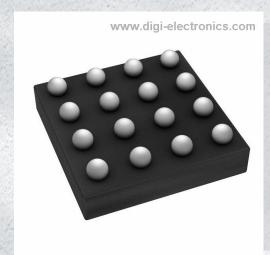


# PAM8902AZER Datasheet



https://www.DiGi-Electronics.com

DiGi Electronics Part Number PAM8902AZER-DG

Manufacturer Diodes Incorporated

Manufacturer Product Number PAM8902AZER

Description IC AMP CLASS D MONO 16CSP

Detailed Description Amplifier IC 1-Channel (Mono) Class D 16-CSP (1.95

x1.95)



Tel: +00 852-30501935

RFQ Email: Info@DiGi-Electronics.com

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8542.33.0001

# **Purchase and inquiry**

Manufacturer Product Number:	Manufacturer:
PAM8902AZER	Diodes Incorporated
Series:	Product Status:
	Obsolete
Type:	Output Type:
Class D	1-Channel (Mono)
Max Output Power x Channels @ Load:	Voltage - Supply:
	2.7V ~ 5.5V
Features:	Mounting Type:
Short-Circuit and Thermal Protection	Surface Mount
Operating Temperature:	Supplier Device Package:
-40°C ~ 85°C (TA)	16-CSP (1.95x1.95)
Package / Case:	Base Product Number:
16-WFBGA, WLCSP	PAM8902

# **Environmental & Export classification**

RoHS Status:	Moisture Sensitivity Level (MSL):
ROHS3 Compliant	1 (Unlimited)
REACH Status:	ECCN:
REACH Unaffected	EAR99
HTSUS:	



## A Product Line of Diodes Incorporated



NOT RECOMMENDED FOR NEW DESIGN USE PAM8902H

**PAM8902A** 

# 30-V<sub>PP</sub> MONO CLASS-D AUDIO AMPLIFIER for PIEZO/CERAMIC SPEAKERS with ANTI-SATURATION/CLIPPING

## **Description**

The PAM8902A is a mono, Class-D audio amplifier with integrated boost converter designed for piezo and ceramic speakers. The PAM8902A is capable for driving a ceramic/piezo speaker with  $30V_{PP}$  (10.6 $V_{RMS}$ ) from a 3.6V power supply. The boost converter operates at a fixed frequency of 1.5MHz and provides a 17.5V supply with a minimum number of external components.

The PAM8902A provide three fixed-gain settings of 18dB, 22dB, and 26dB. PAM's anti-saturation technology detects if the input has been over-driven and automatically decreases the gain to prevent the output signal from clipping. To improve fidelity, an integrated low-pass filter rejects any high-frequency noise. The PAM8902A also includes thermal, current limiting, under-voltage lockout, and over-voltage protection circuitry.

Package options include a 1.95mm x 1.95mm CSP and a 4mm x 4mm QFN.

#### **Features**

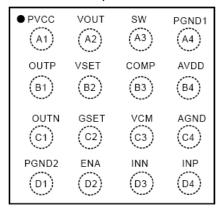
- Supply Voltage Range from 2.7V to 5.5V
- 30V<sub>PP</sub> Boost Converter Output Voltage from a 3.6V Supply
- Selectable Boost Output Voltage of 8V, 12V, and 17.5V
- Integrated Boost Converter Generates 17.5V Supply
- Programmable Soft-Start
- Anti-Saturation Function Eliminates Clipping
- Selectable Gain of 18dB, 22dB, and 26dB
- Low Shutdown Current: < 1µA</li>
- Build in Thermal, Over-Current Protection, Under-Voltage Lock-Out and Over Voltage Protection
- Available in Space Saving CSP and QFN4x4 Packages

## Applications

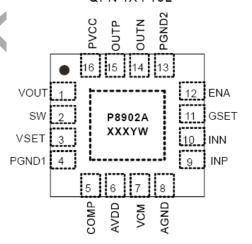
- Wireless or Cellular Handsets
- Portable DVD Player
- Personal Digital Assistants (PDAs)
- Electronic Dictionaries
- Digital Still Cameras

## **Pin Assignments**

### 16 Ball CSP Top View

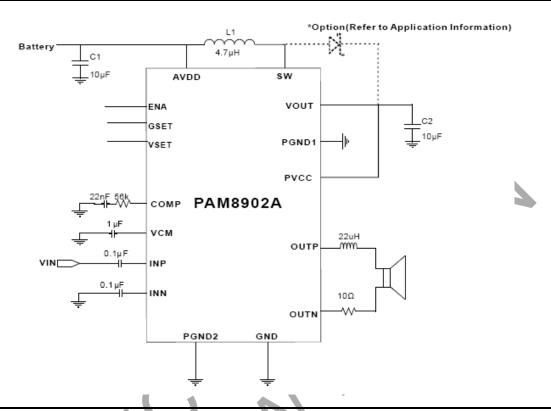


#### Top View QFN 4X4 16L





# **Typical Applications Circuit**

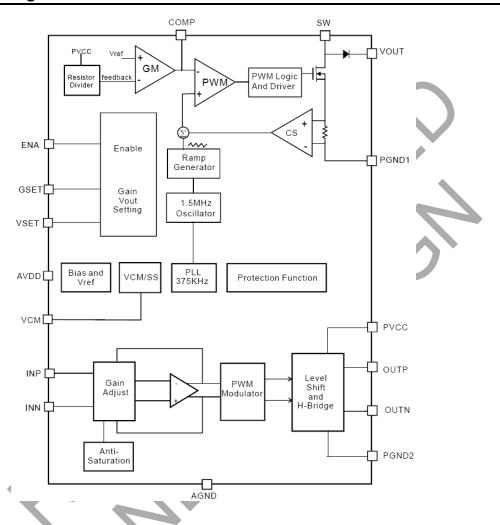


# **Pin Descriptions**

D:				
Pin Name	Bump (CSP)	Pin Number QFN4x4	Function	
PVCC	A1	16	Audio Amplifier Power Supply	
VOUT	A2	1	Boost Converter Output	
SW	A3	2	Boost Converter Switching Node	
PGND1	A4	4	Boost Converter Power Ground	
OUTP	B1	15	Positive Differential Audio Output	
VSET	B2	3	Boost Converter Output Voltage Setting(8V,12V,17.5V)	
COMP	B3	5	Boost Converter Compensation	
AVDD	B4	6	Power Supply	
OUTN	C1	14	Negative Differential Audio Output	
GSET	C2	11	Amplifier Gain Setting (Refer to "Application Information")	
VCM	C3	7	Common Mode Bypass Cap	
AGND	C4	8	Analog Ground	
PGND2	D1	13	Class D Power Ground	
ENA	D2	12	Whole Chip Enable	
INN	D3	10	Negative Differential Audio Input	
INP	D4	9	Positive Differential Audio Input	



## **Functional Block Diagram**



# Absolute Maximum Ratings (@TA = +25°C, unless otherwise specified.)

These are stress ratings only and functional operation is not implied. Exposure to absolute maximum ratings for prolonged time periods may affect device reliability. All voltages are with respect to ground.

Parameter	Rating	Unit
Supply Voltage	6.0	V
Input Voltage	-0.3 to V <sub>DD</sub> +0.3	V
Maximum Junction Temperature	+150	
Storage Temperature	-65 to +150	°C
Soldering Temperature	250, 10 sec	

## Recommended Operating Conditions (@T<sub>A</sub> = +25°C, unless otherwise specified.)

Parameter	Rating	Unit
Supply Voltage Range	2.5 to 5.5	V
Ambient Temperature Range	-40 to +85	°C
Junction Temperature Range	-40 to +125	°C







# **Thermal Information**

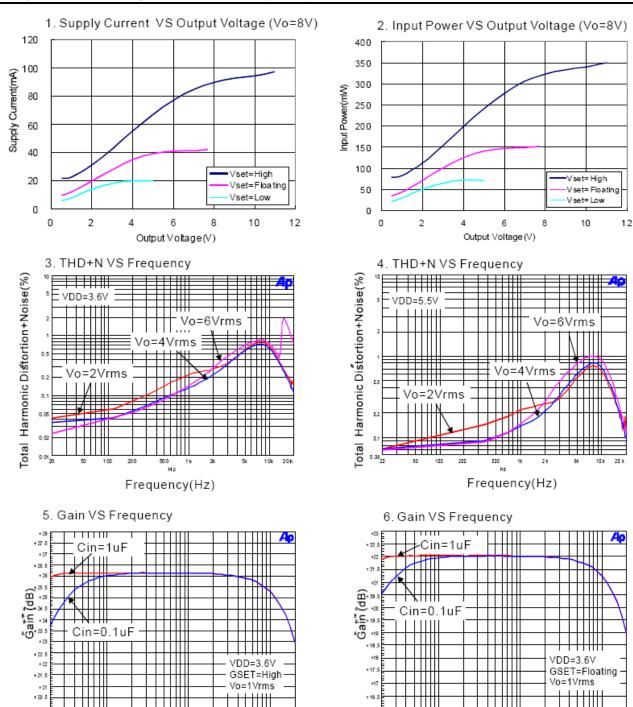
Parameter	Package	Symbol	Max	Unit
Thermal Decistance (Junction to Ambient)	CSP	0	90 – 220	
Thermal Resistance (Junction to Ambient)	QFN4x4-16	θ <sub>JA</sub>	37	°C/W
Thermal Decistance (Junction to Cook)	CSP	0	75	*C/VV
Thermal Resistance (Junction to Case)	QFN4x4-16	θ <sub>JC</sub>	2	]

# **Electrical Characteristics** (@ $T_A$ = +25°C, $V_{DD}$ = 3.6V, $C_L$ = 1 $\mu$ F, $V_{SET}$ floating, unless otherwise specified.)

Parameter	Symbol	Test Conditions	Min	Тур	Max	Units
Input Voltage	$V_{DD}$		2.7		5.5	V
		EN > 1.2V, V <sub>SET</sub> = High		30	48	
Quiescent Current	IQ	EN > 1.2V, V <sub>SET</sub> = Floating		10	18	mA
		EN > 1.2V, V <sub>SET</sub> = GND		5	12	
Shutdown Current	I <sub>SD</sub>	EN = 0V		0.1	1	μΑ
Wake-Up Time	Twu	EN from Low to High		40		mS
Chip Enable	V <sub>EH</sub>		1.2			V
Chip Disable	$V_{EL}$				0.4	V
GSET/ VSET High	$V_{H}$		V <sub>DD</sub> -0.5		$V_{DD}$	
GSET/ VSET Floating	V <sub>F</sub>		1		V <sub>DD</sub> -1	V
GSET/ VSET Low	VL		0		0.5	
Under Voltage Lockout Threshold	UVLO	V <sub>DD</sub> from High to Low		2.2		.,
Under Voltage Lockout Hysteresis	UVLO(H)	V <sub>DD</sub> from Low to High		0.2		V
Thermal Shutdown Threshold	OTP			150		°C
Thermal Shutdown Lockout Hysyeresis	OTP(H)			30		°C
Boost Converter						
	Vo1	V <sub>SET</sub> = Low, No Load	7.2	8.0	8.8	V
Output Voltage	V <sub>O</sub> 2	V <sub>SET</sub> = Floating, No Load	10.8	12.0	13.2	V
	V <sub>0</sub> 3	V <sub>SET</sub> = High, No Load	16.0	17.5	19.0	V
Current Limit	C <sub>L</sub>	Average Input Current		0.8		Α
Lowside MOSFET RDS(ON)	R <sub>LS</sub>	I <sub>O</sub> = 50mA		0.5		Ω
Boost Switching Frequency	fosca		1.1	1.5	1.9	MHz
Over Voltage Protection Threshold	V <sub>OVP</sub>			18		V
Class D			•			•
Class D Amplifier Switching Frequency	fosco	Input AC-GND	225	375	475	KHz
Common Mode Reject Ratio	CMRR	$V_{IN} = \pm 100 \text{mV}, V_{DD} = 3.6 \text{V}$		60		dB
Output Offset Voltage	Vos	Output Offset Voltage		5	50	mV
Parama	D-	High Side		1.5		Ω
R <sub>DS(ON)</sub>	R <sub>P</sub>	Low Side		0.6		Ω
	A <sub>V</sub> 1	$G_{SET} = High, V_O = 1V_{RMS}$	25	26	27	
Closed-Loop Voltage Gain	A <sub>V</sub> 2	$G_{SET}$ = Floating, $V_O$ = $1V_{RMS}$	21	22	23	dB
	A <sub>V</sub> 3	G <sub>SET</sub> = Low, V <sub>O</sub> = 1V <sub>RMS</sub>	17	18	19	
Power Supply Reject Ratio	PSRR	200m V <sub>PP</sub> Supply Ripple @ 217Hz		70		dB
Total Harmonic Distortion Plus Noise	THD+N	V <sub>O</sub> = 5VRMS		0.3		%
Signal to Noise Ratio	SNR	Input AC Ground, A-Weighting		90		dB



**Typical Performance Characteristics** (@ $T_A = +25^{\circ}C$ ,  $V_{DD} = 3.6V$ ,  $CL = 1\mu F$ ,  $L = 4.7\mu H$ ,  $V_{SET}$  floating,  $R_{LOAD} = 10\Omega$ ,  $L_{LOAD} = 22\mu H$  with external schottky diode, unless otherwise specified.)

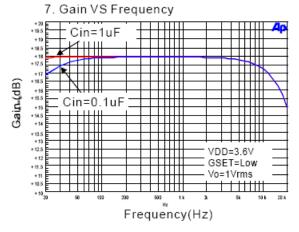


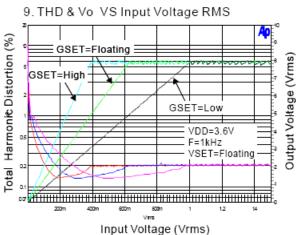
Frequency(Hz)

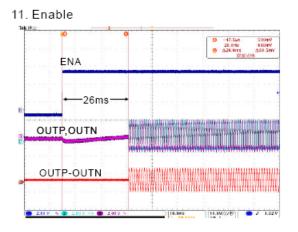
Frequency(Hz)

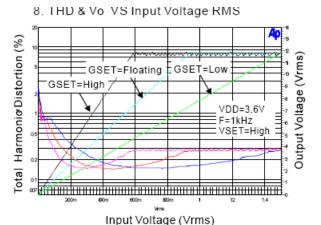


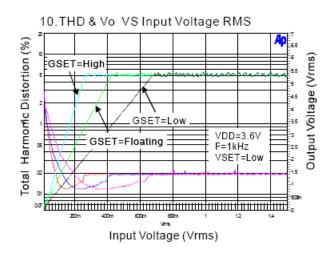
**Typical Performance Characteristics** (cont.) (@ $T_A = +25$ °C,  $V_{DD} = 3.6$ V,  $CL = 1\mu F$ ,  $L = 4.7\mu H$ ,  $V_{SET}$  floating,  $R_{LOAD} = 10\Omega$ ,  $L_{LOAD} = 22\mu H$  with external schottky diode, unless otherwise specified.)

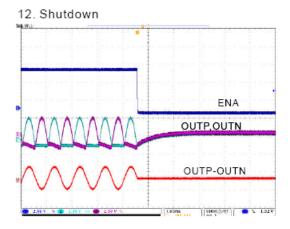














## **Application Information**

#### **Select Boost Converter Output Voltage**

Customer can use V<sub>SET</sub> pin to set boost converor output voltage between 8V, 12V and 17.5V. V<sub>SET</sub> pin configuration table as below:

V <sub>SET</sub> Pin Configuration	Min	Max	PVCC Voltage	Audio Amplifier Maximum Output Voltage
High	AVDD – 0.5	AVDD	17.5V	11 V <sub>RMS</sub> (V <sub>PP</sub> = 31.1V)
Floating	1V	AVDD – 1V	12V	8 V <sub>RMS</sub> (V <sub>PP</sub> = 22.6V)
Low	GND	0.5V	8V	5 V <sub>RMS</sub> (V <sub>PP</sub> = 14.1V)

#### Gain Setting and Input Resistance (R<sub>I</sub>)

Gain setting function is only available in disable condition (ENA = low) and need to follow the sequence as pull ENA to low (disable the IC) first, and change the GSET voltage between high, floating, low, then pull ENA to high (enable the IC).

The input resistors (R<sub>I</sub> = R<sub>IN</sub> + R<sub>EX</sub>) set the gain of the amplifier according to Equation 1 when anti-saturation is inactive

G = 20 Log [1	2.8*R <sub>F</sub> / (R <sub>IN</sub> + R <sub>Ei</sub>	x)] (dB)

G <sub>SET</sub>	R <sub>IN</sub>	R <sub>FB</sub>
High	77.4kΩ	122.6kΩ
Floating	100kΩ	100kΩ
Low	122.6kΩ	77.4kΩ

Where  $R_{IN}$  is a 77.4K $\Omega$  internal resistor,  $R_{EX}$  is the external input resistor,  $R_F$  is a 122.6K $\Omega$  internal resistor. Resistor matching is very important in fully differential amplifiers. The balance of the output on the reference voltage depends on matched ratios of the resistors. CMRR, PSRR, and cancellation of the second harmonic distortion diminish if resistor mismatch occurs. Therefore, it is recommended to use 1% tolerance resistors or better to keep the performance optimized. Matching is more important than overall tolerance. Resistor arrays with 1% matching can be used with a tolerance greater than 1%.

Place the input resistors very close to the PAM8902A to limit noise injection on the high-impedance nodes. For optimal performance the gain should be set to lower. Lower gain allows the PAM8902A to operate at its best, and keeps a high voltage at the input making the inputs less susceptible to noise. In addition to these features, higher value of R<sub>I</sub> minimizes pop noise.

#### **Anti-Saturation Function**

Anti-saturation is active by detecting the duty cycle of the PWM output. When the mode is detected, the gain is automatically adjusted to the value that the output is not clip step by step. The attack time is 100µs per step and the release time is 186ms per step. Following table shows the anti-saturation variable description.

## PAM8902A Anti-Saturation Variable Description

Variable	Description	Value
Gain	The original gain of the device when the Anti-saturation is inactive. The fixed gain is also the initial gainwhen the device comes out of shutdown mode.	26dB ( $G_{SET}$ = High) 22dB ( $G_{SET}$ = Floating) 18dB ( $G_{SET}$ = Low)
Attenuation Range	The gain control range when trigger anti-saturation function.	-26dB (G <sub>SET</sub> = High) -22dB (G <sub>SET</sub> = Floating) -18dB (G <sub>SET</sub> = Low)
Step Size	Gain adjust step size when attack and release.	0.25dB/Step
Attack Time	The minimum time between two gain decrements.	100µs
Release Time	The minimum time between two gain increments.	186ms



## **Application Information (cont.)**

#### Input Capacitors (C<sub>I</sub>)

In the typical application, an input capacitor,  $C_I$ , is required to allow the amplifier to bias the input signal to the proper DC level for optimum operation. In this case,  $C_I$  and the minimum input impedance  $R_I$  form is a high-pass filter with the corner frequency determined in the follow equation:

$$f_C = \frac{1}{2_{\Pi} R_I C_I}$$

It is important to consider the value of  $C_l$  as it directly affects the low frequency performance of the circuit. For example, when  $R_l$  is 150k and the specification calls for a flat bass response are down to 150Hz.

Equation is reconfigured as followed:

$$C_{I} = \frac{1}{2_{\Pi} R_{IFC}}$$

When input resistance variation is considered, the  $C_I$  is 7nF, so one would likely choose a value of 10nF. A further consideration for this capacitor is the leakage path from the input source through the input network ( $C_I$ ,  $R_I + R_F$ ) to the load. This leakage current creates a DC offset voltage at the input to the amplifier that reduces useful headroom, especially in high gain applications. For this reason, a low-leakage tantalum or ceramic capacitor is the best choice. When polarized capacitors are used, the positive side of the capacitor should face the amplifier input in most applications as the DC level is held at  $V_{DD}$  /2, which is likely higher than the source DC level. Please note that it is important to confirm the capacitor polarity in the application.

## **Decoupling Capacitor**

The PAM8902 is a high-performance CMOS audio amplifier that requires adequate power supply decoupling to ensure the output total harmonic distortion (THD) as low as possible.

The optimum decoupling is achieved by using two different types of capacitors that target on different types of noise on the power supply leads. For higher frequency transients, spikes, or digital hash on the line, a good low equivalent series-resistance (ESR) ceramic capacitor, typically  $1\mu F$  is placed as close as possible to the device AVDD pin for the best operation. For filtering lower frequency noise signals, a large ceramic capacitor of  $10\mu F$  or greater placed near the AVDD supply trace is recommended.

#### **External Schottky Diode**

Use external schottky diode can get the best driving capability and efficiency.

Since internal power diode has limited driving capability, only in following conditions customer can remove the external schottky diode to reduce the cost.

- 1.  $V_{SET}$  = Low or Floating and  $C_L$  less than  $1\mu F$
- 2. The signal frequency less than 4KHz.
- 3. Haptic application (50-500Hz).

#### **Shutdown Operation**

In order to reduce power consumption while not in use, the PAM8902A contains shutdown circuitry amplifier off when a logic low is placed on the ENA pin. By switching the ENA pin connected to GND, the PAM8902A supply current draw will be minimized in idle mode.

### **Under-Voltage Lock-Out (UVLO)**

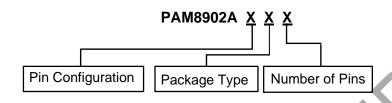
The PAM8902A incorporates circuitry designed to detect supply voltage. When the supply voltage drops to 2.2V or below, the PAM8902A goes into a state of shutdown, and the device comes out of its shutdown state and restore to normal function only when reset the power supply or ENA pin.

#### **Over-Temperature Protection (OTP)**

Thermal protection on the PAM8902A prevents the device from damage when the internal die temperature exceeds +150°C. There is a 15°C tolerance on this trip point from device to device. Once the die temperature exceeds the set point, the device will enter the shutdown state and the outputs are disabled, in this condition both OUTP and OUTN will become high impedance. This is not a latched fault. The thermal fault is cleared once the temperature of the die decreased by 30°C. This large hysteresis will prevent motor boating sound well and the device begins normal operation at this point with no external system interaction.



# **Ordering Information**



Part Number	Part Marking	Package Type	Standard Package
PAM8902AZER	BN YW	CSP-16L	3000Units/Tape&Reel
PAM8902AKER	P8902A XXXYW	QFN4x4-16L	3000Units/Tape&Reel

## **Marking Information**

### 16 Ball CSP Top View

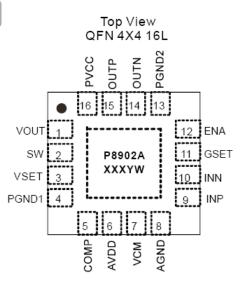
● PVCC (A1)	VOUT	sw (A3)	PGND1
OUTP	VSET	COMP	AVDD
(B1)	(B2)	(B3)	(B4)
OUTN	GSET	VCM	AGND
(C1)	(C2)	(C3)	(C4)
PGND2	ENA	INN	INP
	(D2)	(D3)	(D4)

# Marking

BN YW

BN: Product Code of PAM8902A

Y: Year W: Week



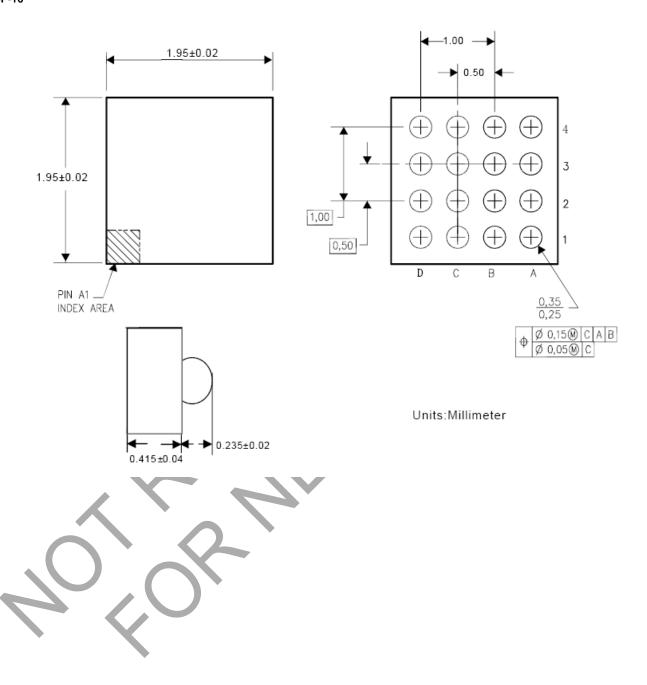
Y: Year W: Week

X: Internal Code



# Package Outline Dimensions (All dimensions in mm.)

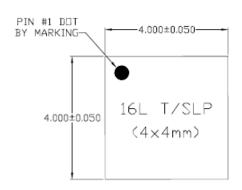
CSP-16



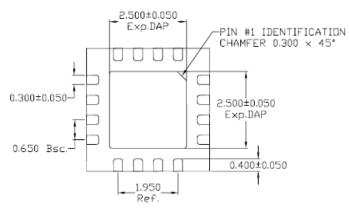


## Package Outline Dimensions (cont.) (All dimensions in mm.)

### QFN4x4-16





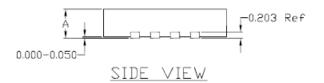


BOTTOM VIEW

## NOTE:

1) TSLP AND SLP SHARE THE SAME EXPOSE DUTLINE BUT WITH DIFFERENT THICKNESS:

		TSLP	SLP
Α	MAX.	0.800	0.900
	N□M.	0.750	0.850
	MIN.	0.700	0.800









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