (commonly used settings on CC1101), 49-kHz RX bandwidth, BER = 10^{-2}		–117		dBm
Selectivity, ±100 kHz, CC1101 compatible mode, 1.2 kbps	Wanted signal 3 dB above sensitivity limit. 1.2 kbps, GFSK, 5.2-kHz deviation (commonly used settings on CC1101), 49-kHz RX bandwidth, BER = 10^{-2}		45, 44		dB
Selectivity, ±200 kHz, CC1101 compatible mode, 1.2 kbps	Wanted signal 3 dB above sensitivity limit. 1.2 kbps, GFSK, 5.2-kHz deviation (commonly used settings on CC1101), 49-kHz RX bandwidth, BER = 10^{-2}		51, 47		dB
Blocking, ±2 MHz, CC1101 compatible mode, 1.2 kbps	Wanted signal 3 dB above sensitivity limit. 1.2 kbps, GFSK, 5.2-kHz deviation (commonly used settings on CC1101), 49-kHz RX bandwidth, BER = 10^{-2}		63, 62		dB
Blocking, ±10 MHz, CC1101 compatible mode, 1.2 kbps	Wanted signal 3 dB above sensitivity limit. 1.2 kbps, GFSK, 5.2-kHz deviation (commonly used settings on CC1101), 49-kHz RX bandwidth, BER = 10^{-2}		81, 81		dB
Image rejection (image compensation enabled, the image compensation is calibrated in production), CC1101 compatible mode, 1.2 kbps	Wanted signal 3 dB above sensitivity limit. 1.2 kbps, GFSK, 5.2-kHz deviation (commonly used settings on CC1101), 49-kHz RX bandwidth, BER = 10^{-2}		45		dB
Receiver sensitivity, legacy long-range mode, 625 bps	10 ksym/s, GFSK, 5-kHz deviation, FEC (half rate), DSSS = 8, 40-kHz RX bandwidth, BER = 10^{-2}. 868 MHz and 915 MHz.		–124		dBm
Selectivity, ±100 kHz, legacy long-range mode, 625 bps	Wanted signal 3 dB above sensitivity limit. 10 ksym/s, GFSK, 5-kHz deviation, FEC (half rate), DSSS = 8, 40-kHz RX bandwidth, BER = 10^{-2}		56, 56		dB
Selectivity, ±200 kHz, legacy long-range mode, 625 bps	Wanted signal 3 dB above sensitivity limit. 10 ksym/s, GFSK, 5-kHz deviation, FEC (half rate), DSSS = 8, 40-kHz RX bandwidth, BER = 10^{-2}		62, 65		dB
Blocking ±1 MHz, legacy long-range mode, 625 bps	Wanted signal 3 dB above sensitivity limit. 10 ksym/s, GFSK, 5-kHz deviation, FEC (half rate), DSSS = 8, 40-kHz RX bandwidth, BER = 10^{-2}		73, 77		dB
Blocking ±2 MHz, legacy long-range mode, 625 bps	Wanted signal 3 dB above sensitivity limit. 10 ksym/s, GFSK, 5-kHz deviation, FEC (half rate), DSSS = 8, 40-kHz RX bandwidth, BER = 10^{-2}		79, 79		dB
Blocking ±10 MHz, legacy long-range mode, 625 bps	Wanted signal 3 dB above sensitivity limit. 10 ksym/s, GFSK, 5-kHz deviation, FEC (half rate), DSSS = 8, 40-kHz RX bandwidth, BER = 10^{-2}		91, 91		dB
Receiver sensitivity, OOK, 4.8 kbps	4.8 kbps, OOK, 40-kHz RX bandwidth, BER = 10^{-2}. 868 MHz and 915 MHz		–115		dBm

5.7 Receive (RX) Parameters, 431 MHz to 527 MHz

Measured on the Texas Instruments CC1310EM-7XD-4251 reference design with T_c = 25°C, V_{DDS} = 3.0 V, DC/DC enabled, f_{RF} = 433.92 MHz, unless otherwise noted. All measurements are performed at the antenna input with a combined RX and TX path. This frequency band is supported on die Revision B and later.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Receiver sensitivity, 50 kbps	50 kbps, GFSK, 25-kHz deviation, 100-kHz RX bandwidth (same modulation format as IEEE 802.15.4g mandatory mode), BER = 10^{-2}		–110		dBm
Receiver saturation	50 kbps, GFSK, 25-kHz deviation, 100-kHz RX bandwidth (same modulation format as IEEE 802.15.4g mandatory mode), BER = 10^{-2}		10		dBm
Selectivity, ±200 kHz, 50 kbps	Wanted signal 3 dB above sensitivity limit. 50 kbps, GFSK, 25-kHz deviation, 100-kHz RX bandwidth (same modulation format as IEEE 802.15.4g mandatory mode), BER = 10^{-2}		40, 42		dB
Selectivity, ±400 kHz, 50 kbps	Wanted signal 3 dB above sensitivity limit. 50 kbps, GFSK, 25-kHz deviation, 100-kHz RX bandwidth (same modulation format as IEEE 802.15.4g mandatory mode), BER = 10^{-2}		42, 50		dB
Blocking ±1 MHz, 50 kbps	Wanted signal 3 dB above sensitivity limit. 50 kbps, GFSK, 25-kHz deviation, 100-kHz RX bandwidth (same modulation format as IEEE 802.15.4g mandatory mode), BER = 10^{-2}		53, 58		dB
Blocking ±2 MHz, 50 kbps	Wanted signal 3 dB above sensitivity limit. 50 kbps, GFSK, 25-kHz deviation, 100-kHz RX bandwidth (same modulation format as IEEE 802.15.4g mandatory mode), BER = 10^{-2}		59, 60		dB
Blocking ±10 MHz, 50 kbps	Wanted signal 3 dB above sensitivity limit. 50 kbps, GFSK, 25-kHz deviation, 100-kHz RX bandwidth (same modulation format as IEEE 802.15.4g mandatory mode), BER = 10^{-2}		74, 74		dB
Spurious emissions 1 GHz to 13 GHz (VCO leakage at 3.5 GHz) and 30 MHz to 1 GHz	Conducted emissions measured according to ETSI EN 300 220		–74		dBm
Image rejection (image compensation enabled, the image compensation is calibrated in production), 50 kbps	Wanted signal 3 dB above sensitivity limit. 50 kbps, GFSK, 25-kHz deviation, 100-kHz RX bandwidth (same modulation format as IEEE 802.15.4g mandatory mode), BER = 10^{-2}		43		dB
Receiver sensitivity, long-range mode, 5 kbps	20 ksym/s, GFSK, 5-kHz deviation, FEC (half rate), DSSS = 2, 49-kHz RX bandwidth, BER = 10^{-2}. 433 MHz		–119		dBm
Receiver sensitivity, long-range mode, 2.5 kbps	20 ksym/s, GFSK, 5-kHz deviation, FEC (half rate), DSSS = 4, 49-kHz RX bandwidth, BER = 10^{-2}. 433 MHz		–120		dBm
Receiver sensitivity, long-range mode, 1.25 kbps	20 ksym/s, GFSK, 5-kHz deviation, FEC (half rate), DSSS = 8, 49-kHz RX bandwidth, BER = 10^{-2}. 433 MHz		–121		dBm
Receiver sensitivity, legacy long-range mode, 625 bps	10 ksym/s, GFSK, 5-kHz deviation, FEC (half rate), DSSS = 8, 40-kHz RX bandwidth, BER = 10^{-2}. 868 MHz and 915 MHZ.		–124		dBm
Selectivity, ±100 kHz, legacy long-range mode, 625 bps	Wanted signal 3 dB above sensitivity limit. 10 ksym/s, GFSK, 5-kHz deviation, FEC (half rate), DSSS = 8, 40-kHz RX bandwidth, BER = 10^{-2}		57, 58		dB
Selectivity, ±200 kHz, legacy long-range mode, 625 bps	Wanted signal 3 dB above sensitivity limit. 10 ksym/s, GFSK, 5-kHz deviation, FEC (half rate), DSSS = 8, 40-kHz RX bandwidth, BER = 10^{-2}		56, 60		dB
Blocking ±1 MHz, legacy long-range mode, 625 bps	Wanted signal 3 dB above sensitivity limit. 10 ksym/s, GFSK, 5-kHz deviation, FEC (half rate), DSSS = 8, 40-kHz RX bandwidth, BER = 10^{-2}		68, 73		dB
Blocking ±2 MHz, legacy long-range mode, 625 bps	Wanted signal 3 dB above sensitivity limit. 10 ksym/s, GFSK, 5-kHz deviation, FEC (half rate), DSSS = 8, 40-kHz RX bandwidth, BER = 10^{-2}		74, 74		dB
Blocking ±10 MHz, legacy long-range mode, 625 bps	Wanted signal 3 dB above sensitivity limit. 10 ksym/s, GFSK, 5-kHz deviation, FEC (half rate), DSSS = 8, 40-kHz RX bandwidth, BER = 10^{-2}		88, 89		dB

Receive (RX) Parameters, 431 MHz to 527 MHz *(continued)*

Measured on the Texas Instruments CC1310EM-7XD-4251 reference design with T_c = 25°C, V_{DDS} = 3.0 V, DC/DC enabled, f_{RF} = 433.92 MHz, unless otherwise noted. All measurements are performed at the antenna input with a combined RX and TX path. This frequency band is supported on die Revision B and later.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Image rejection (image compensation enabled, the image compensation is calibrated in production), legacy long-range mode, 625 bps	Wanted signal 3 dB above sensitivity limit. 10 ksym/s, GFSK, 5-kHz deviation, FEC (half rate), DSSS = 8, 40-kHz RX bandwidth, BER = 10^{-2}		55		dB

5.8 Transmit (TX) Parameters, 861 MHz to 1054 MHz

Measured on the Texas Instruments CC1310EM-7XD-7793 reference design with T_c = 25°C, V_{DDS} = 3.0 V, DC/DC enabled, f_{RF} = 868 MHz, unless otherwise noted. All measurements are performed at the antenna input with a combined RX and TX path.

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
Maximum output power, boost mode		VDDR = 1.95 V Minimum VDDS for boost mode is 2.1 V 868 MHz and 915 MHz		14		dBm
Maximum output power		868 MHz and 915 MHz		12		dBm
Output power programmable range				24		dB
Output power variation		Tested at +10-dBm setting		±0.9		dB
Output power variation, boost mode		+14 dBm		±0.5		dB
Spurious emissions (excluding harmonics)(1)	30 MHz to 1 GHz	Transmitting +14 dBm ETSI restricted bands		<–59		dBm
		Transmitting +14 dBm outside ETSI restricted bands		<–51		
	1 GHz to 12.75 GHz	Transmitting +14 dBm measured in 1-MHz bandwidth (ETSI)		<–37		
Harmonics	Second harmonic	Transmitting +14 dBm, conducted 868 MHz, 915 MHz		–52, –55		dBm
	Third harmonic	Transmitting +14 dBm, conducted 868 MHz, 915 MHz		–58, –55		
	Fourth harmonic	Transmitting +14 dBm, conducted 868 MHz, 915 MHz		–56, –56		
Spurious emissions out-of-band, 915 MHz(1)	30 MHz to 88 MHz (within FCC restricted bands)	Transmitting +14 dBm, conducted		<–66		dBm
	88 MHz to 216 MHz (within FCC restricted bands)	Transmitting +14 dBm, conducted		<–65		
	216 MHz to 960 MHz (within FCC restricted bands)	Transmitting +14 dBm, conducted		<–65		
	960 MHz to 2390 MHz and above 2483.5 MHz (within FCC restricted band)	Transmitting +14 dBm, conducted		<–52		
	1 GHz to 12.75 GHz (outside FCC restricted bands)	Transmitting +14 dBm, conducted		<–43		
Spurious emissions out-of-band, 920.6 MHz(1)	Below 710 MHz (ARIB T-108)	Transmitting +14 dBm, conducted		<–50		dBm
	710 MHz to 900 MHz (ARIB T-108)	Transmitting +14 dBm, conducted		<–60		
	900 MHz to 915 MHz (ARIB T-108)	Transmitting +14 dBm, conducted		<–57		
	930 MHz to 1000 MHz (ARIB T-108)	Transmitting +14 dBm, conducted		<–57		
	1000 MHz to 1215 MHz (ARIB T-108)	Transmitting +14 dBm, conducted		<–59		
	Above 1215 MHz (ARIB T-108)	Transmitting +14 dBm, conducted		<–45		

(1) Suitable for systems targeting compliance with EN 300 220, EN 54-25, EN 303 204, FCC CFR47 Part 15, ARIB STD-T108.

5.9 Transmit (TX) Parameters, 431 MHz to 527 MHz

Measured on the Texas Instruments CC1310EM-7XD-4251 reference design with T_c = 25°C, V_{DDS} = 3.0 V, DC/DC enabled, f_{RF} = 433.92 MHz, unless otherwise noted. All measurements are performed at the antenna input with a combined RX and TX path. This frequency band is supported on die Revision B and later.

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
Maximum output power, boost mode		VDDR = 1.95 V Minimum VDDS for boost mode is 2.1 V		15		dBm
Maximum output power				14		dBm
Spurious emissions (excluding harmonics)(1)	30 MHz to 1 GHz	Transmitting +10 dBm, 433 MHz Inside ETSI restricted bands		<–63		dBm
		Transmitting +10 dBm, 433 MHz Outside ETSI restricted bands		<–39		
	1 GHz to 12.75 GHz	Transmitting +10 dBm, 433 MHz Outside ETSI restricted bands, measured in 1-MHz bandwidth (ETSI)		<–52		
		Transmitting +10 dBm, 433 MHz Inside ETSI restricted bands, measured in 1-MHz bandwidth (ETSI)		<–58		

(1) Suitable for systems targeting compliance with EN 300 220, EN 54-25, EN 303 204, FCC CFR47 Part 15, ARIB STD-T108.

5.10 PLL Parameters

Measured on the Texas Instruments CC1310EM-7XD-7793 reference design with T_c = 25°C, V_{DDS} = 3.0 V

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Phase noise in the 868-MHz band	±100-kHz offset		–101		dBc/Hz
	±200-kHz offset		–108		
	±400-kHz offset		–115		
	±1000-kHz offset		–124		
	±2000-kHz offset		–131		
	±10000-kHz offset		–140		
Phase noise in the 915-MHz band	±100-kHz offset		–98		dBc/Hz
	±200-kHz offset		–106		
	±400-kHz offset		–114		
	±1000-kHz offset		–122		
	±2000-kHz offset		–130		
	±10000-kHz offset		–140		

5.11 ADC Characteristics

T_c = 25°C, V_{DDS} = 3.0 V, DC/DC disabled. Input voltage scaling enabled, unless otherwise noted.(1)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
	Input voltage range		0		V_{DDS}	V
	Resolution			12		Bits
	Sample rate				200	ksamples/s
	Offset	Internal 4.3-V equivalent reference(2)		2.1		LSB
	Gain error	Internal 4.3-V equivalent reference(2)		–0.14		LSB
DNL(3)	Differential nonlinearity			>–1		LSB
INL(4)	Integral nonlinearity			±2		LSB

(1) Using IEEE Std 1241™ 2010 for terminology and test methods.
(2) Input signal scaled down internally before conversion, as if voltage range was 0 to 4.3 V. Applied voltage must be within the absolute maximum ratings (see Section 5.1) at all times.
(3) No missing codes. Positive DNL typically varies from 0.3 to 1.7, depending on the device (see Figure 5-7).
(4) For a typical example, see Figure 5-6.

ADC Characteristics *(continued)*

T_c = 25°C, V_{DDS} = 3.0 V, DC/DC disabled. Input voltage scaling enabled, unless otherwise noted.(1)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
ENOB	Effective number of bits	Internal 4.3-V equivalent reference(2), 200 ksamples/s, 9.6-kHz input tone		10.0		Bits
		VDDS as reference, 200 ksamples/s, 9.6-kHz input tone		10.2		
		Internal 1.44-V reference, voltage scaling disabled, 32 samples average, 200 ksamples/s, 300-Hz input tone		11.1		
THD	Total harmonic distortion	Internal 4.3-V equivalent reference(2), 200 ksamples/s, 9.6-kHz input tone		–65		dB
		VDDS as reference, 200 ksamples/s, 9.6-kHz input tone		–72		
		Internal 1.44-V reference, voltage scaling disabled, 32 samples average, 200 ksamples/s, 300-Hz input tone		–75		
SINAD and SNDR	Signal-to-noise and distortion ratio	Internal 4.3-V equivalent reference(2), 200 ksamples/s, 9.6-kHz input tone		62		dB
		VDDS as reference, 200 ksamples/s, 9.6-kHz input tone		63		
		Internal 1.44-V reference, voltage scaling disabled, 32 samples average, 200 ksamples/s, 300-Hz input tone		69		
SFDR	Spurious-free dynamic range	Internal 4.3-V equivalent reference(2), 200 ksamples/s, 9.6-kHz input tone		74		dB
		VDDS as reference, 200 ksamples/s, 9.6-kHz input tone		75		
		Internal 1.44-V reference, voltage scaling disabled, 32 samples average, 200 ksamples/s, 300-Hz input tone		75		
	Conversion time	Including sampling time		5		µs
	Current consumption	Internal 4.3-V equivalent reference(2)		0.66		mA
	Current consumption	VDDS as reference		0.75		mA
	Reference voltage	Equivalent fixed internal reference(voltage scaling enabled) (2) For best accuracy, the ADC conversion should be initiated through the TI-RTOS API in order to include the gain/offset compensation factors stored in FCFG1.		4.3		V
	Reference voltage	Fixed internal reference (input voltage scaling disabled). (2) For best accuracy, the ADC conversion should be initiated through the TI-RTOS API in order to include the gain/offset compensation factors stored in FCFG1. This value is derived from the scaled value (4.3 V) as follows: V_{ref} = 4.3 V × 1408 / 4095		1.48		V
	Reference voltage	VDDS as reference (Also known as RELATIVE) (input voltage scaling enabled)		VDDS		V
	Reference voltage	VDDS as reference (Also known as RELATIVE) (input voltage scaling disabled)		VDDS / 2.82		V
	Input Impedance	200 ksamples/s, voltage scaling enabled. Capacitive input, input impedance depends on sampling frequency and sampling time		>1		MΩ

5.12 Temperature Sensor

Measured on the Texas Instruments CC1310EM-7XD-7793 reference design with T_c = 25°C, V_{DDS} = 3.0 V, unless otherwise noted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Resolution			4		°C
Range		–40		85	°C
Accuracy			±5		°C
Supply voltage coefficient(1)			3.2		°C/V

(1) Automatically compensated when using supplied driver libraries.

5.13 Battery Monitor

Measured on the Texas Instruments CC1310EM-7XD-7793 reference design with T_c = 25°C, V_{DDS} = 3.0 V, unless otherwise noted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Resolution			50		mV
Range		1.8		3.8	V
Accuracy			13		mV

5.14 Continuous Time Comparator

T_c = 25°C, V_{DDS} = 3.0 V, unless otherwise noted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Input voltage range		0		V_{DDS}	V
External reference voltage		0		V_{DDS}	V
Internal reference voltage	DCOUPL as reference		1.27		V
Offset			3		mV
Hysteresis			<2		mV
Decision time	Step from –10 mV to 10 mV		0.72		µs
Current consumption when enabled(1)			8.6		µA

(1) Additionally, the bias module must be enabled when running in standby mode.

5.15 Low-Power Clocked Comparator

T_c = 25°C, V_{DDS} = 3.0 V, unless otherwise noted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Input voltage range		0		V_{DDS}	V
Clock frequency			32.8		kHz
Internal reference voltage, VDDS / 2			1.49 to 1.51		V
Internal reference voltage, VDDS / 3			1.01 to 1.03		V
Internal reference voltage, VDDS / 4			0.78 to 0.79		V
Internal reference voltage, DCOUPL / 1			1.25 to 1.28		V
Internal reference voltage, DCOUPL / 2			0.63 to 0.65		V
Internal reference voltage, DCOUPL / 3			0.42 to 0.44		V
Internal reference voltage, DCOUPL / 4			0.33 to 0.34		V
Offset			<2		mV
Hysteresis			<5		mV
Decision time	Step from –50 mV to 50 mV		1		clock-cycle
Current consumption when enabled			362		nA

5.16 Programmable Current Source

T_c = 25°C, V_{DDS} = 3.0 V, unless otherwise noted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Current source programmable output range			0.25 to 20		µA
Resolution			0.25		µA
Current consumption(1)	Including current source at maximum programmable output		23		µA

(1) Additionally, the bias module must be enabled when running in standby mode.

5.17 DC Characteristics

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
T_A = 25°C, V_{DDS} = 1.8 V					
GPIO VOH at 8-mA load	IOCURR = 2, high-drive GPIOs only	1.32	1.54		V
GPIO VOL at 8-mA load	IOCURR = 2, high-drive GPIOs only		0.26	0.32	V
GPIO VOH at 4-mA load	IOCURR = 1	1.32	1.58		V
GPIO VOL at 4-mA load	IOCURR = 1		0.21	0.32	V
GPIO pullup current	Input mode, pullup enabled, Vpad = 0 V		71.7		µA
GPIO pulldown current	Input mode, pulldown enabled, Vpad = VDDS		21.1		µA
GPIO high/low input transition, no hysteresis	IH = 0, transition between reading 0 and reading 1		0.88		V
GPIO low-to-high input transition, with hysteresis	IH = 1, transition voltage for input read as 0 → 1		1.07		V
GPIO high-to-low input transition, with hysteresis	IH = 1, transition voltage for input read as 1 → 0		0.74		V
GPIO input hysteresis	IH = 1, difference between 0 → 1 and 1 → 0 voltage transition points		0.33		V
T_A = 25°C, V_{DDS} = 3.0 V					
GPIO VOH at 8-mA load	IOCURR = 2, high-drive GPIOs only		2.68		V
GPIO VOL at 8-mA load	IOCURR = 2, high-drive GPIOs only		0.33		V
GPIO VOH at 4-mA load	IOCURR = 1		2.72		V
GPIO VOL at 4-mA load	IOCURR = 1		0.28		V
T_A = 25°C, V_{DDS} = 3.8 V					
GPIO pullup current	Input mode, pullup enabled, Vpad = 0 V		277		µA
GPIO pulldown current	Input mode, pulldown enabled, Vpad = VDDS		113		µA
GPIO high/low input transition, no hysteresis	IH = 0, transition between reading 0 and reading 1		1.67		V
GPIO low-to-high input transition, with hysteresis	IH = 1, transition voltage for input read as 0 → 1		1.94		V
GPIO high-to-low input transition, with hysteresis	IH = 1, transition voltage for input read as 1 → 0		1.54		V
GPIO input hysteresis	IH = 1, difference between 0 → 1 and 1 → 0 voltage transition points		0.4		V
VIH	Lowest GPIO input voltage reliably interpreted as a *High*			0.8	VDDS(1)
VIL	Highest GPIO input voltage reliably interpreted as a *Low*	0.2			VDDS(1)

(1) Each GPIO is referenced to a specific VDDS pin. See the technical reference manual listed in Section 8.3 for more details.

5.18 Thermal Characteristics

THERMAL METRIC[1]		CC1310			UNIT[2]
		RSM (VQFN)	RHB (VQFN)	RGZ (VQFN)	
		32 PINS	32 PINS	48 PINS	
$R_{\theta JA}$	Junction-to-ambient thermal resistance	36.9	32.8	29.6	°C/W
$R_{\theta JC(top)}$	Junction-to-case (top) thermal resistance	30.3	24.0	15.7	°C/W
$R_{\theta JB}$	Junction-to-board thermal resistance	7.6	6.8	6.2	°C/W
ψ_{JT}	Junction-to-top characterization parameter	0.4	0.3	0.3	°C/W
ψ_{JB}	Junction-to-board characterization parameter	7.4	6.8	6.2	°C/W
$R_{\theta JC(bot)}$	Junction-to-case (bottom) thermal resistance	2.1	1.9	1.9	°C/W

(1) For more information about traditional and new thermal metrics, see *Semiconductor and IC Package Thermal Metrics*.
(2) °C/W = degrees Celsius per watt.

5.19 Timing and Switching Characteristics

5.19.1 Reset Timing

Table 5-1. Reset Timing

PARAMETER	MIN	TYP	MAX	UNIT
RESET_N low duration	1			µs

5.19.2 Wakeup Timing

Table 5-2. Wakeup Timing

Measured on the Texas Instruments CC1310EM-7XD-7793 reference design with T_c = 25°C, V_{DDS} = 3.0 V, unless otherwise noted. The times listed here do not include RTOS overhead.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
MCU, Idle → Active			14		µs
MCU, Standby → Active			174		µs
MCU, Shutdown → Active			1097		µs

5.19.3 Clock Specifications

Table 5-3. 24-MHz Crystal Oscillator (XOSC_HF)

T_c = 25°C, V_{DDS} = 3.0 V, unless otherwise noted.Section 5.19.1

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
ESR equivalent series resistanceSection 5.19.2	6 pF < C_L ≤ 9 pF		20	60	Ω
ESR equivalent series resistanceSection 5.19.2	5 pF < C_L ≤ 6 pF			80	Ω
L_M motional inductanceSection 5.19.2	Relates to load capacitance (C_L in Farads)		< 1.6 × 10^{-24} / C_L ^{2}		H
C_L crystal load capacitanceSection 5.19.2		5		9	pF
Crystal frequencySection 5.19.2			24		MHz
Start-up time			150		µs

Table 5-4. 32.768-kHz Crystal Oscillator (XOSC_LF)

Measured on the Texas Instruments CC1310EM-7XD-7793 reference design with T_c = 25°C, V_{DDS} = 3.0 V, unless otherwise noted.(1)

	MIN	TYP	MAX	UNIT
Crystal frequency		32.768		kHz
ESR equivalent series resistance		30	100	kΩ
Crystal load capacitance (C_L)	6		12	pF

(1) Probing or otherwise stopping the crystal while the DC/DC converter is enabled may cause permanent damage to the device.

Table 5-5. 48-MHz RC Oscillator (RCOSC_HF)

Measured on the Texas Instruments CC1310EM-7XD-7793 reference design with T_c = 25°C, V_{DDS} = 3.0 V, unless otherwise noted.

	MIN	TYP	MAX	UNIT
Frequency		48		MHz
Uncalibrated frequency accuracy		±1%		
Calibrated frequency accuracy(1)		±0.25%		
Startup time		5		µs

(1) Accuracy relative to the calibration source (XOSC_HF)

Table 5-6. 32-kHz RC Oscillator (RCOSC_LF)

Measured on the Texas Instruments CC1310EM-7XD-7793 reference design with T_c = 25°C, V_{DDS} = 3.0 V, unless otherwise noted.

	MIN	TYP	MAX	UNIT
Calibrated frequency(1)		32.768		kHz
Temperature coefficient		50		ppm/°C

(1) The frequency accuracy of the Real Time Clock (RTC) is not directly dependent on the frequency accuracy of the 32-kHz RC Oscillator. The RTC can be calibrated by measuring the frequency error of RCOSC_LF relative to XOSC_HF and compensating for the RTC tick speed.

5.19.4 Flash Memory Characteristics

Table 5-7. Flash Memory Characteristics

T_c = 25°C, V_{DDS} = 3.0 V, unless otherwise noted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Supported flash erase cycles before failure		100			k Cycles
Flash page or sector erase current	Average delta current		12.6		mA
Flash page or sector erase time(1)			8		ms
Flash page or sector size			4		KB
Flash write current	Average delta current, 4 bytes at a time		8.15		mA
Flash write time(1)	4 bytes at a time		8		µs

(1) This number is dependent on flash aging and increases over time and erase cycles.

5.19.5 Synchronous Serial Interface (SSI) Characteristics

Table 5-8. Synchronous Serial Interface (SSI) Characteristics

T_c = 25°C, V_{DDS} = 3.0 V, unless otherwise noted.

PARAMETER NO.	PARAMETER		MIN	TYP	MAX	UNIT
S1	t_{clk_per}	SSIClk cycle time	12		65024	system clocks
S2(1)	t_{clk_high}	SSIClk high time		0.5 × t_{clk_per}		
S3(1)	t_{clk_low}	SSIClk low time		0.5 × t_{clk_per}		

(1) See the SSI timing diagrams, Figure 5-1, Figure 5-2, and Figure 5-3.

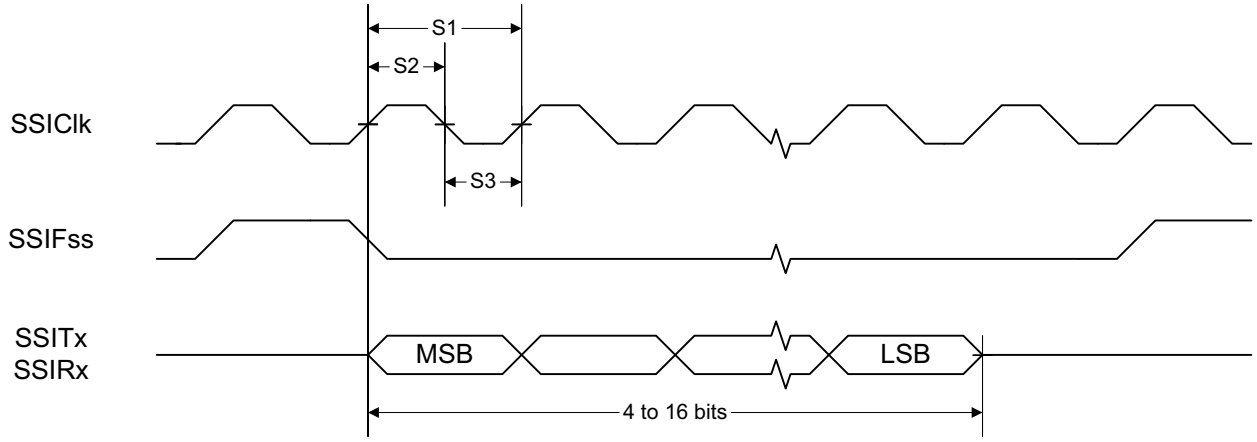


Figure 5-1. SSI Timing for TI Frame Format (FRF = 01), Single Transfer Timing Measurement

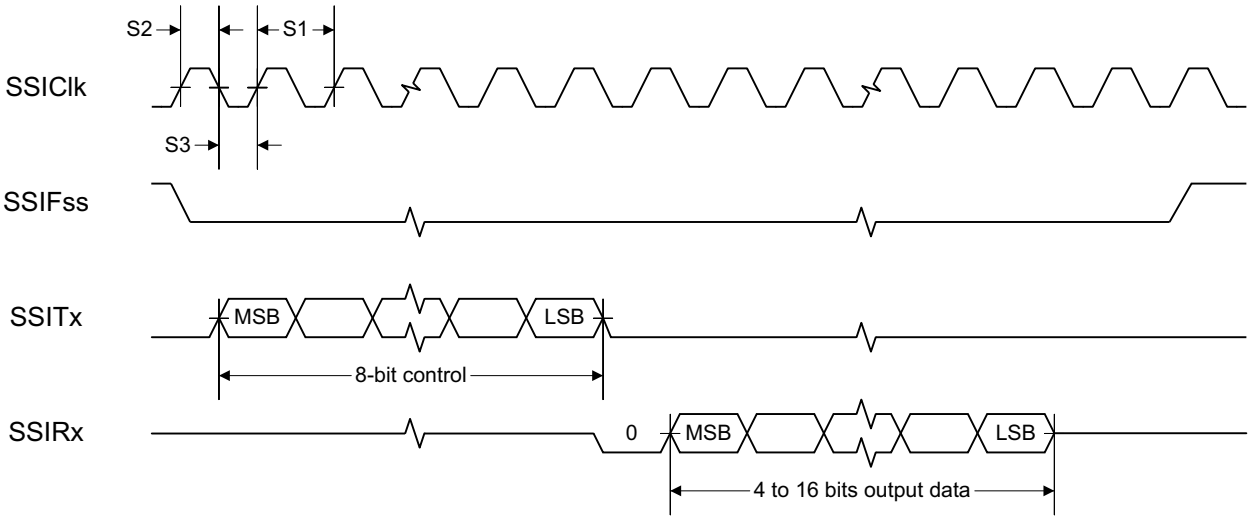


Figure 5-2. SSI Timing for MICROWIRE Frame Format (FRF = 10), Single Transfer

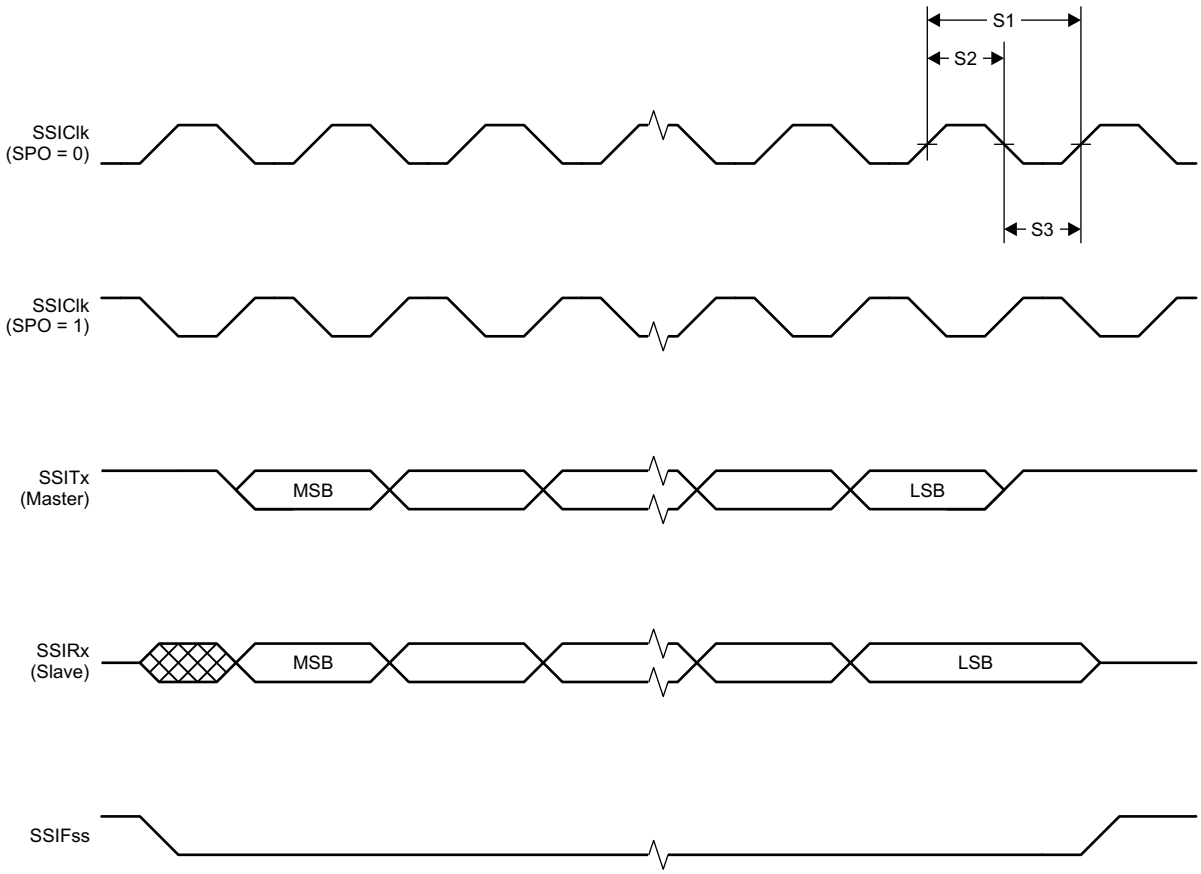


Figure 5-3. SSI Timing for SPI Frame Format (FRF = 00), With SPH = 1

5.20 Typical Characteristics

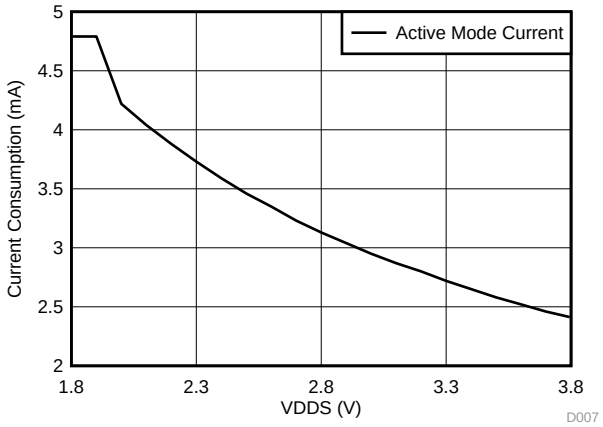


Figure 5-4. Active Mode (MCU) Current Consumption vs Supply Voltage (VDDS)

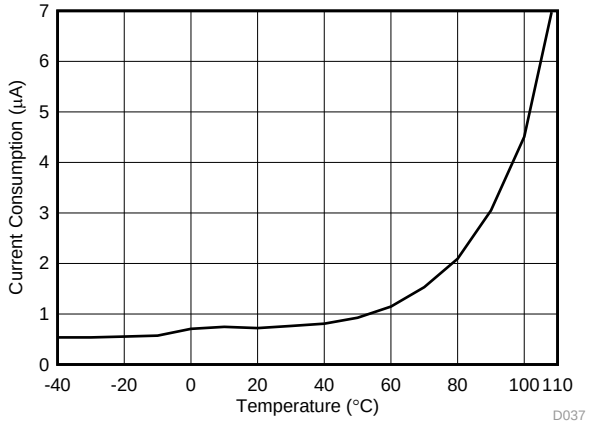


Figure 5-5. Standby MCU Current Consumption, 32-kHz Clock, RAM and MCU Retention

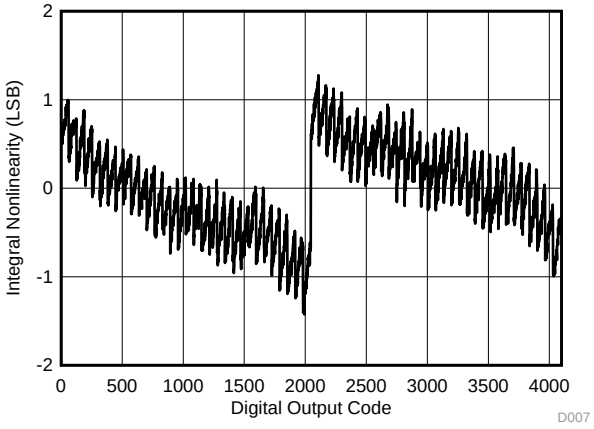


Figure 5-6. SoC ADC, Integral Nonlinearity vs Digital Output Code

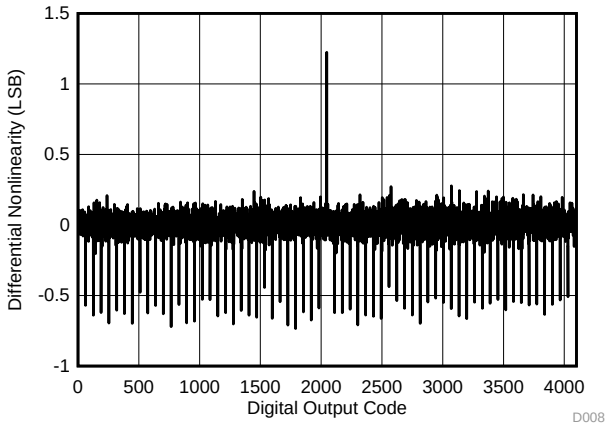


Figure 5-7. SoC ADC, Differential Nonlinearity vs Digital Output Code

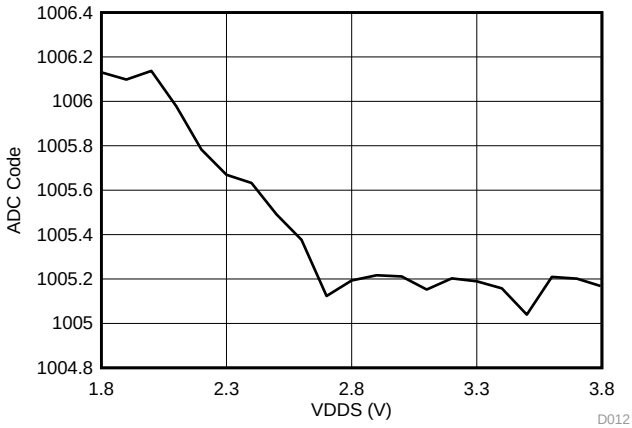


Figure 5-8. SoC ADC Output vs Supply Voltage (Fixed Input, Internal Reference, No Scaling)

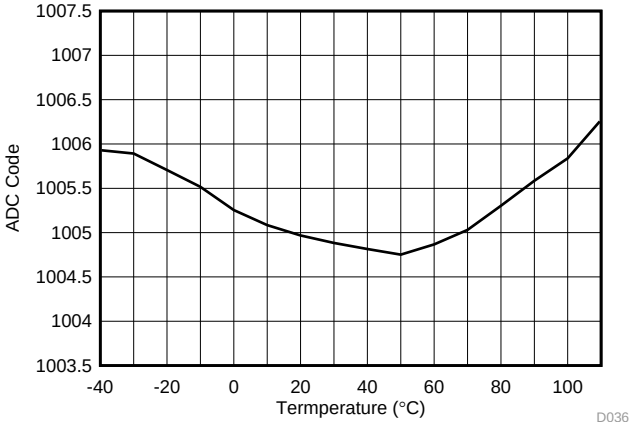


Figure 5-9. SoC ADC Output vs Temperature (Fixed Input, Internal Reference, No Scaling)

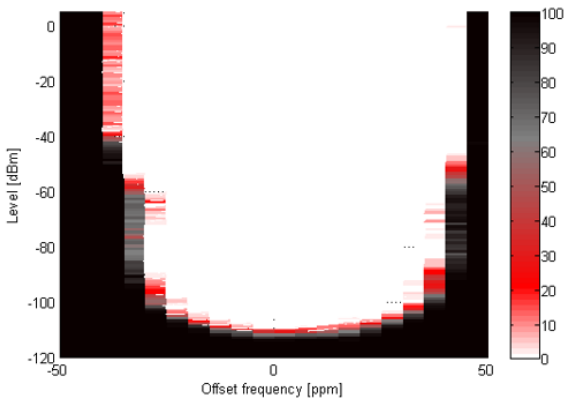


Figure 5-10. RX, (50-kbps) Packet Error Rate (PER) vs Input RF Level vs Frequency Offset, 868 MHz

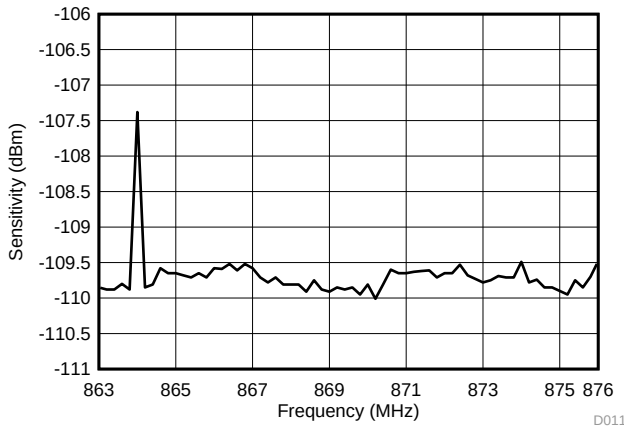


Figure 5-11. RX (50-kbps) Sensitivity vs Frequency

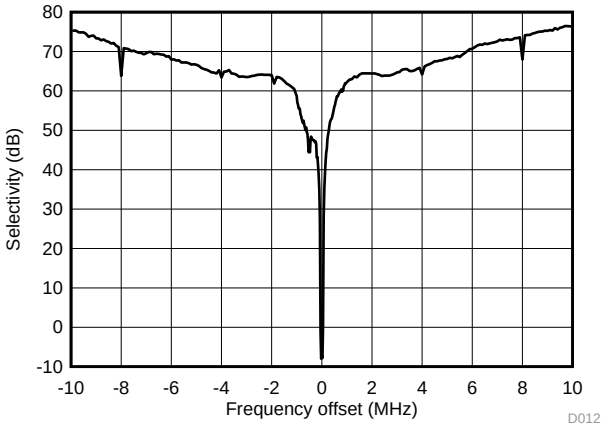


Figure 5-12. RX (50-kbps) Selectivity 868 MHz

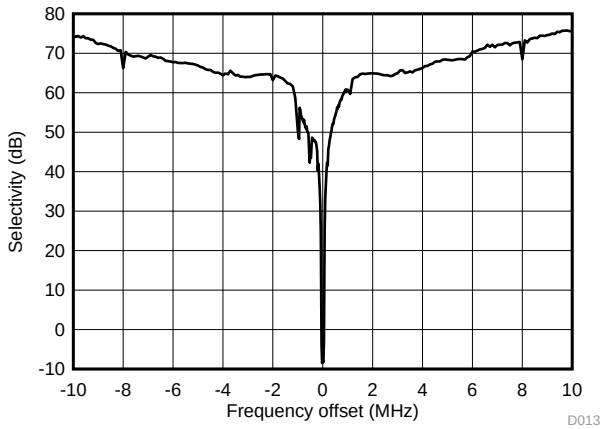


Figure 5-13. RX (50-kbps) Selectivity 915 MHz

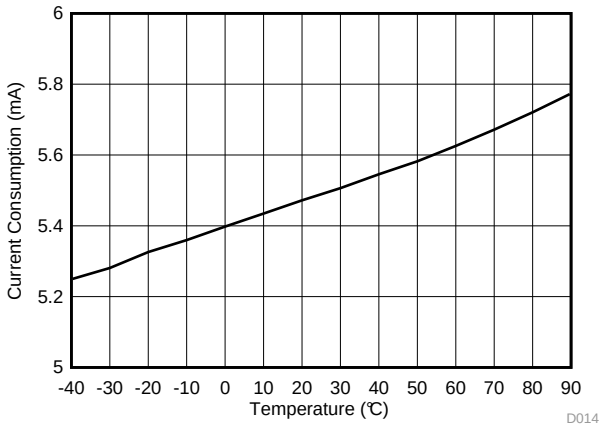


Figure 5-14. RX (50-kbps) Current Consumption vs Temperature 868 MHz

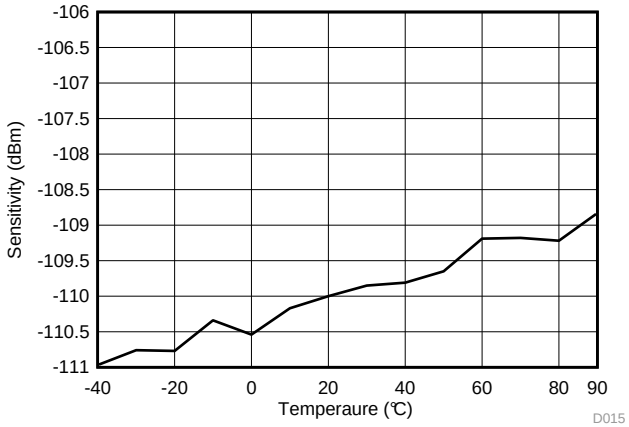


Figure 5-15. RX (50-kbps) Sensitivity vs Temperature 868 MHz

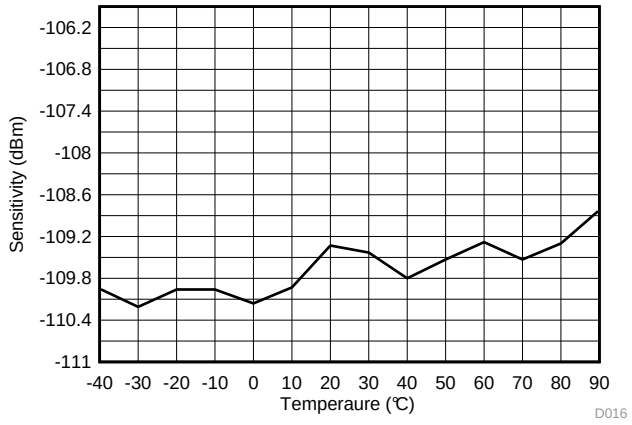


Figure 5-16. RX (50-kbps) Sensitivity vs Temperature 915 MHz

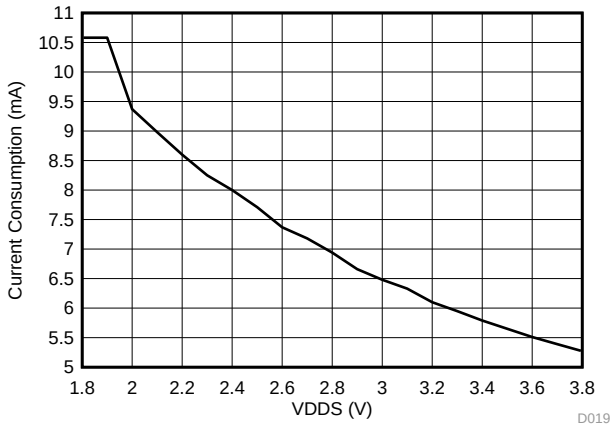


Figure 5-17. RX (50-kbps) Current Consumption vs Supply Voltage 915 MHz

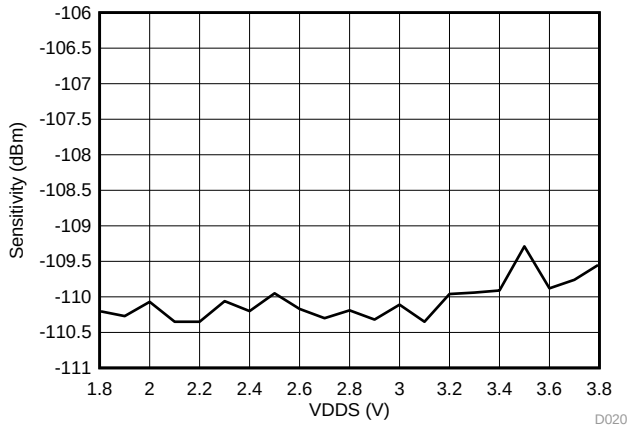
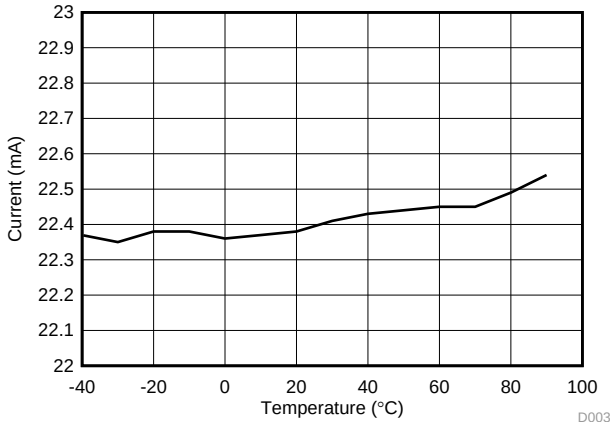


Figure 5-18. RX (50-kbps) Sensitivity vs Supply Voltage 868 MHz



DCDC On, 3.6 V

Figure 5-19. TX Current Consumption With Maximum Output Power vs Temperature 868 MHz

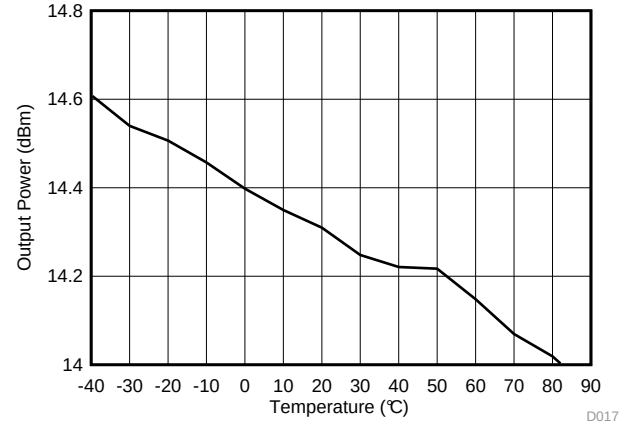


Figure 5-20. TX Maximum Output vs Temperature 868 MHz

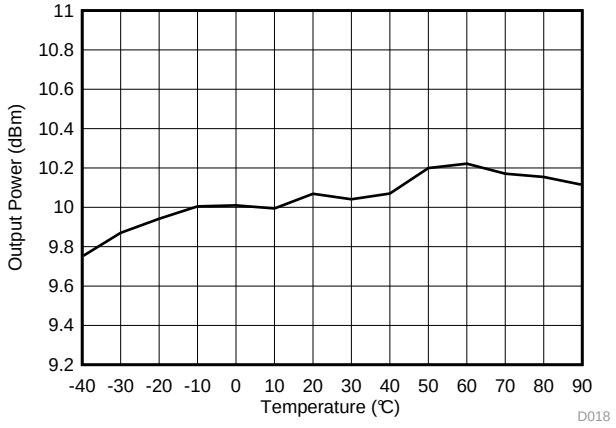


Figure 5-21. TX 10-dBm Output Power vs Temperature 868 MHz

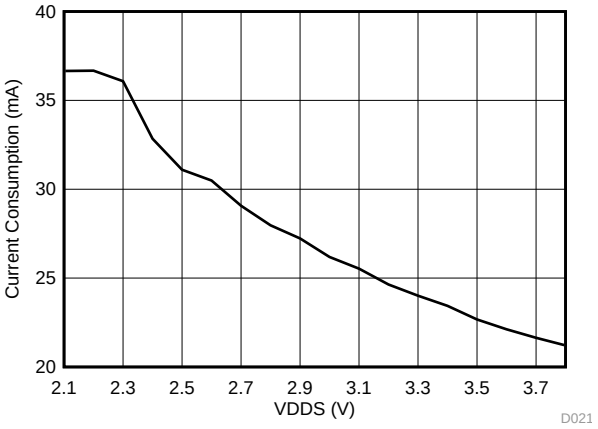


Figure 5-22. TX Current Consumption Maximum Output Power vs Supply Voltage 868 MHz

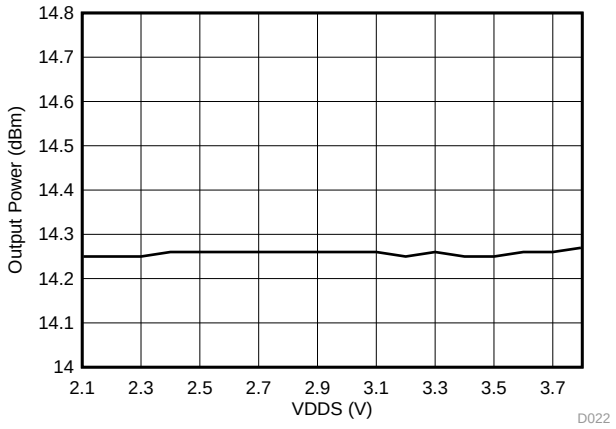


Figure 5-23. TX Maximum Output Power vs Supply Voltage 915 MHz

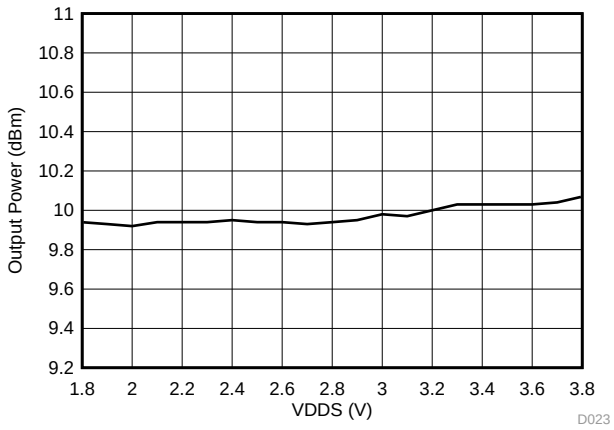


Figure 5-24. TX 10-dBm Output Power vs Supply Voltage 868 MHz

6 Detailed Description

6.1 Overview

Section 1.4 shows a block diagram of the core modules of the CC13xx product family.

6.2 Main CPU

The CC1310 SimpleLink Wireless MCU contains an ARM Cortex-M3 (CM3) 32-bit CPU, which runs the application and the higher layers of the protocol stack.

The CM3 processor provides a high-performance, low-cost platform that meets the system requirements of minimal memory implementation and low-power consumption, while delivering outstanding computational performance and exceptional system response to interrupts.

The CM3 features include the following:

- 32-bit ARM Cortex-M3 architecture optimized for small-footprint embedded applications
- Outstanding processing performance combined with fast interrupt handling
- ARM Thumb®-2 mixed 16- and 32-bit instruction set delivers the high performance expected of a 32-bit ARM core in a compact memory size usually associated with 8- and 16-bit devices, typically in the range of a few kilobytes of memory for microcontroller-class applications:
  - Single-cycle multiply instruction and hardware divide
  - Atomic bit manipulation (bit-banding), delivering maximum memory use and streamlined peripheral control
  - Unaligned data access, enabling data to be efficiently packed into memory
- Fast code execution permits slower processor clock or increases sleep mode time
- Harvard architecture characterized by separate buses for instruction and data
- Efficient processor core, system, and memories
- Hardware division and fast digital-signal-processing oriented multiply accumulate
- Saturating arithmetic for signal processing
- Deterministic, high-performance interrupt handling for time-critical applications
- Enhanced system debug with extensive breakpoint and trace capabilities
- Serial wire trace reduces the number of pins required for debugging and tracing
- Migration from the ARM7™ processor family for better performance and power efficiency
- Optimized for single-cycle flash memory use
- Ultra-low power consumption with integrated sleep modes
- 1.25 DMIPS per MHz

6.3 RF Core

The RF core is a highly flexible and capable radio system that interfaces the analog RF and baseband circuits, handles data to and from the system side, and assembles the information bits in a given packet structure.

The RF core can autonomously handle the time-critical aspects of the radio protocols, thus offloading the main CPU and leaving more resources for the user application. The RF core offers a high-level, command-based API to the main CPU.

The RF core supports a wide range of modulation formats, frequency bands, and accelerator features, which include the following:

- Wide range of data rates:
  - From 625 bps (offering long range and high robustness) to as high as 4 Mbps
- Wide range of modulation formats:
  - Multilevel (G) FSK and MSK
  - On-Off Keying (OOK) with optimized shaping to minimize adjacent channel leakage
  - Coding-gain support for long range
- Dedicated packet handling accelerators:
  - Forward error correction
  - Data whitening
  - 802.15.4g mode-switch support
  - Automatic CRC
- Automatic listen-before-talk (LBT) and clear channel assist (CCA)
- Digital RSSI
- Highly configurable channel filtering, supporting channel spacing schemes from 40 kHz to 4 MHz
- High degree of flexibility, offering a future-proof solution

The RF core interfaces a highly flexible radio, with a high-performance synthesizer that can support a wide range of frequency bands.

6.4 Sensor Controller

The Sensor Controller contains circuitry that can be selectively enabled in standby mode. The peripherals in this domain may be controlled by the Sensor Controller Engine, which is a proprietary power-optimized CPU. This CPU can read and monitor sensors or perform other tasks autonomously; thereby significantly reducing power consumption and offloading the main CM3 CPU.

A PC-based development tool called *Sensor Controller Studio* is used to write, test, and debug code for the Sensor Controller. The tool produces C driver source code, which the System CPU application uses to control and exchange data with the Sensor Controller. Typical use cases may be (but are not limited to) the following:

- Analog sensors using integrated ADC
- Digital sensors using GPIOs with bit-banged I^2C or SPI
- Capacitive sensing
- Waveform generation
- Pulse counting
- Key scan
- Quadrature decoder for polling rotational sensors

The peripherals in the Sensor Controller include the following:

- The low-power clocked comparator can be used to wake the device from any state in which the comparator is active. A configurable internal reference can be used with the comparator. The output of the comparator can also be used to trigger an interrupt or the ADC.
- Capacitive sensing functionality is implemented through the use of a constant current source, a time-to-digital converter, and a comparator. The continuous time comparator in this block can also be used as a higher-accuracy alternative to the low-power clocked comparator. The Sensor Controller takes care of baseline tracking, hysteresis, filtering, and other related functions.
- The ADC is a 12-bit, 200-ksamples/s ADC with 8 inputs and a built-in voltage reference. The ADC can be triggered by many different sources, including timers, I/O pins, software, the analog comparator, and the RTC.
- The analog modules can be connected to up to eight different GPIOs (see Table 6-1).

The peripherals in the Sensor Controller can also be controlled from the main application processor.

Table 6-1. GPIOs Connected to the Sensor Controller(1)

ANALOG CAPABLE	CC13x0		
	7 × 7 RGZ DIO NUMBER	5 × 5 RHB DIO NUMBER	4 × 4 RSM DIO NUMBER
Y	30	14	
Y	29	13	
Y	28	12	
Y	27	11	9
Y	26	9	8
Y	25	10	7
Y	24	8	6
Y	23	7	5
N	7	4	2
N	6	3	1
N	5	2	0
N	4	1	
N	3	0	
N	2		
N	1		
N	0		

(1) Depending on the package size, up to 15 pins can be connected to the Sensor Controller. Up to eight of these pins can be connected to analog modules.

6.5 Memory

The flash memory provides nonvolatile storage for code and data. The flash memory is in-system programmable.

The SRAM (static RAM) is split into two 4-KB blocks and two 6-KB blocks and can be used to store data and execute code. Retention of the RAM contents in standby mode can be enabled or disabled individually for each block to minimize power consumption. In addition, if flash cache is disabled, the 8-KB cache can be used as general-purpose RAM.

The ROM provides preprogrammed, embedded TI-RTOS kernel and Driverlib. The ROM also contains a bootloader that can be used to reprogram the device using SPI or UART.

6.6 Debug

The on-chip debug support is done through a dedicated cJTAG (IEEE 1149.7) or JTAG (IEEE 1149.1) interface.

6.7 Power Management

To minimize power consumption, the CC1310 device supports a number of power modes and power-management features (see Table 6-2).

Table 6-2. Power Modes

MODE	SOFTWARE-CONFIGURABLE POWER MODES				RESET PIN HELD
	ACTIVE	IDLE	STANDBY	SHUTDOWN	
CPU	Active	Off	Off	Off	Off
Flash	On	Available	Off	Off	Off
SRAM	On	On	On	Off	Off
Radio	Available	Available	Off	Off	Off
Supply System	On	On	Duty Cycled	Off	Off
Current	1.2 mA + 25.5 µA/MHz	570 µA	0.6 µA	185 nA	0.1 µA
Wake-up Time to CPU Active(1)	–	14 µs	174 µs	1015 µs	1015 µs
Register Retention	Full	Full	Partial	No	No
SRAM Retention	Full	Full	Full	No	No
High-Speed Clock	XOSC_HF or RCOSC_HF	XOSC_HF or RCOSC_HF	Off	Off	Off
Low-Speed Clock	XOSC_LF or RCOSC_LF	XOSC_LF or RCOSC_LF	XOSC_LF or RCOSC_LF	Off	Off
Peripherals	Available	Available	Off	Off	Off
Sensor Controller	Available	Available	Available	Off	Off
Wake-up on RTC	Available	Available	Available	Off	Off
Wake-up on Pin Edge	Available	Available	Available	Available	Off
Wake-up on Reset Pin	Available	Available	Available	Available	Available
Brown Out Detector (BOD)	Active	Active	Duty Cycled(2)	Off	N/A
Power On Reset (POR)	Active	Active	Active	Active	N/A

(1) Not including RTOS overhead.
(2) The Brown Out Detector is disabled between recharge periods in STANDBY. Lowering the supply voltage below the BOD threshold between two recharge periods while in STANDBY may cause the BOD to lock the device upon wakeup until a Reset/POR releases it. To avoid this, it is recommended that STANDBY mode is avoided if there is a risk that the supply voltage (VDDS) may drop below the specified operating voltage range. For the same reason, it is also good practice to ensure that a power cycling operation, such as a battery replacement, triggers a Power-on-reset by ensuring that the VDDS decoupling network is fully depleted before applying supply voltage again (for example, inserting new batteries). This restriction does not apply to CC1310 die revision B or later.

In active mode, the application CM3 CPU is actively executing code. Active mode provides normal operation of the processor and all of the peripherals that are currently enabled. The system clock can be any available clock source (see Table 6-2).

In idle mode, all active peripherals can be clocked, but the Application CPU core and memory are not clocked and no code is executed. Any interrupt event returns the processor to active mode.

In standby mode, only the always-on (AON) domain is active. An external wake-up event, RTC event, or Sensor Controller event is required to return the device to active mode. MCU peripherals with retention do not need to be reconfigured when waking up again, and the CPU continues execution from where it went into standby mode. All GPIOs are latched in standby mode.

In shutdown mode, the device is entirely turned off (including the AON domain and Sensor Controller), and the I/Os are latched with the value they had before entering shutdown mode. A change of state on any I/O pin defined as a *wake from shutdown pin* wakes up the device and functions as a reset trigger. The CPU can differentiate between reset in this way and reset-by-reset pin or POR by reading the reset status register. The only state retained in this mode is the latched I/O state and the flash memory contents.

The Sensor Controller is an autonomous processor that can control the peripherals in the Sensor Controller independent of the main CPU. This means that the main CPU does not have to wake up, for example to execute an ADC sample or poll a digital sensor over SPI, thus saving both current and wake-up time that would otherwise be wasted. The Sensor Controller Studio lets the user configure the Sensor Controller and choose which peripherals are controlled and which conditions wake up the main CPU.

6.8 Clock Systems

The CC1310 device supports two external and two internal clock sources.

A 24-MHz external crystal is required as the frequency reference for the radio. This signal is doubled internally to create a 48-MHz clock.

The 32.768-kHz crystal is optional. The low-speed crystal oscillator is designed for use with a 32.768-kHz watch-type crystal.

The internal high-speed RC oscillator (48-MHz) can be used as a clock source for the CPU subsystem.

The internal low-speed RC oscillator (32-kHz) can be used as a reference if the low-power crystal oscillator is not used.

The 32-kHz clock source can be used as external clocking reference through GPIO.

6.9 General Peripherals and Modules

The I/O controller controls the digital I/O pins and contains multiplexer circuitry to assign a set of peripherals to I/O pins in a flexible manner. All digital I/Os are interrupt and wake-up capable, have a programmable pullup and pulldown function, and can generate an interrupt on a negative or positive edge (configurable). When configured as an output, pins can function as either push-pull or open-drain. Five GPIOs have high-drive capabilities, which are marked in **bold** in Section 4.

The SSIs are synchronous serial interfaces that are compatible with SPI, MICROWIRE, and TI's synchronous serial interfaces. The SSIs support both SPI master and slave up to 4 MHz.

The UART implements a universal asynchronous receiver and transmitter function. The UART supports flexible baud-rate generation up to a maximum of 3 Mbps.

Timer 0 is a general-purpose timer module (GPTM) that provides two 16-bit timers. The GPTM can be configured to operate as a single 32-bit timer, dual 16-bit timers, or as a PWM module.

Timer 1, Timer 2, and Timer 3 are also GPTMs; each timer is functionally equivalent to Timer 0.

In addition to these four timers, a separate timer in the RF core handles timing for RF protocols; the RF timer can be synchronized to the RTC.

The I2S interface is used to handle digital audio (for more information, see the *CC13x0, CC26x0 SimpleLink™ Wireless MCU Technical Reference Manual*).

The I^2C interface is used to communicate with devices compatible with the I^2C standard. The I^2C interface can handle 100-kHz and 400-kHz operation, and can serve as both I^2C master and I^2C slave.

The TRNG module provides a true, nondeterministic noise source for the purpose of generating keys, initialization vectors (IVs), and other random number requirements. The TRNG is built on 24 ring oscillators that create unpredictable output to feed a complex nonlinear-combinatorial circuit.

The watchdog timer is used to regain control if the system fails due to a software error after an external device fails to respond as expected. The watchdog timer can generate an interrupt or a reset when a predefined time-out value is reached.

The device includes a direct memory access (µDMA) controller. The µDMA controller provides a way to offload data-transfer tasks from the CM3 CPU, thus allowing for more efficient use of the processor and the available bus bandwidth. The µDMA controller can perform transfer between memory and peripherals. The µDMA controller has dedicated channels for each supported on-chip module and can be programmed to automatically perform transfers between peripherals and memory when the peripheral is ready to transfer more data.

Some features of the µDMA controller follow (this is not an exhaustive list):

- Highly flexible and configurable channel operation of up to 32 channels
- Transfer modes: memory-to-memory, memory-to-peripheral, peripheral-to-memory, and peripheral-to-peripheral
- Data sizes of 8, 16, and 32 bits

The AON domain contains circuitry that is always enabled, except when in shutdown mode (where the digital supply is off). This circuitry includes the following:

- The RTC can be used to wake the device from any state where it is active. The RTC contains three compare registers and one capture register. With software support, the RTC can be used for clock and calendar operation. The RTC is clocked from the 32-kHz RC oscillator or crystal. The RTC can also be compensated to tick at the correct frequency even when the internal 32-kHz RC oscillator is used instead of a crystal.
- The battery monitor and temperature sensor are accessible by software and provide a battery status indication as well as a coarse temperature measure.

6.10 Voltage Supply Domains

The CC1310 device can interface to two or three different voltage domains depending on the package type. On-chip level converters ensure correct operation as long as the signal voltage on each input/output pin is set with respect to the corresponding supply pin (VDDS, VDDS2, or VDDS3). Table 6-3 lists the pin-to-VDDS mapping.

Table 6-3. Pin Function to VDDS Mapping Table

	Package		
	VQFN 7 × 7 (RGZ)	**VQFN 5 × 5 (RHB)**	**VQFN 4 × 4 (RSM)**
VDDS(1)	DIO 23–30 Reset_N	DIO 7–14 Reset_N	DIO 5–9 Reset_N
VDDS2	DIO 1–11	DIO 0–6 JTAG_TCKC JTAG_TMSC	DIO 0–4 JTAG_TCKC JTAG_TMSC
VDDS3	DIO 12–22 JTAG_TCKC JTAG_TMSC	NA	NA

(1) The VDDS_DCDC pin must always be connected to the same voltage as the VDDS pin.

6.11 System Architecture

Depending on the product configuration, the CC1310 device can function as a wireless network processor (WNP – a device running the wireless protocol stack, with the application running on a separate host MCU), or as a system-on-chip (SoC) with the application and protocol stack running on the ARM CM3 core inside the device.

In the first case, the external host MCU communicates with the device using SPI or UART. In the second case, the application must be written according to the application framework supplied with the wireless protocol stack.

7 Application, Implementation, and Layout

NOTE

Information in the following Applications section is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

7.1 Application Information

Few external components are required for the operation of the CC1310 device. Figure 7-1 shows a typical application circuit.

The board layout greatly influences the RF performance of the CC1310 device.

On the Texas Instruments CC1310EM-7XD-7793 reference design, the optimal differential impedance seen from the RF pins into the balun and filter and antenna is 44 + j15.

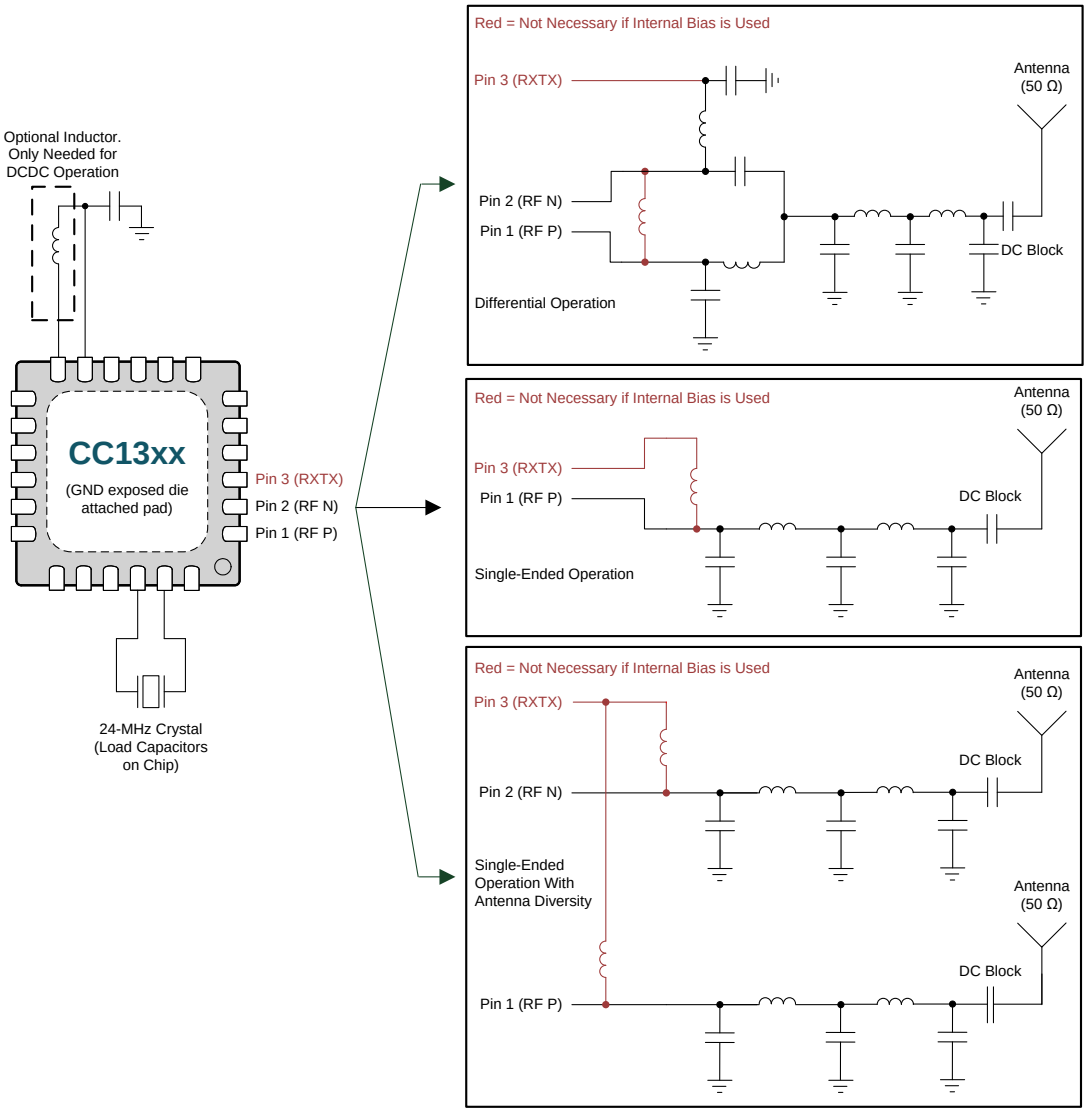


Figure 7-1 does not show decoupling capacitors for power pins. For a complete reference design, see the product folder on www.ti.com.

Figure 7-1. CC1310 Application Circuits

7.2 TI Design or Reference Design

The TI Designs Reference Design Library is a robust reference design library spanning analog, embedded processor, and connectivity. Created by TI experts to help you jumpstart your system design, all TI Designs include schematic or block diagrams, BOMs, and design files to speed your time to market.

Humidity and Temperature Sensor Node for Sub-1 GHz Star Networks Enabling 10+ Year Coin Cell Battery Life

This reference design uses TI's nano-power system timer, boost converter, SimpleLink™ ultra-low-power Sub-1GHzwireless MCU platform, and humidity-sensing technologies to demonstrate an ultra-low-power method to duty-cycle sensor end nodes leading to extremely long battery life. The TI Design includes techniques for system design, detailed test results, and information to get the design operating running quickly.

SimpleLink™ Sub-1 GHz Sensor to Cloud Gateway Reference Design for TI-RTOS Systems

This reference design demonstrates how to connect sensors to the cloud over a long-range Sub-1 GHz wireless network, suitable for industrial settings such as building control and asset tracking. The solution is based on a TI-RTOS gateway. This design provides a complete end-to-end solution for creating a Sub-1 GHz sensor network with an Internet of Things (IoT) gateway solution and cloud connectivity. The gateway solution is based on the low-power, SimpleLink™ Wi-Fi® CC3220 wireless microcontroller (MCU), which hosts the gateway application and the SimpleLink Sub-1 GHz CC1310/CC1312R or the multi-band CC1350/ CC1352R wireless MCU as the MAC Co-Processor. The reference design also includes sensor node example applications running on the SimpleLink Sub-1 GHz CC1312R/CC1310 and multi-band CC1352R/CC1350 wireless MCUs.

Low-Power Wireless M-Bus Communications Module Reference Design

This reference design explains how to use the TI wireless M-Bus stack for CC1310 and CC1350 wireless MCUs and integrate it into a smart meter or data-collector product. This software stack is compatible with the Open Metering System (OMS) v3.0.1 specification. This design offers ready-to-use binary images for any of the wireless M-Bus S-, T-, or C-modes at 868 MHz with unidirectional (meter) or bidirectional configurations (both meter and data collector).

Low-Power Water Flow Measurement With Inductive Sensing Reference Design

This reference design demonstrates a highly-integrated solution for this application using an inductive sensing technique enabled by the CC1310/CC1350 SimpleLink™ Wireless MCU and FemtoFET™ MOSFET. This reference design also provides the platform for integration of wireless communications such as wireless M-Bus, Sigfox™, or a proprietary protocol.

Heat Cost Allocator with wM-Bus at 868 MHz Reference Design

This reference design implements a heat cost allocator system following the EN834 standard with the 'two-sensor measurement method'. The solution achieves better than 0.5 degrees Celsius accuracy across a range of +20 to +85°C. Two analog temperature sensors are available as matched pairs to eliminate the need for calibration during manufacturing and lowering OEM system cost. The CC1310 wireless MCU provides a single-chip solution for heat measurement (control of the two temperature sensors) and RF communications (example code using 868 MHz wM-Bus S, T and C-modes "Meter" device).

Sub-1 GHz Sensor to Cloud Industrial IoT Gateway Reference Design for Linux Systems

This reference design demonstrates how to connect sensors to the cloud over a long-range Sub-1 GHz wireless network, suitable for industrial settings such as building control and asset tracking. This design provides a complete end-to-end solution for creating a Sub-1 GHz sensor network with an Internet of Things (IoT) gateway solution and cloud connectivity. The gateway solution is based on the low-power, SimpleLink™ Wi-Fi® CC3220 wireless microcontroller (MCU), which hosts the gateway application and the SimpleLink Sub-1 GHz CC1312R/CC1310 or the multi-band CC1352R/CC1350 wireless MCU as the MAC Co-Processor.

8 Device and Documentation Support

TI offers an extensive line of development tools. Tools and software to evaluate the performance of the device, generate code, and develop solutions are listed in the following.

8.1 Device Nomenclature

To designate the stages in the product development cycle, TI assigns prefixes to all part numbers and/or date-code. Each device has one of three prefixes/identifications: X, P, or null (no prefix) (for example, CC1310 is in production; therefore, no prefix/identification is assigned).

Device development evolutionary flow:

X Experimental device that is not necessarily representative of the final device's electrical specifications and may not use production assembly flow.

P Prototype device that is not necessarily the final silicon die and may not necessarily meet final electrical specifications.

null Production version of the silicon die that is fully qualified.

Production devices have been characterized fully, and the quality and reliability of the device have been demonstrated fully. TI's standard warranty applies.

Predictions show that prototype devices (X or P) have a greater failure rate than the standard production devices. Texas Instruments recommends that these devices not be used in any production system because their expected end-use failure rate still is undefined. Only qualified production devices are to be used.

TI device nomenclature also includes a suffix with the device family name. This suffix indicates the package type (for example, *RGZ*).

For orderable part numbers of *CC1310* devices in the *RSM (4-mm × 4-mm), RHB (5-mm × 5-mm), or RGZ (7-mm × 7-mm)* package types, see the *Package Option Addendum* of this document, the TI website (www.ti.com), or contact your TI sales representative.

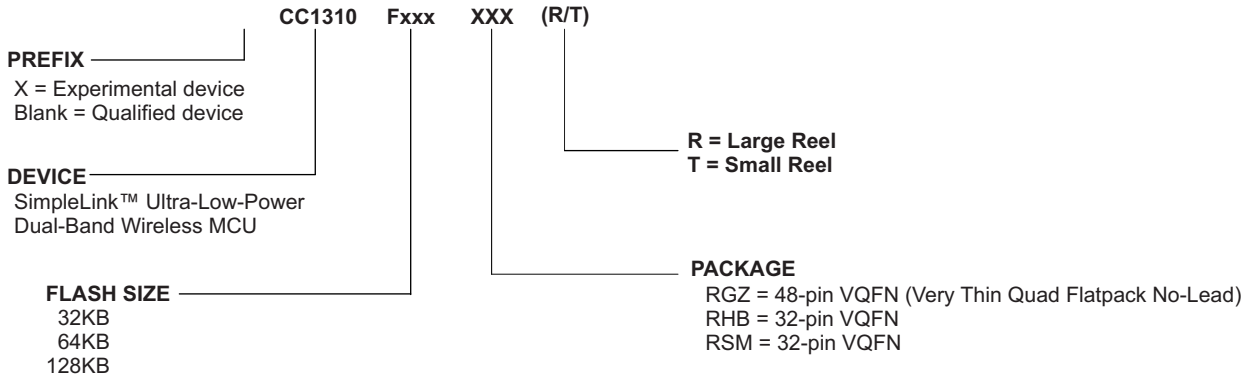


Figure 8-1. Device Nomenclature

8.2 Tools and Software

Development Kit:

SimpleLink™ Sub-1 GHz CC1310 Wireless MCU LaunchPad™ Development Kit

The SimpleLink™ Sub-1 GHz CC1310 wireless microcontroller (MCU) LaunchPad™ development kit is the first LaunchPad kit with a Sub-1 GHz radio, which offers long-range connectivity, combined with a 32-bit Arm® Cortex®-M3 processor on a single chip.

The CC1310 device is a wireless MCU targeting low-power, long-range wireless applications. The CC1310 wireless MCU contains a 32-bit Arm Cortex-M3 processor that runs at 48 MHz as the main processor and a rich peripheral feature set that includes a unique ultra-low power sensor controller. This sensor controller is great for interfacing external sensors and for collecting analog and digital data autonomously while the rest of the system is in sleep mode.

Software:

SimpleLink™ CC13x0 SDK

The SimpleLink™ Sub-1 GHz CC13x0 software development kit (SDK) provides a comprehensive Sub-1 GHz software package for the Sub-1 GHz CC1310 and Dual-band CC1350 wireless MCUs and includes the following:

- TI 15.4-Stack - IEEE 802.15.4e/g-based star topology networking solution for Sub-1 GHz ISM bands (433 MHz, 868 MHz and 915 MHz).
- Support for proprietary solutions - proprietary RF examples for Sub-1 GHz based on the RF driver and EasyLink Abstraction Layer.
- Bluetooth Low Energy – Stack including support for all Bluetooth core specification 4.2 features as well as a BLE micro-stack to support customers using the Dual-Band CC1350 wireless MCU.

The SimpleLink CC13x0 SDK is part of the TI SimpleLink MCU platform, offering a single development environment that delivers flexible hardware, software and tool options for customers developing wired and wireless applications. For more information about the SimpleLink MCU Platform, visit www.ti.com/simplelink.

Software Tools:

SmartRF™ Studio 7

SmartRF™ Studio is a PC application that helps designers of radio systems to easily evaluate the RF-IC at an early stage in the design process.

- Test functions for transmitting and receiving radio packets, continuous wave transmit and receive
- Evaluate RF performance on custom boards by wiring it to a supported evaluation board or debugger
- Can also be used without any hardware, but then only to generate, edit and export radio configuration settings
- Can be used in combination with several development kits for Texas Instruments' CC1310 RF-ICs

Sensor Controller Studio

Sensor Controller Studio provides a development environment for the CC1310 Sensor Controller. The Sensor Controller is a proprietary, power-optimized CPU inside the CC1310, which can perform simple background tasks autonomously and independent of the System CPU state.

- Allows for Sensor Controller task algorithms to be implemented using a C-like programming language
- Outputs a Sensor Controller Interface driver, which incorporates the generated Sensor Controller machine code and associated definitions
- Allows for rapid development by using the integrated Sensor Controller task testing and debugging functionality. This allows for live visualization of sensor data and algorithm verification.

IDEs and Compilers:

Code Composer Studio™ IDE

- An integrated development environment (IDE) with project management tools and editor
- Code Composer Studio (CCS) 6.1 and later has built-in support for the CC1310 device family
- Best support for XDS debuggers; XDS100v3, XDS110 and XDS200
- High integration with TI-RTOS with support for TI-RTOS Object View

Code Composer Studio™ Cloud IDE

Code Composer Studio™ (CCS) Cloud is a web-based IDE that allows you to create, edit, and build CCS and Energia projects. After you have successfully built your project, you can download and run on your connected LaunchPad™ development kit. Basic debugging, including features like setting breakpoints and viewing variable values is now supported with CCS Cloud.

CCS UniFlash

CCS UniFlash is a standalone tool used to program on-chip flash memory on TI MCUs. UniFlash has a GUI, command line, and scripting interface. CCS UniFlash is available free of charge.

IAR Embedded Workbench® for Arm

- Integrated development environment with project management tools and editor
- IAR EWARM 7.30.3 and later has built-in support for the CC1310 device family
- Broad debugger support, supporting XDS100v3, XDS200, IAR I-jet® and SEGGER J-Link™
- Integrated development environment with project management tools and editor
- RTOS plugin available for TI-RTOS

For a complete listing of development-support tools for the CC1310 platform, visit the Texas Instruments website at www.ti.com. For information on pricing and availability, contact the nearest TI field sales office or authorized distributor.

8.3 Documentation Support

To receive notification of documentation updates, navigate to the device product folder on ti.com (**CC1310**). In the upper right corner, click on Alert me to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

The current documentation that describes the CC1310, related peripherals, and other technical collateral is listed in the following.

Errata

CC1310 SimpleLink™ Ultra-Low-Power Sub-1 GHz Wireless MCU Silicon Revisions B, A Silicon Errata

Technical Reference Manual

CC13xx, CC26xx SimpleLink™ Wireless MCU Technical Reference Manual

Reference Guide

CC26xx/CC13xx Power Management Software Developer's Reference Guide

8.4 Texas Instruments Low-Power RF Website

TI's Low-Power RF website has all the latest products, application and design notes, FAQ section, news and events updates. Go to www.ti.com/longrange.

8.5 Additional Information

Texas Instruments offers a wide selection of cost-effective, low-power RF solutions for proprietary and standard-based wireless applications for use in industrial and consumer applications. The selection includes RF transceivers, RF transmitters, RF front ends, and Systems-on-Chips as well as various software solutions for the Sub-1 GHz and 2.4-GHz frequency bands.

In addition, Texas Instruments provides a large selection of support collateral such as development tools, technical documentation, reference designs, application expertise, customer support, third-party and university programs.

Other than providing technical support forums, videos, and blogs, the Low-Power RF E2E Online Community also presents the opportunity to interact with engineers from all over the world.

With a broad selection of product solutions, end-application possibilities, and a range of technical support, Texas Instruments offers the broadest low-power RF portfolio.

8.6 Community Resources

The following links connect to TI community resources. Linked contents are provided "AS IS" by the respective contributors. They do not constitute TI specifications and do not necessarily reflect TI's views; see TI's Terms of Use.

TI E2E™ Online Community The TI engineer-to-engineer (E2E) community was created to foster collaboration among engineers. At e2e.ti.com, you can ask questions, share knowledge, explore ideas and help solve problems with fellow engineers.

TI Embedded Processors Wiki ***Texas Instruments Embedded Processors Wiki.*** Established to help developers get started with Embedded Processors from Texas Instruments and to foster innovation and growth of general knowledge about the hardware and software surrounding these devices.

8.7 Trademarks

SimpleLink, SmartRF, Code Composer Studio, Texas Instruments, FemtoFET, E2E are trademarks of Texas Instruments.
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ULPBench is a trademark of Embedded Microprocessor Benchmark Consortium.
CoreMark is a registered trademark of Embedded Microprocessor Benchmark Consortium.
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IEEE Std 1241 is a trademark of Institute of Electrical and Electronics Engineers, Incorporated.
IEEE is a registered trademark of Institute of Electrical and Electronics Engineers, Incorporated.
J-Link is a trademark of SEGGER Microcontroller GmbH.
Wi-Fi is a registered trademark of Wi-Fi Alliance.
Wi-SUN is a trademark of Wi-SUN Alliance, Inc.
Zigbee is a registered trademark of Zigbee Alliance.
All other trademarks are the property of their respective owners.

8.8 Electrostatic Discharge Caution

This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

8.9 Export Control Notice

Recipient agrees to not knowingly export or re-export, directly or indirectly, any product or technical data (as defined by the U.S., EU, and other Export Administration Regulations) including software, or any controlled product restricted by other applicable national regulations, received from disclosing party under nondisclosure obligations (if any), or any direct product of such technology, to any destination to which such export or re-export is restricted or prohibited by U.S. or other applicable laws, without obtaining prior authorization from U.S. Department of Commerce and other competent Government authorities to the extent required by those laws.

8.10 Glossary

TI Glossary This glossary lists and explains terms, acronyms, and definitions.

9 Mechanical, Packaging, and Orderable Information

9.1 Packaging Information

The following pages include mechanical packaging and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead/Ball Finish (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
CC1310F128RGZR	ACTIVE	VQFN	RGZ	48	2500	Green (RoHS & no Sb/Br)	CU NIPDAU | CU NIPDAUAG	Level-3-260C-168 HR	-40 to 85	CC1310 F128	Samples
CC1310F128RGZT	ACTIVE	VQFN	RGZ	48	250	Green (RoHS & no Sb/Br)	CU NIPDAU | CU NIPDAUAG	Level-3-260C-168 HR	-40 to 85	CC1310 F128	Samples
CC1310F128RHBR	ACTIVE	VQFN	RHB	32	3000	Green (RoHS & no Sb/Br)	CU NIPDAU | CU NIPDAUAG	Level-3-260C-168 HR	-40 to 85	CC1310 F128	Samples
CC1310F128RHBT	ACTIVE	VQFN	RHB	32	250	Green (RoHS & no Sb/Br)	CU NIPDAU | CU NIPDAUAG	Level-3-260C-168 HR	-40 to 85	CC1310 F128	Samples
CC1310F128RSMR	ACTIVE	VQFN	RSM	32	3000	Green (RoHS & no Sb/Br)	CU NIPDAU | CU NIPDAUAG	Level-3-260C-168 HR	-40 to 85	CC1310 F128	Samples
CC1310F128RSMT	ACTIVE	VQFN	RSM	32	250	Green (RoHS & no Sb/Br)	CU NIPDAU | CU NIPDAUAG	Level-3-260C-168 HR	-40 to 85	CC1310 F128	Samples
CC1310F32RGZR	ACTIVE	VQFN	RGZ	48	2500	Green (RoHS & no Sb/Br)	CU NIPDAU | CU NIPDAUAG	Level-3-260C-168 HR	-40 to 85	CC1310 F32	Samples
CC1310F32RGZT	ACTIVE	VQFN	RGZ	48	250	Green (RoHS & no Sb/Br)	CU NIPDAU | CU NIPDAUAG	Level-3-260C-168 HR	-40 to 85	CC1310 F32	Samples
CC1310F32RHBR	ACTIVE	VQFN	RHB	32	3000	Green (RoHS & no Sb/Br)	CU NIPDAU | CU NIPDAUAG	Level-3-260C-168 HR	-40 to 85	CC1310 F32	Samples
CC1310F32RHBT	ACTIVE	VQFN	RHB	32	250	Green (RoHS & no Sb/Br)	CU NIPDAU | CU NIPDAUAG	Level-3-260C-168 HR	-40 to 85	CC1310 F32	Samples
CC1310F32RSMR	ACTIVE	VQFN	RSM	32	3000	Green (RoHS & no Sb/Br)	CU NIPDAU | CU NIPDAUAG	Level-3-260C-168 HR	-40 to 85	CC1310 F32	Samples
CC1310F32RSMT	ACTIVE	VQFN	RSM	32	250	Green (RoHS & no Sb/Br)	CU NIPDAU | CU NIPDAUAG	Level-3-260C-168 HR	-40 to 85	CC1310 F32	Samples
CC1310F64RGZR	ACTIVE	VQFN	RGZ	48	2500	Green (RoHS & no Sb/Br)	CU NIPDAU | CU NIPDAUAG	Level-3-260C-168 HR	-40 to 85	CC1310 F64	Samples
CC1310F64RGZT	ACTIVE	VQFN	RGZ	48	250	Green (RoHS & no Sb/Br)	CU NIPDAU | CU NIPDAUAG	Level-3-260C-168 HR	-40 to 85	CC1310 F64	Samples
CC1310F64RHBR	ACTIVE	VQFN	RHB	32	3000	Green (RoHS & no Sb/Br)	CU NIPDAU | CU NIPDAUAG	Level-3-260C-168 HR	-40 to 85	CC1310 F64	Samples
CC1310F64RHBT	ACTIVE	VQFN	RHB	32	250	Green (RoHS & no Sb/Br)	CU NIPDAU | CU NIPDAUAG	Level-3-260C-168 HR	-40 to 85	CC1310 F64	Samples
CC1310F64RSMR	ACTIVE	VQFN	RSM	32	3000	Green (RoHS & no Sb/Br)	CU NIPDAU | CU NIPDAUAG	Level-3-260C-168 HR	-40 to 85	CC1310 F64	Samples

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead/Ball Finish (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
CC1310F64RSMT	ACTIVE	VQFN	RSM	32	250	Green (RoHS & no Sb/Br)	CU NIPDAU | CU NIPDAUAG	Level-3-260C-168 HR	-40 to 85	CC1310 F64	Samples

(1) The marketing status values are defined as follows:
ACTIVE: Product device recommended for new designs.
LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.
NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.
PREVIEW: Device has been announced but is not in production. Samples may or may not be available.
OBSOLETE: TI has discontinued the production of the device.

(2) **RoHS:** TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".
RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.
Green: TI defines "Green" to mean the content of Chlorine (Cl) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

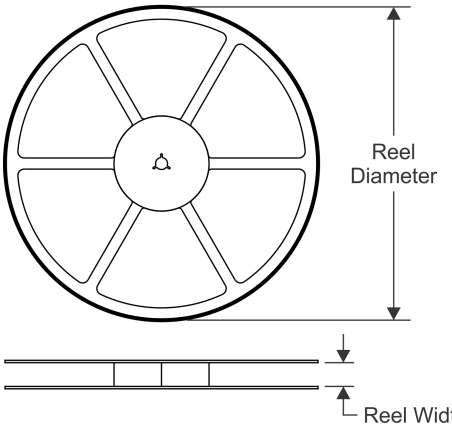
(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

(6) Lead/Ball Finish - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead/Ball Finish values may wrap to two lines if the finish value exceeds the maximum column width.

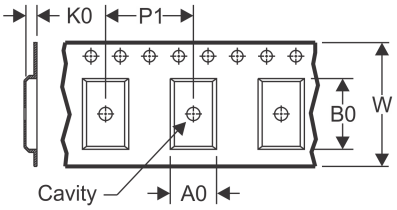
PACKAGE MATERIALS INFORMATION

TAPE AND REEL INFORMATION

REEL DIMENSIONS

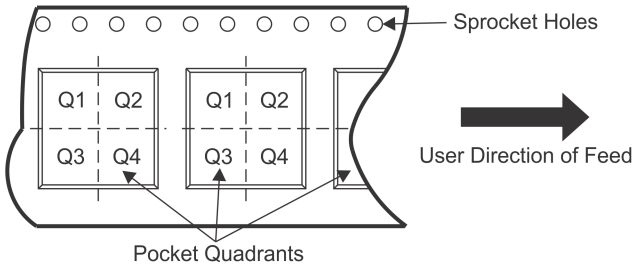


TAPE DIMENSIONS



A0	Dimension designed to accommodate the component width
B0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE

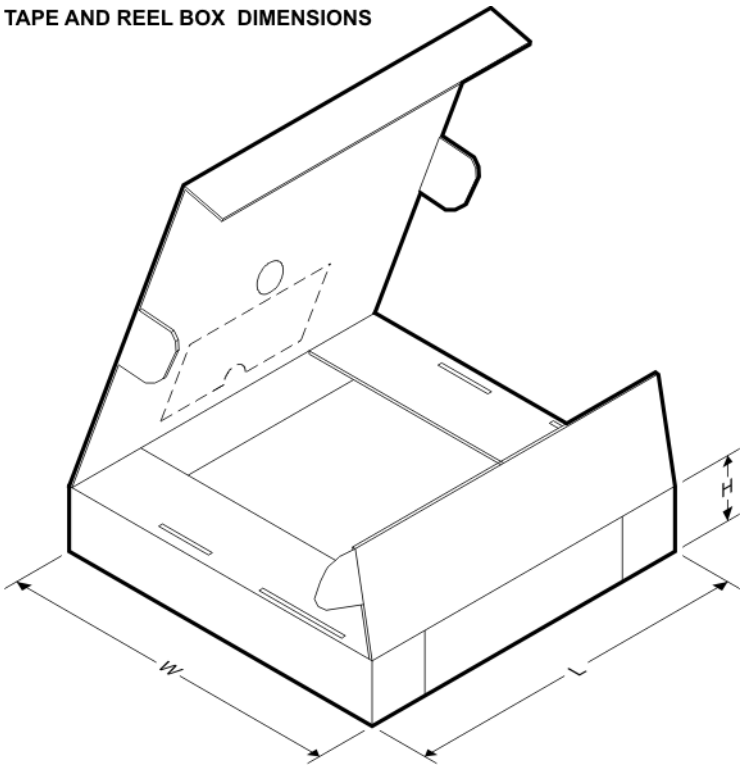


*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
CC1310F128RSMR	VQFN	RSM	32	3000	330.0	12.4	4.25	4.25	1.15	8.0	12.0	Q2
CC1310F64RSMR	VQFN	RSM	32	3000	330.0	12.4	4.25	4.25	1.15	8.0	12.0	Q2

PACKAGE MATERIALS INFORMATION

TAPE AND REEL BOX DIMENSIONS



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
CC1310F128RSMR	VQFN	RSM	32	3000	336.6	336.6	31.8
CC1310F64RSMR	VQFN	RSM	32	3000	336.6	336.6	31.8

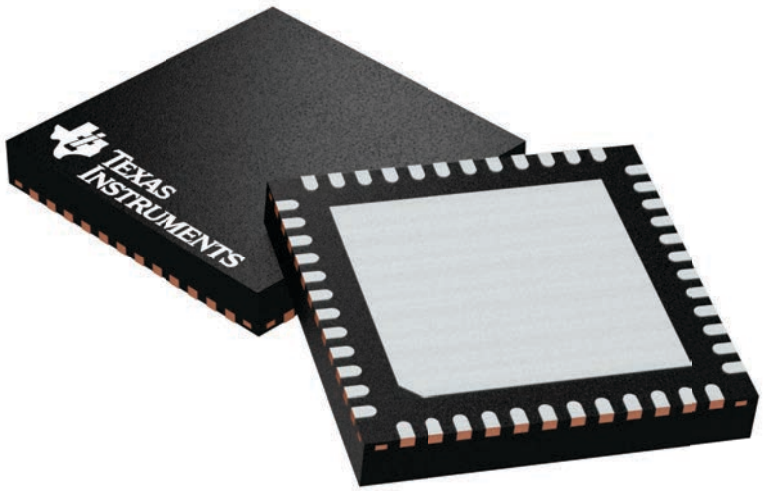
GENERIC PACKAGE VIEW

RGZ 48

VQFN - 1 mm max height

7 x 7, 0.5 mm pitch

PLASTIC QUADFLAT PACK- NO LEAD



Images above are just a representation of the package family, actual package may vary.
Refer to the product data sheet for package details.

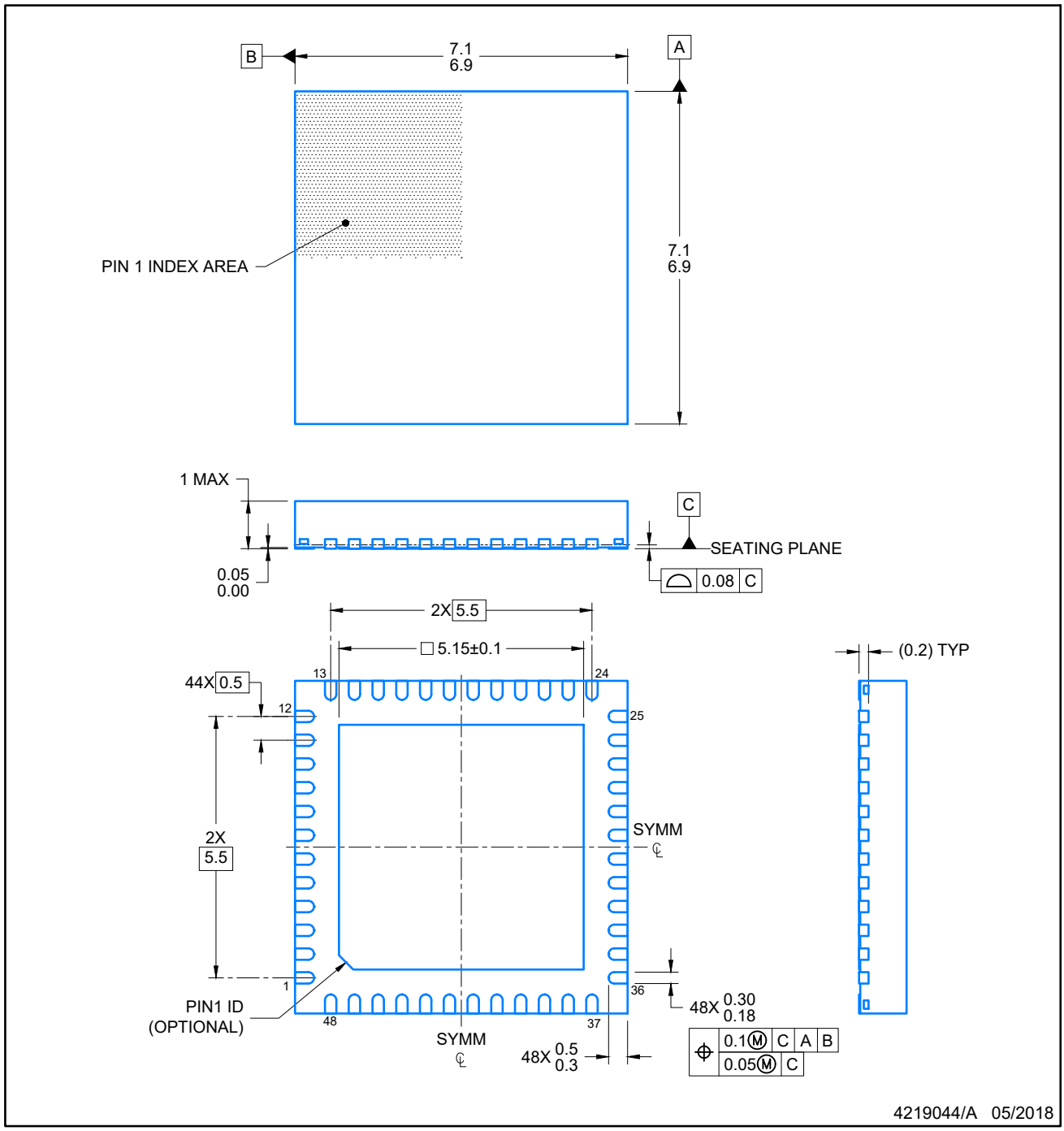
4224671/A

PACKAGE OUTLINE

RGZ0048A

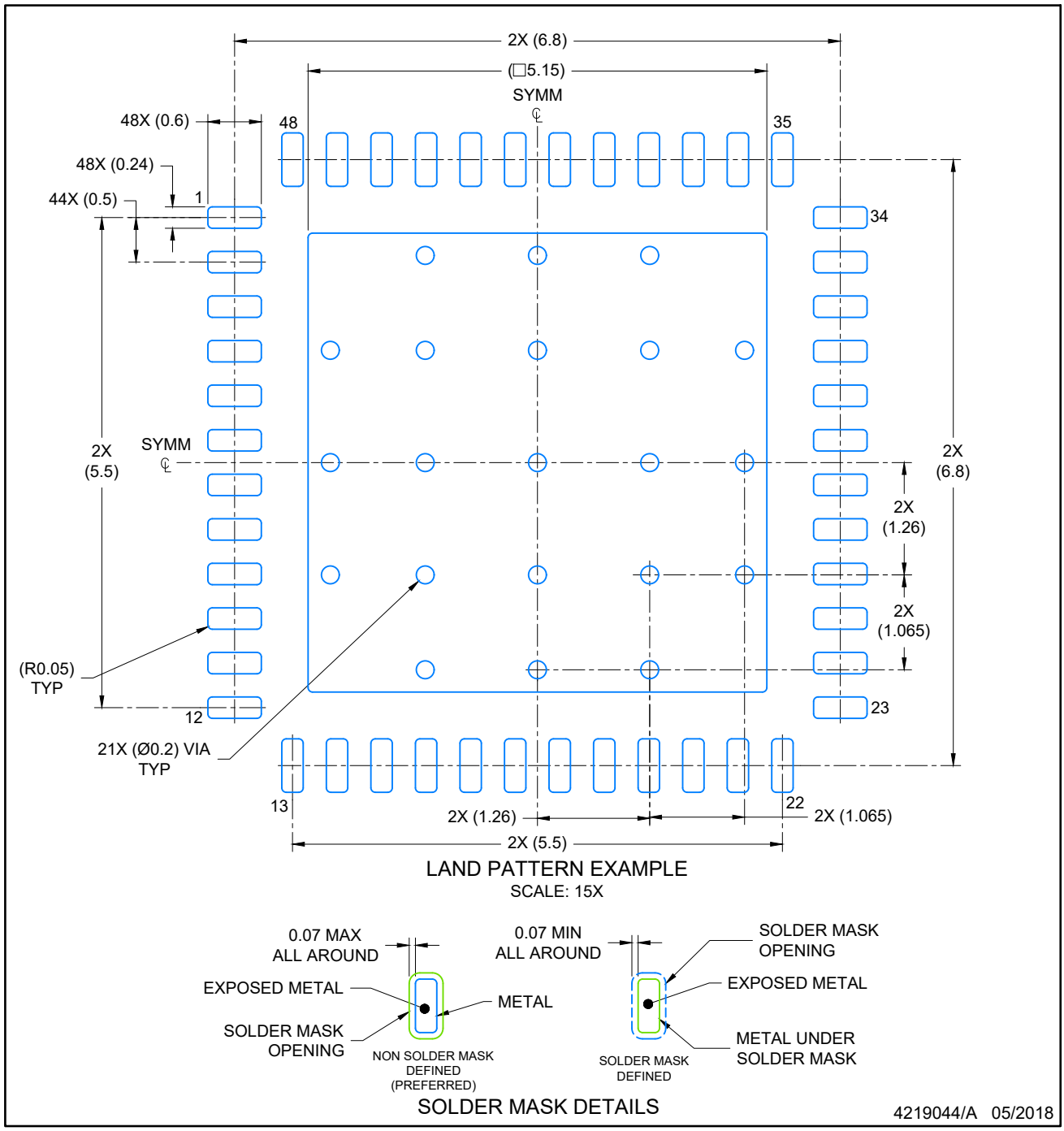
VQFN - 1 mm max height

PLASTIC QUADFLAT PACK- NO LEAD



NOTES:

1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. The package thermal pad must be soldered to the printed circuit board for optimal thermal and mechanical performance.



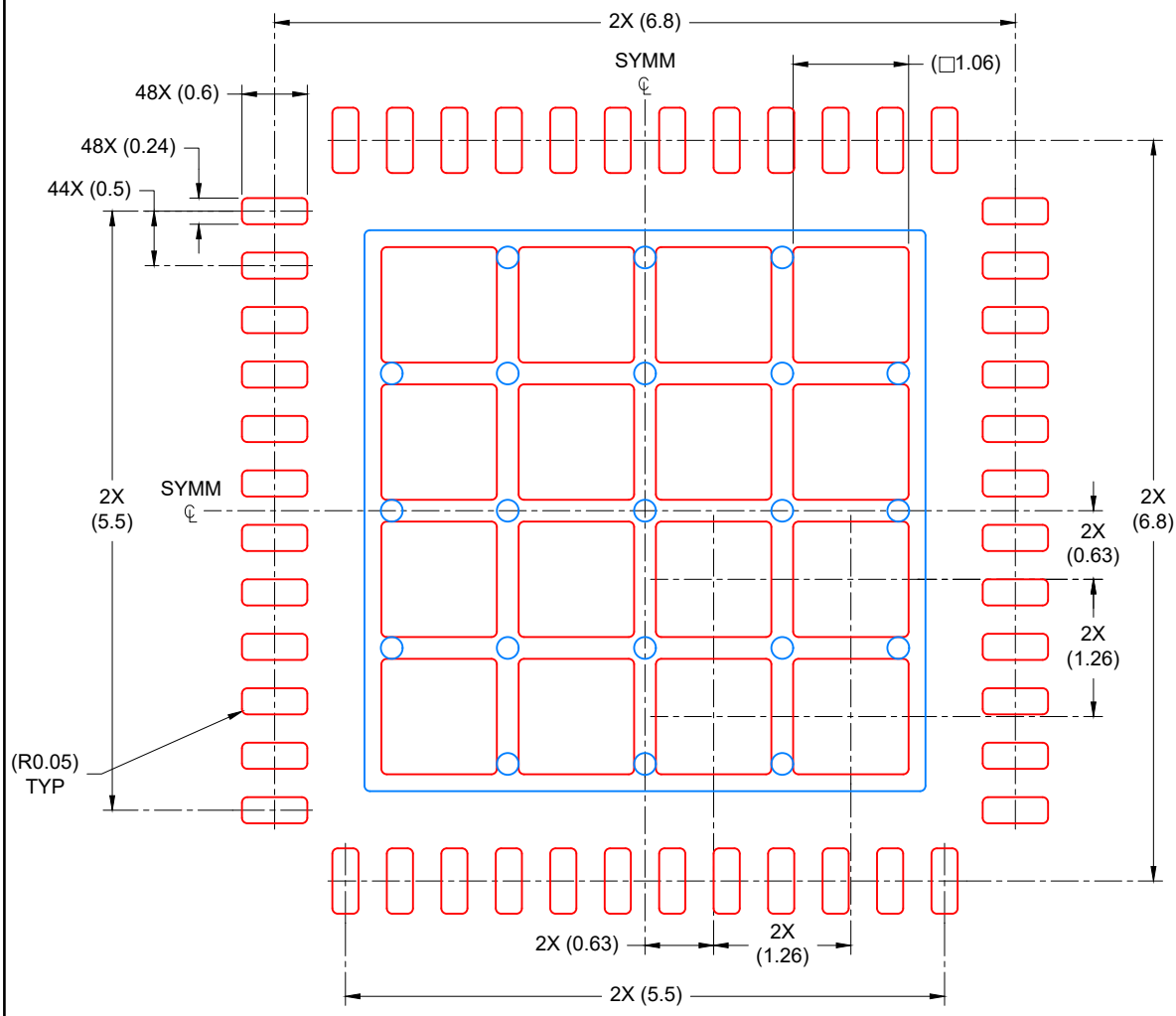
NOTES: (continued)

4. This package is designed to be soldered to a thermal pad on the board. For more information, see Texas Instruments literature number SLUA271 (www.ti.com/lit/slua271).
5. Vias are optional depending on application, refer to device data sheet. If any vias are implemented, refer to their locations shown on this view. It is recommended that vias under paste be filled, plugged or tented.

EXAMPLE STENCIL DESIGN

VQFN - 1 mm max height

PLASTIC QUADFLAT PACK- NO LEAD



SOLDER PASTE EXAMPLE
BASED ON 0.125 mm THICK STENCIL

EXPOSED PAD
67% PRINTED COVERAGE BY AREA
SCALE: 15X

4219044/A 05/2018

NOTES: (continued)

6. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.

GENERIC PACKAGE VIEW

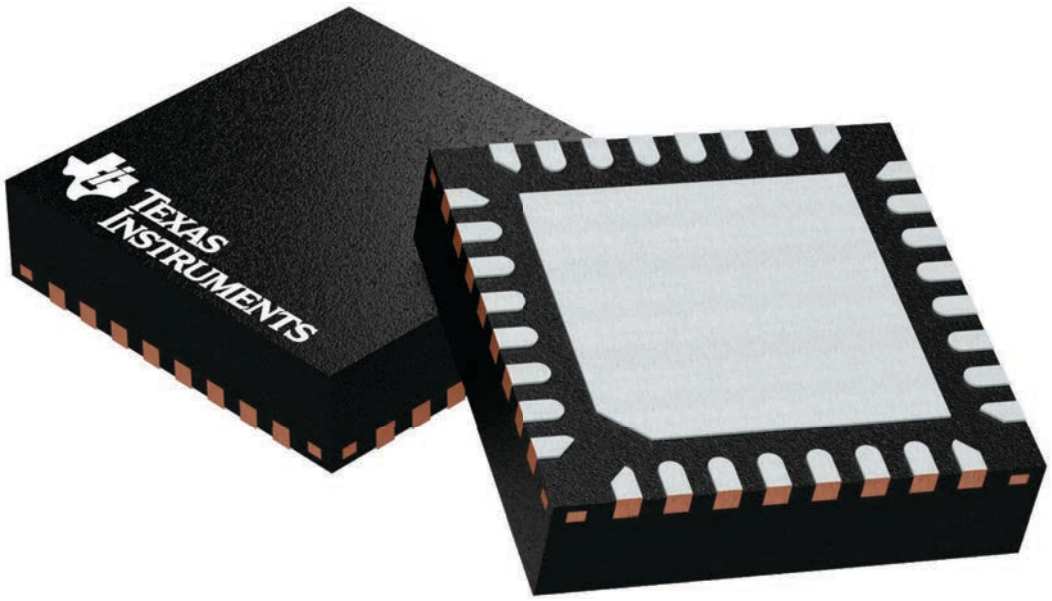
RSM 32

VQFN - 1 mm max height

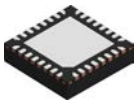
4 x 4, 0.4 mm pitch

PLASTIC QUAD FLATPACK - NO LEAD

This image is a representation of the package family, actual package may vary.
Refer to the product data sheet for package details.



4224982/A

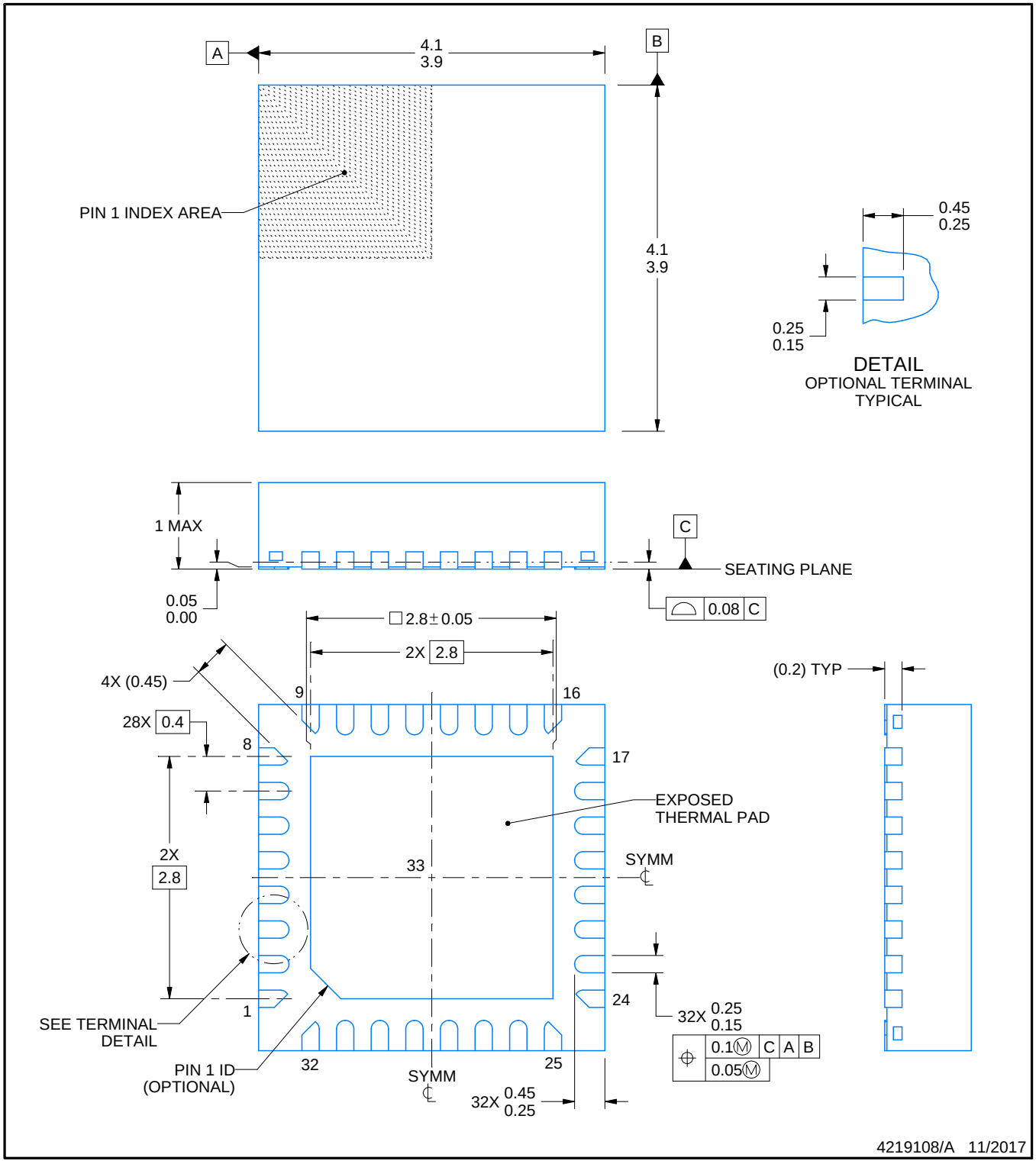


PACKAGE OUTLINE

RSM0032B

VQFN - 1 mm max height

PLASTIC QUAD FLATPACK - NO LEAD



NOTES:

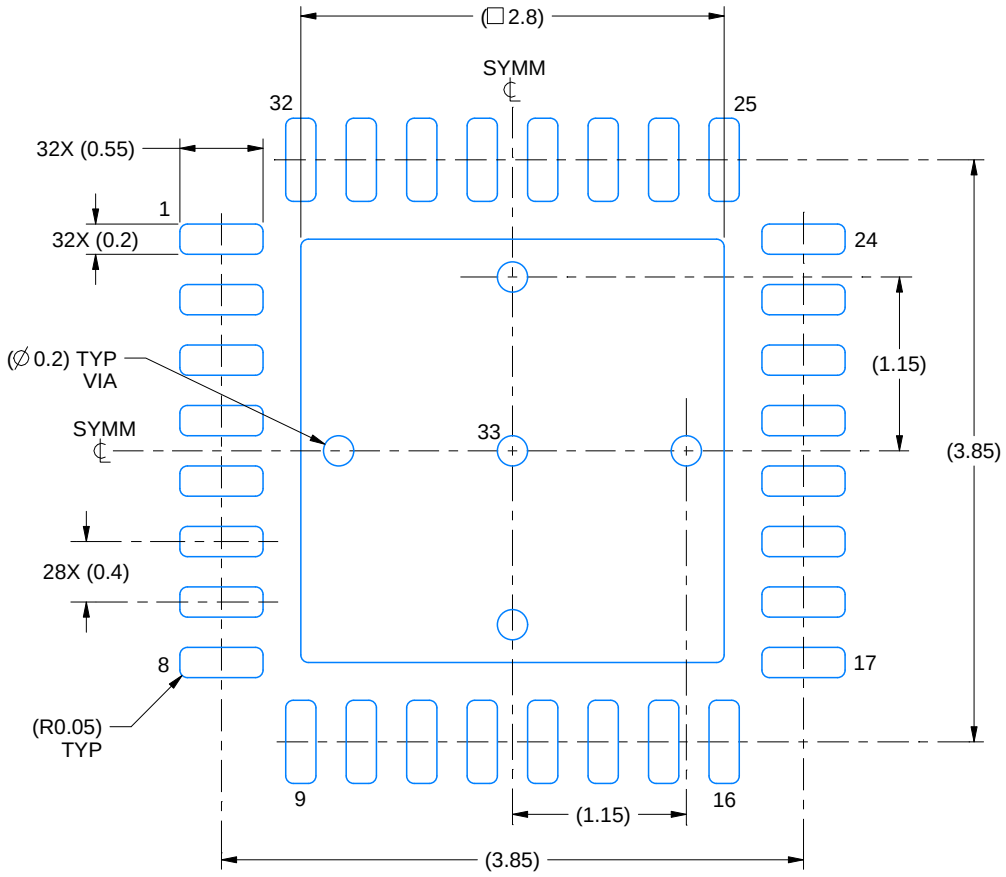
1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. The package thermal pad must be soldered to the printed circuit board for thermal and mechanical performance.

EXAMPLE BOARD LAYOUT

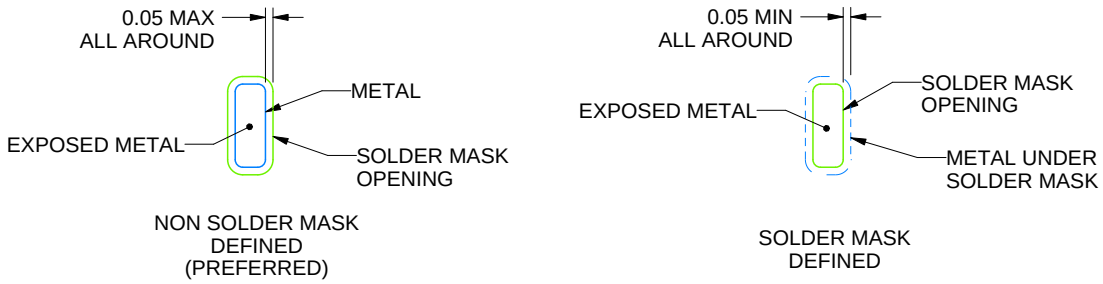
RSM0032B

VQFN - 1 mm max height

PLASTIC QUAD FLATPACK - NO LEAD



LAND PATTERN EXAMPLE
EXPOSED METAL SHOWN
SCALE:20X



SOLDER MASK DETAILS

4219108/A 11/2017

NOTES: (continued)

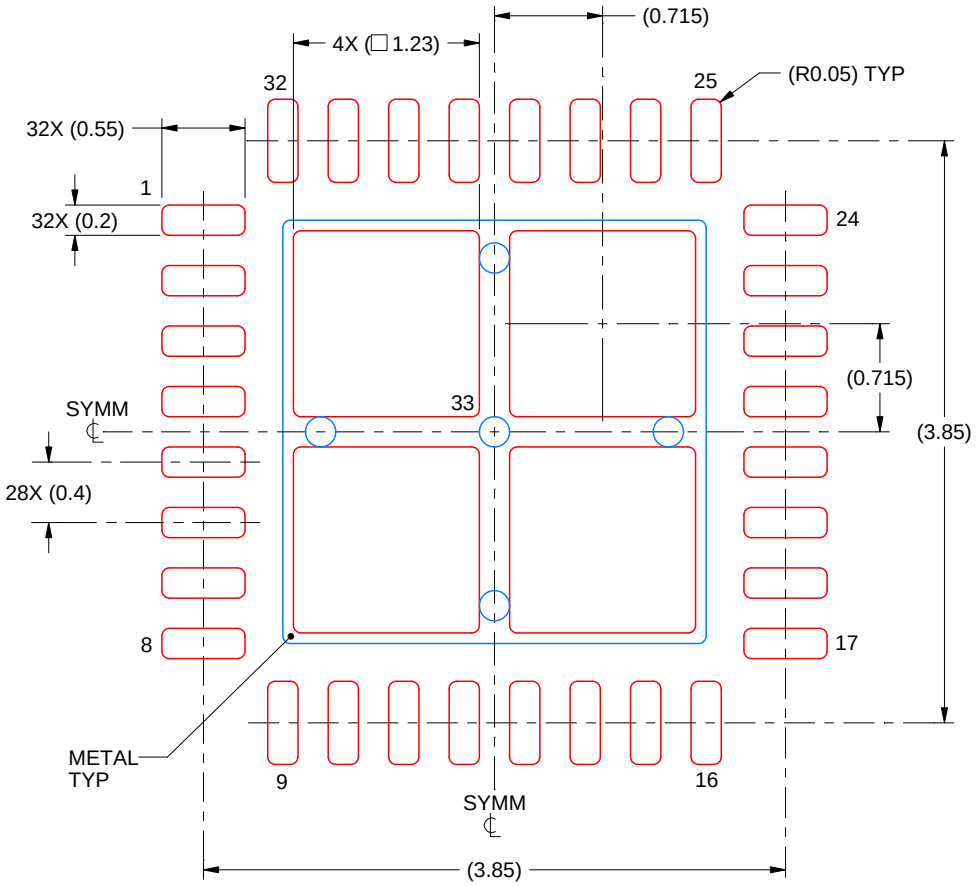
4. This package is designed to be soldered to a thermal pad on the board. For more information, see Texas Instruments literature number SLUA271 (www.ti.com/lit/slua271).
5. Vias are optional depending on application, refer to device data sheet. If any vias are implemented, refer to their locations shown on this view. It is recommended that vias under paste be filled, plugged or tented.

EXAMPLE STENCIL DESIGN

RSM0032B

VQFN - 1 mm max height

PLASTIC QUAD FLATPACK - NO LEAD



SOLDER PASTE EXAMPLE
BASED ON 0.1 mm THICK STENCIL

EXPOSED PAD 33:
77% PRINTED SOLDER COVERAGE BY AREA UNDER PACKAGE
SCALE:20X

4219108/A 11/2017

NOTES: (continued)

6. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.

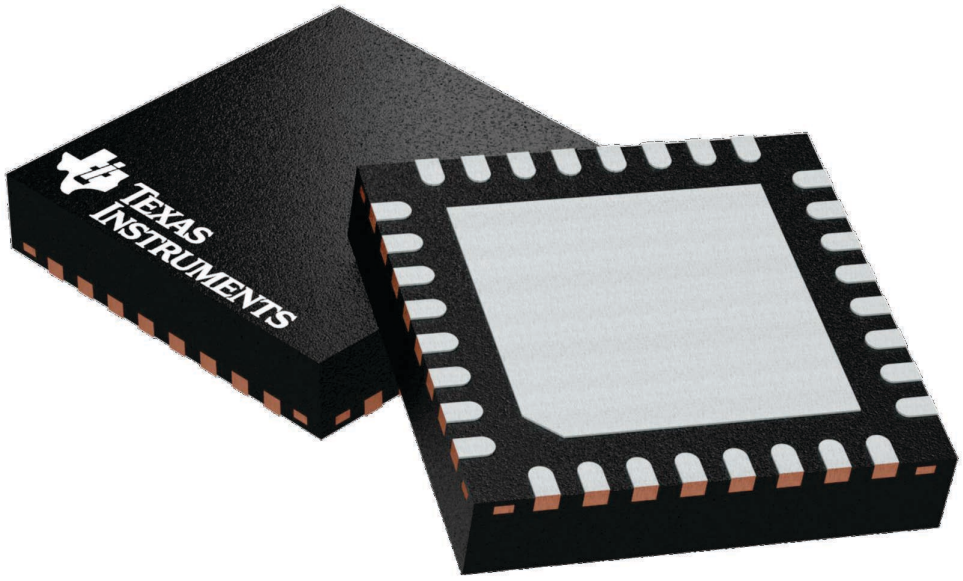
GENERIC PACKAGE VIEW

RHB 32

VQFN - 1 mm max height

5 x 5, 0.5 mm pitch

PLASTIC QUAD FLATPACK - NO LEAD



Images above are just a representation of the package family, actual package may vary.
Refer to the product data sheet for package details.

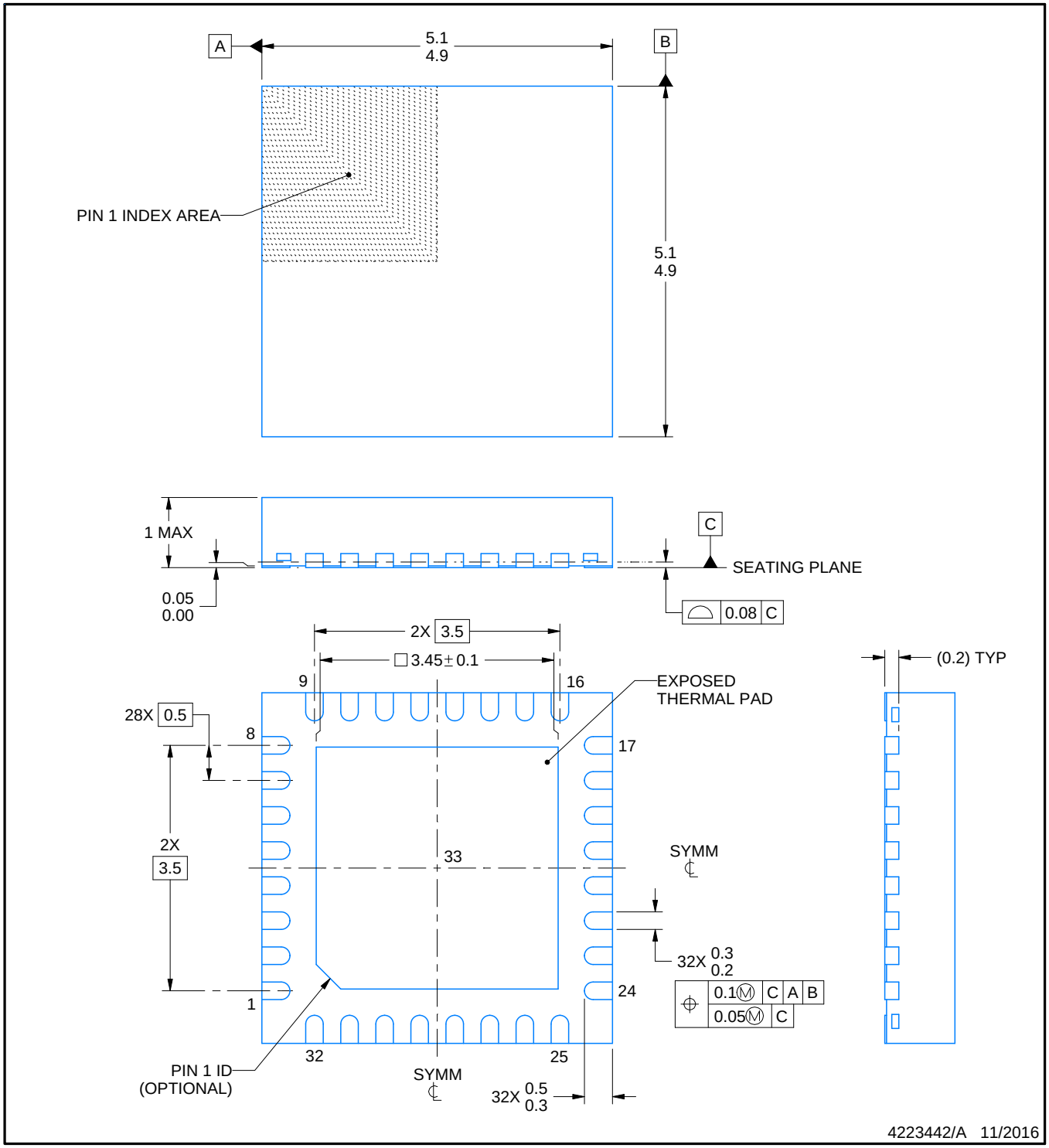
4224745/A

PACKAGE OUTLINE

RHB0032E

VQFN - 1 mm max height

PLASTIC QUAD FLATPACK - NO LEAD



NOTES:

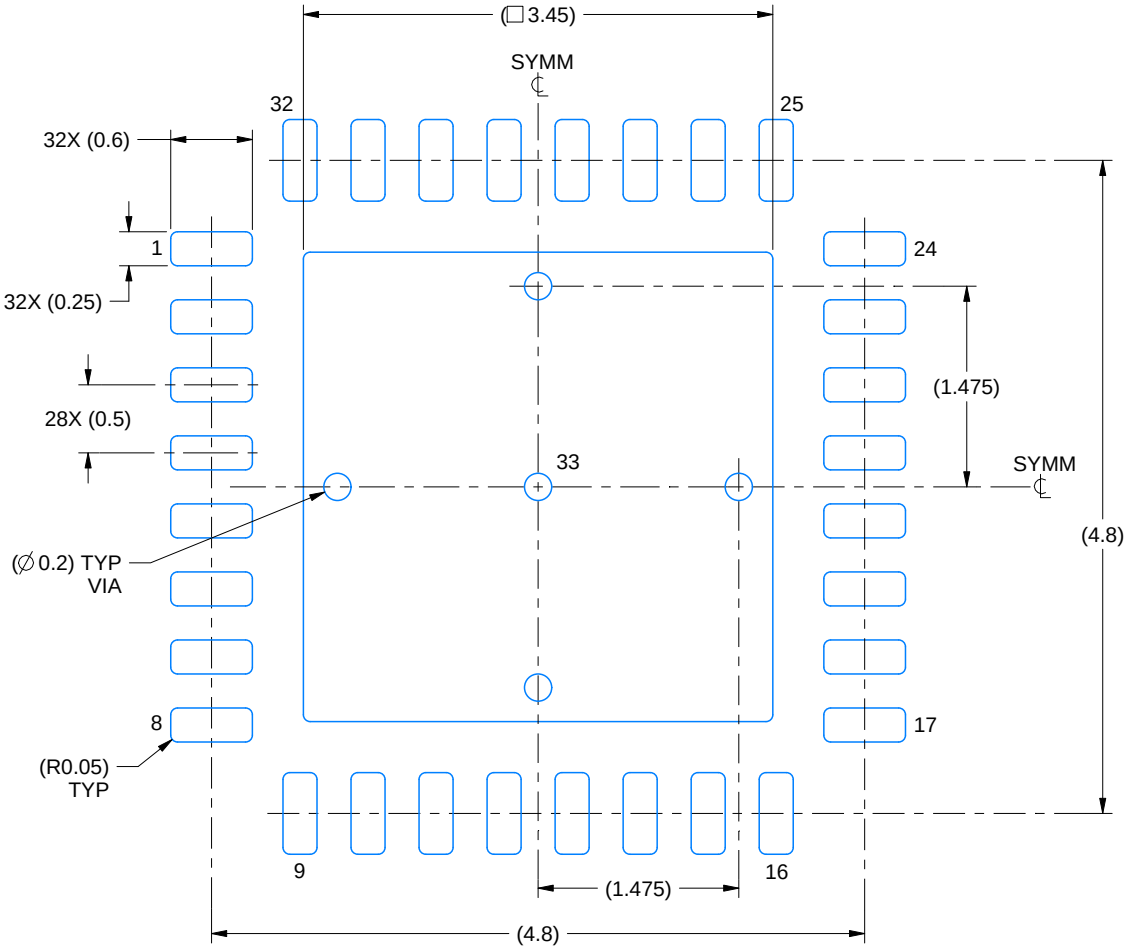
1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. The package thermal pad must be soldered to the printed circuit board for thermal and mechanical performance.

EXAMPLE BOARD LAYOUT

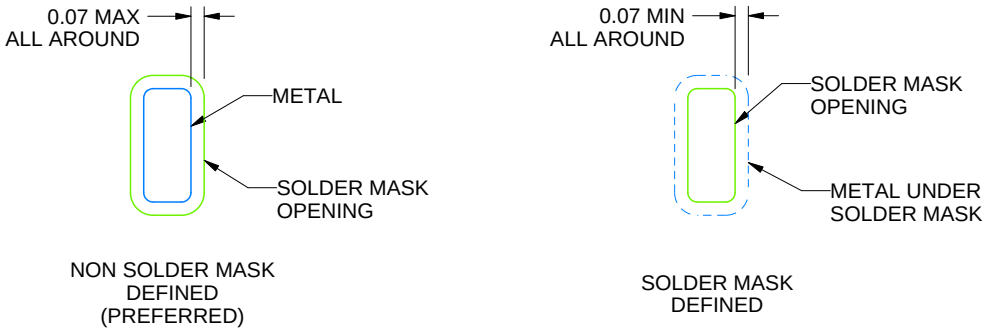
RHB0032E

VQFN - 1 mm max height

PLASTIC QUAD FLATPACK - NO LEAD



LAND PATTERN EXAMPLE
SCALE:18X



SOLDER MASK DETAILS

4223442/A 11/2016

NOTES: (continued)

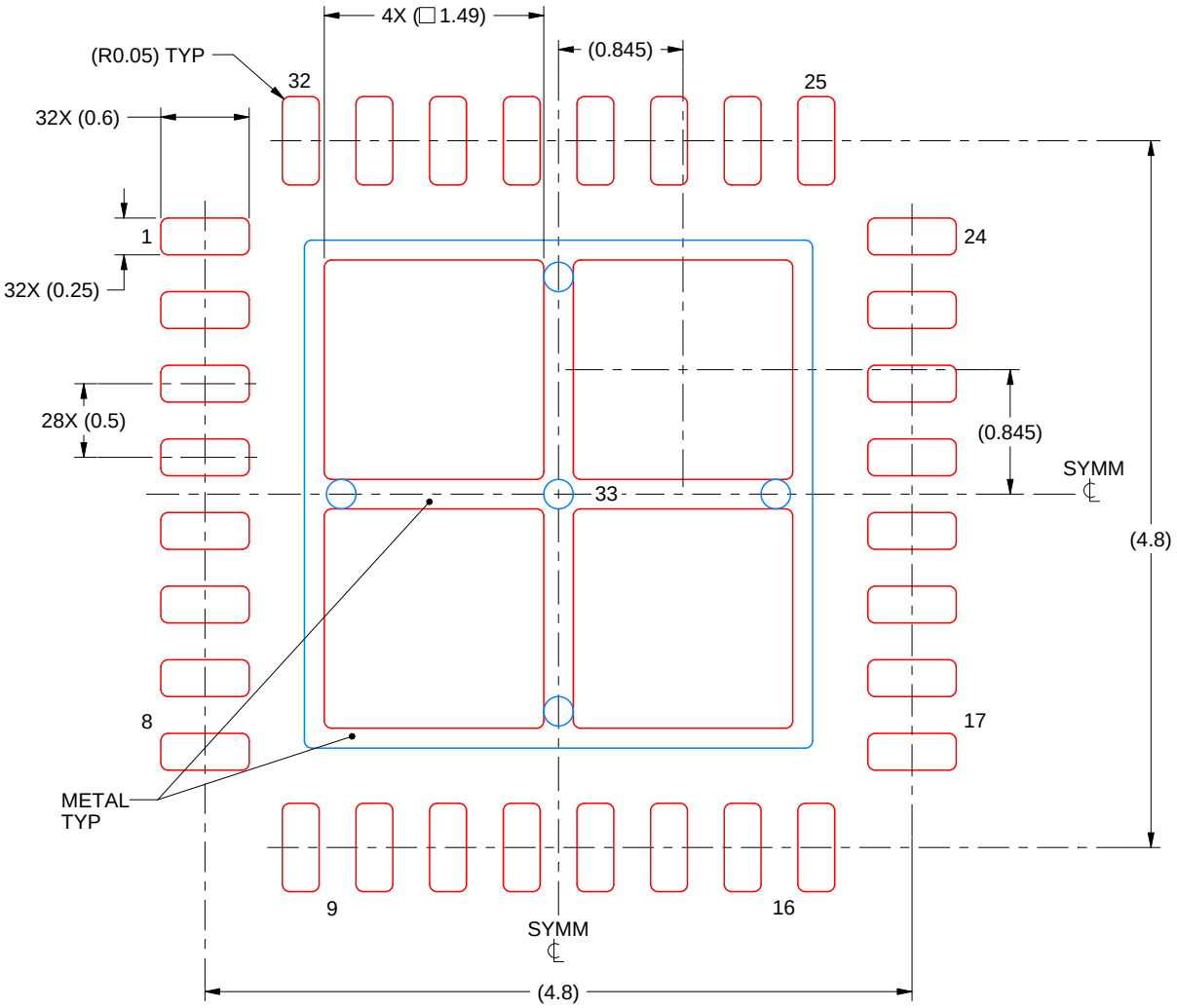
4. This package is designed to be soldered to a thermal pad on the board. For more information, see Texas Instruments literature number SLUA271 (www.ti.com/lit/slua271).
5. Vias are optional depending on application, refer to device data sheet. If any vias are implemented, refer to their locations shown on this view. It is recommended that vias under paste be filled, plugged or tented.

EXAMPLE STENCIL DESIGN

RHB0032E

VQFN - 1 mm max height

PLASTIC QUAD FLATPACK - NO LEAD



SOLDER PASTE EXAMPLE
BASED ON 0.125 mm THICK STENCIL

EXPOSED PAD 33:
75% PRINTED SOLDER COVERAGE BY AREA UNDER PACKAGE
SCALE:20X

4223442/A 11/2016

NOTES: (continued)

6. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.

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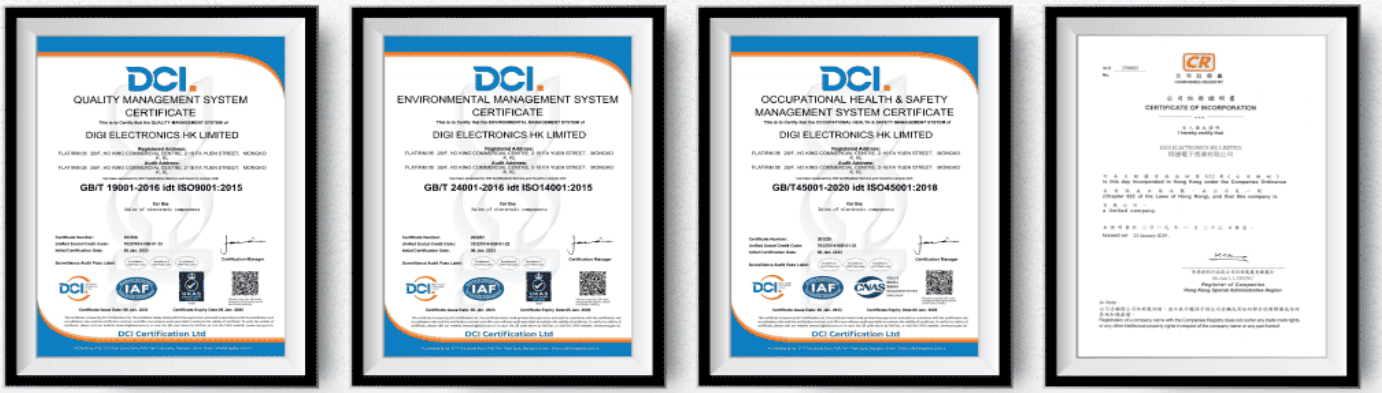

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